

# BUILD THIS HI-FI NOISE REDUCER 



## THE NEW


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## UNBEATABLE PRICE/ PERFORMANCE RATIO

 The Model 7000 Universal Counter/Timer

Built for your budget, the Model 7000 is a micro-processor controlled reciprocal universal counter. It is capable of measuring both input signal frequency and period over the full 5 Hz to 80 MHz range in one second with six digit resolution. The autoranging unit has both frequency and multiple period averaging measurement capability. Its microprocessor executes the optimum measurement and displays the desired format, frequency or period.

1. High resolution, $\mu \mathrm{P}$ controlled reciprocal counting design provides both input signal frequency and period measurements.
2. 80 MHz frequency measurement plus event counting to 1 bil-
lion and elasped time measurement from $100 \mu \mathrm{~S}$ to 100 hours.
3. Single function knob for easier operation and built in self testing confidence test circuit.
Its ease of operation, versatility and accuracy make the Model 7000 an ideal instrument for the hobbyist, technician or engineer. The Model 7000 can also be ordered with a temperature-compensated oscillator for applications where higher accuracy is needed.
See your Triplett distributor, Mod Center or representative for a free no-obligation demonstration. Triplett Corp., Bluffton, Ohio 45817. (419) 358-5015, TWX (810) 490-2400.

Triplett performance... a tough act to follow


It started with the Bone Fone. And this very unusual stereo system has created a whole new series of products.

The Bone Fone is an AM/FM stereo radio that drapes around your neck like a scarf. Two speakers, placed near your ears, not only provide excellent stereo separation, but vibrate slightly through your bones to give you the same sensation as standing in front of your home stereo system.

## UNEXPECTED APPLICATIONS

Shortly after it was introduced, the Bone Fone became a very popular product for a variety of reasons. A lady in Helena, Montana who bought the unit for her son told us, "It's made a significant contribution to my sanity. No more rock $n$ ' roll blasting through the house, the sound goes where my son goes."

A jogger in Rowlett, Texas wrote us "Amazing separation, fantastic stereo response, helps my jogging tremendously. I wasn't really expecting this type of quality through a magazine ad at this price."

But one of the most unexpected letters came from a man in Belle Center, Ohio. "You don't have to be young and jog to enjoy Bone Fone. You see, I'm 73 years old. I just sit and listen."

## LETTERS EVERYWHERE

Letters have come from mailmen, roller skaters, skiers, cyclists, motrocycle enthusiasts, hikers and even people who listen to the Bone Fone stereo while walking their dog. The Bone Fone appeals to practically every American.

The Bone Fone was designed by an engineer who wanted to listen to good stereo music without carrying heavy box radios or bulky headphones. Headphones block out all other sounds-even warnings which could be dangerous outdoors, and box radios are heavy and disturb those around you. So he invented the Bone Fone-"the stereo sound you wear around.'

Weighing only 17 ounces and powered by

4AA cell batteries the Bone Fone stereo provides a sound that would be impossible to describe in an advertisement. The cliche, "you've got to hear it to believe it," certainly applies here. And for $\$ 69.95$ it's the lowest priced stereo entertainment product available.

But what about the sport enthusiast who can care less about stereo music? Or the person who wants just the news? Or simply the person that just listens to AM radio and doesn't want to spend $\$ 69.95$ ?


The Bone Fone drapes around your neck like a scart and has a sound that you find incredible when you first hear it.

Enter NUTS! NUTS is the AM version of the Bone Fone for sports nuts, news nuts, jogging nuts or anybody who wants a low cost Bone Fone without FM or stereo. NUTS sells for $\$ 39.95$ complete with two speakers and a strap that firmly attaches the unit to you for any physical activity.
Sitting at a football game, walking your dog, jogging-NUTS gives you a convenient way to listen to music, news and sports without paying a premium for stereo.

But the Bone Fone spinoffs don't end there. There's the Neck Fone-a device you place over your shoulders and plugs into your home stereo system. This lets you enjoy your home stereo without disturbing those around you and without the bulk of headphones. The Neck

Fone sells for $\$ 34.95$.
So there you have it. Three exciting products-Bone Fone, NUTS, and the Neck Fone-three unusual solutions designed to solve any gift-giving problem.

## LOWEST-PRICED STEREO

Compare the Bone Fone price with any box radio, stereo system or even the new $\$ 200$ Sony Walkman. The Bone Fone is the lowestpriced quality personal stereo system you can buy. It is also safer than headphones as it leaves you free to hear the sounds around you and keeps you in touch with the environment.
To order any of the above products, simply send your check or money order for the amount listed above plus $\$ 2.50$ for postage and handling (III. residents add $6 \%$ sales tax) to the address below, or credit card buyers may call our toll-free number below. Each unit is backed by a 90 -day limited warranty and a service-by-mail facility as close as your mailbox. Service should rarely be required as the units use solid-state components and are designed to take rugged treatment. JS\&A is America's largest single source of space-age products-further assurance that your modest investment is well protected.
The Bone Fone started a small revolution. Be part of that revolution with the space-age way to listen to music, news and sports. Order a Bone Fone product at no obligation, today.


Dept.RA One JS\&A Plaza Northbrook, III. 60062 (312) 564-7000 Call TOLL-FREE ........ 800 323-6400 In Illinois Call . . . . . . . . . . (312) 564-7000 © JS\&A Group. Inc., 1980

# Beckman brings a new dimension to hand held Digital Multimeters 



## True RMS capability at an affordable price

Now you can measure the exact power content of any signal - regardless of waveform. Beckman delivers the new TECH ${ }^{\text {TM }} 330$ multimeter with true RMS capability and many more fine performance features for just $\$ 210$.
Unlike most multimeters calibrated to read only the true power content of sine waves, the TECH 330 extends its true RMS capability to give you accurate readings of both sine and non-sine waveforms.

True RMS makes a significant difference in accuracy when measuring switching power supplies, flyback power circuits, SCR or TRIAC controlled power supplies or any other circuit generating a non-sine signal.

The TECH 330 also accurately measures the entire audio band up to 20 kHz . But that's not all you can expect from Beckman's top-of-the-line multimeter.

| Waveforms (Peak $=1$ Volt) | Average Responding Meter | $\begin{aligned} & \text { Beckman } \\ & \text { TECH } 330 \end{aligned}$ | Correct <br> Reading |
| :---: | :---: | :---: | :---: |
| Sine Wave | 0.707 V | 0.707 V | 0.707V |
| Full Wave Rectified Sine Wave 0 $\qquad$ | 0,298V | 0.707V | 0.707 V |
| Half Wave Rectified Sine Wave $0$ $\qquad$ ค $\square$ $\square$ | 0.382 V | 0.500 V | 0.500 V |
| Square Wave <br>  | 1.110V | 1,000V | 1.000 V |
| Triangular Sawtooth Wave - WWWMW | 0.545 V | 0.577 V | 0.577 V |

You also get $0.1 \%$ basic dc accuracy, instant continuity checks, 10 amp current ranges, a separate diode test function, 22 megohm dc input impedance, and an easy-to-use rotary switch.

With so much capability in hand, you'll be able to depend on the TECH 330 for a long time. That's why Beckman designed it tough enough to go the distance.

Enclosed in a rugged water-resistant case, the TECH 330 can take a 6 -foot fall onto concrete and still perform up to spec. And to further ensure reliable, trouble-free operation, the TECH 330 gives you 1500 Vdc overload protection, RF shielding, 2000-hour battery life, gold switch contacts, and fewer electronic components to worry about.

Add another dimension to your world of electronics. Visit your Beckman distributor today for more information on the TECH 330 and Beckman's complete line of digital multimeters, starting at \$120.
For your nearest distributor, or a free brochure:

## CALL TOLL FREE

## 24 HOURS A DAY, 7 DAYS A WEEK

1-(800)-821-7700 (ext. 517)
in Missouri 1-(800)-892-7655 (ext. 517)
BECKMAN

## ON THE COVER

The ASRU (Audio Signal Restoration Unit) is a combination noise-reduction and signal-expander device that offers features not even found on some commercial units. Build one yourself and hear things from your records and tapes that you never heard before. The first part of this project begins on page 41.


FIND OUT HOW laser videodisc players handle the complex signals inscribed on those shiny platters. The story starts on page 67.


UNMARKED IC's can be a bargain. Learn how to find out what's inside those plain black packages on page 55.

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## TYPEWRITER WILL TRAVEL



Combining high-density magnetic recording, miniaturization and consumer-electronics design principles, Sony has introduced what could be the forerunner of a major and basic new kind of product-a noiseless, completely portable electronic typewriter. Sony's Typecorder actually is far more than a typewriter. In a slim $8^{1 / 2} \times 11 \times 1^{1 / 2}$ inches, weighing only three pounds, it combines a word-processing terminal, a dictating machine and a stenotype system. A built-in microcassette can store up to 120 pages of typed material and make sound recordings. There is a 40 -character (half-line) LCD readout on the machine.
A portable printer makes the Typecorder a complete out-of-office typing system, or with various peripherals, it can be integrated into word-processing systems. An interface lets it operate any IBM Selectric typewriter, either directly or via phone line. Another accessory will punch out a Telex tape from its cassette, or it can be interfaced as a remote terminal to any word-processing system via a modem. The Typecorder will be available around midyear at $\$ 1400$. The portable printer accessory will be about $\$ 600$. Typecorder's debut marks Sony's entry into the office-equipment word-processing business. A new Sony line of word-processors is geared to a new $3^{1 / 2}$-inch floppy-disk drive, with a recording density 1.47 times that of a 5.25 -inch floppy disk. Sony's Series- 35 word-processing equipment also employs microcassettes and is compatible with the Typecorder.

Is video the next wave in hi-fi? Apparently many audio manufacturers think so. Advent was the first audio manufacturer to go into video, through its VideoBeam projection system. Pioneer is now making and selling the LaserDisc optical videodisc system and will add a vido projector. Who's next? Well, Fisher will have a deluxe console containing audio equipment, an optical disc player and Beta VCR, scheduled for next fall. And Sansui plans to bring out a VHD videodisc player later this year, followed by a VCR. Superscope is expected to go into video under the Marantz brand, possibly bringing out both a projection TV set and a videodisc player. Advent will add an optical videodisc player, and both Kenwood and Aiwa are studying possible entry into the video market in the U.S.

This is the month that RCA's CED SelectaVision videodisc player makes its debut-scheduled to go on sale the week of March 22 in 5,000 stores coast-to-coast, backed by a record advertising campaign on television, in newspapers and magazines. RCA already has trained technicians in 1,000 company-owned and independent service shops. The number of players on hand for the start of sale will be at least 20,000 and possibly as high as 30,000 . The player will carry a suggested list price of $\$ 499.95$, and the initial disc catalog will contain 100 titles, selling at $\$ 14.98$ to $\$ 27.98$, but most of the discs will sell for under $\$ 20$ with the highest-priced selections containing more than one two-hour disc. Movies will be the mainstay of RCA's first offerings, although there will also be classic TV shows, sports, and music, as well as children's and educational selections. RCA says it will offer 25 new titles in May; 25 in September, and 10 monthly thereafter. RCA's disc-ad campaign will stress programming, simplicity of operation, and low cost.

Projection TV, too, is sweeping the American television industry, seven years after Advent popularized the first home color-projection system. Major holdout RCA has introduced a single-piece three-tube unit, to be followed soon by Zenith, Magnavox, Sylvania and Curtis Mathes, with giant-screen sets now fielded by virtually all American TV brands and most Japanese ones. Sears Roebuck added its own projector late last year. Three-tube projection sets are also being offered by GE, Sony, Quasar, Panasonic, Toshiba, MGA and Kloss Video, with Hitachi, Sanyo and Toshiba expected to come along soon. Sales of specially built TV projectors (as opposed to modifications of small-screen sets) totaled about 28,000 in 1979, rising to nearly 50,000 in 1980 and are expected to reach 75,000 this year.

With all of the new video information sources available to us, will we soon be required to watch and listen at a faster pace to take it all in? VSC Corporation thinks we will, and it is forecasting that the next generation of home VCR's will incorporate its new variable speech-control IC which permits speeded-up audio without a change in pitch. Although most VCR brands now offer special-effects models that permit high-speed viewing, all except JVC disable the audio. One JVC model contains the VSC IC and allows watching and listening to programs at less time than it took to record them. Just think of all the extra viewing you can do if you can watch 60 Minutes in 30 .

## Magnavision is Gourmet Video. Video for people who know and love video.

If you seek the ultimate in your electronic gear, Magnavox has a bright idea for you called Magnavision. It is Gourmet Video for the video gourmet. A picture that's clearer than tape and less costly, too.

Magnavision is an advanced LaserVision ${ }^{\text {TM }}$ videodisc player. Its optical laser scanner, a videodisc and your TV set team up to give you a picture that's amazingly sharp and clear.

Even better, the Magnavision picture remains this good even after thousands of viewings. That's because there is no direct contact between our laser and the disc. Unlike your phonograph. Magnavision doesn't use a needle.


O 1963 ALFRED $J$. HITCHCOCK PRODUCTIONS. INC

The hearing's as good as the seeing.

Magnavision is designed to be played through your home stereo system so you hear what you see in full high-fidelity stereophonic sound. And since there is no disc
wear, the Magnavision sound stays crystal clear, playing after playing Studio-like controllability puts you in command of the action.

Now the real fun begins. You not only watch and hear Magnavision. You play with it, too. Reverse, Slow Motion, Still, Fast Forward, Search, Numerical Index, Stereo Sound. Only LaserVision systems like Magnavision let you watch and play so many different ways.
Watch what you want whenever you want.

With Magnavision, you have a complete library of MCA DiscoVision ${ }^{*}$ programming to choose from. Blockbuster movies like The Electric Horseman. Classic films like The Bride of Frankenstein. Cooking lessons by Julia Child. Documentaries from Jacques Cousteau. How-to-do-it tennis, golf, swimming and crafts. Music, concerts, cartoons, the arts and NFL football.

Discover Gourmet Video today. Call toll-free 800-447-4700 for the Magnavision dealer nearest you. In Illinois, call 800-322-4400.

## The brightest ideas in the world are here to play.

# What's news 

## Encyclopedia on one disc?

Drs. Alan E. Bell and Robert A. Bartolini of RCA Labs, Princeton, NJ, have received U.S. patent for a recording method that makes it possible to put 100 billion bits of information on the two sides of a single disc. That is 10 to 100 times the capacity of any magnetic-storage disc used today.
The information is recorded with an intense beam from a novel semiconductor laser, which burns a series of microscopic holes in a thin tellurium layer deposited in the coating structure of the plastic disc. The information is read out by a less intense beam from the same laser. It shines through the microscopic holes and is reflected by a layer of aluminum located below the tellurium.

The high-density system has many po-
tential applications. A multi-volume encyclopedia could be stored on the two sides of a single disc. It may also be possible to replace conventional X-ray film, which requires expensive silver. Other possible applications are in word processing, still and motion pictures, and in storing bulky business and government records.

## Video viewers are passive

The average video viewer of prospective video viewer shows little interest in programs that call for viewer participation, such as do-it-yourself programs. Those interest only 15.9 percent of the potential audience, while new movies would attract 66.5 percent, or two-thirds, according to a recent survey by Venture Development of Wellesley, MA.


THE HIGH DENSITY DISC is held by Dr. Bell. Dr. Bartolini is at left.

Movie classics are next highly regarded, with a 50.2 percent rating, and PBS series and specials follow with 31.6 percent.

Between 20 and 30 percent of the prospective audience would be interested in pop, rock, and jazz concerts, educational courses, plays and dramatic specials, sports events, and old television series. Comedy specials, classical music concerts, and news and documentaries all rank between 14 and 20 percent, while musicalvariety specials would interest only 14 percent. At the bottom of the list are children's programs, sports lessons, and foreign-language movies, all with ratings lower than 10 percent.

## Power savings for UHF TV

A committee of nineteen Public Television engineers have reported that the efficiency of UHF stations using klystron transmitting tubes can be increased to make a power saving of nearly 50 percent with no decrease in output, or with a considerable increase in output if input power is maintained. The statement was backed up with a 67-page report describing experiments at several television stations and with a transmitter manufacturer.

One of the techniques is to cut the beam perveance, reducing beam current while raising the voltage to maintain the same output power. Another is to pulse the modulation anode-the element in a klystron that corresponds most closely to the grid in a triode tube-during the TV sync periods. That increases efficiency notably. In effect, the power of the tube is increased during the sync pulses.
Greater efficiency was also achieved by using a system of tuning, or "alignment" developed by the BBC, and now used by most European stations. A significant amount of power was also saved by cutting the ratio of audio to video power from the conventional 20 to 10 percent and installing a special audio coupler that would increase efficiency.
Because UHF stations require several times as much power as VHF stations, savings on power can mean a significantly greater net income to many UHF TV stations, most of which use klystrons.

## Allocated territories for Tronics 2000

Tronics 2000-a new nationwide franchising organization designed to give the independent service shop owner the prestige that small business proprietors in other fields have achieved through national franchise organizations - has sold its first three territories, during its first three weeks of operation.

Robert P. Neal, of Able Electronics, Waukegan, IL, has purchased a master francontinued on page 12

## Direct readings in decibels: Keeping track of your gains and losses.

If you'd rather forget about the last time you got wrapped up in an audio jungle, you'll want to respond to this ad.

Meet our new $4^{1 / 2}$-digit Model 8050A Multimeter - the first low-cost DMM with self-calculating dB features that let you keep your mind on your mission instead of on conversions and formulas.

While most analog meters read dBm referenced only to 600 ohms , the Fluke 8050A delivers direct readouts in decibels over a 108 dB range referenced to any one of 16 impedances ( 8 to 1200 ohms) with 0.01 dB resolution.

Push one button, and the microprocessor in the 8050A scrolls
through its reference impedances. Simply stop at the one that matches your system and get back to work. No more math; just action. And with the 8050A's relative reference feature you can measure gains or losses in dB throughout your system faster than you thought possible.

When you're dealing with voltage, current or resistance, an offset function provides a means of comparing stored inputs with all subsequent inputs, automatically displaying the difference. A real timesaver.

And there's more. True RMS measurements to $50 \mathrm{kHz} ; 0.03 \%$ basic dc accuracy; conductance (measures leakage and high resistance); extensive overload protection and safety features; a full line of accessories; and a low price of $\$ 349$ U.S.

For all the facts on how to maximize your gains with the 8050A, call toll free 800-426-0361; use the coupon below;or contact your Fluke stocking distributor, sales office or representative.

## FL- KKE



For technical data circle no. 37

> NRI Iraining inTV and Audio Servicing keeps up with the state of the art. Now you can learn to service video cassette and disc systems.


You build color TV, hi-fi, professional instruments.
Now, in addition to learning color TV and audio systems servicing, you get state-of-the-art lessons in maintaining and repairing video cassette recorders, and the amazing new video disc players, both mechanical and laser-beam types.

## Learn at Home in Your Spare Time

And you learn right at home, at your own convenience, without quitting your job or going to night school. NRI "bite-size" lessons make learning easier...NRI "hands-on" training gives you practical bench experience as you progress. You not only get theory, you actually build and test electronic circuits, a complete audio system, even a color TV.

## Build Color TV with Computer Programming

As part of your training in NRI's Master Course in TV/Audio/ Video Systems Servicing, you actually assemble and keep NRI's exclusive designed-for-learning 25 "
(diagonal) color TV. It's the only one that comes complete with builtin computer tuning that lets you program an entire evening's entertainment. As you build it, you introduce and correct electronic faults, study circuit operation, get practical bench experience that gives you extra confidence.

You also construct a solid-state stereo tuner and amplifier complete with speakers. You even assemble pro-fessional-grade test instruments so you know what makes them tick, too. Then you use them in your course, keep them for actual TV and audio servicing work.

## NRI Includes the Instruments You Need

You start by building a transistorized volt-ohm meter which you use for basic training in electronic theory. Then you assemble a digital CMOS frequency counter for use with lessons in analog and digital circuitry, FM principles. You also get an integrated circuit TV pattern generator, and an advanced design solid-state 5 " trig-gered-sweep oscilloscope. Use them for learning, then use them for earning.

## NRI Training Works... Choice of the Pros

More than 60 years and a million students later, NRI is still first choice in home study schools. A national survey of successful TV repairmen shows that more than half have had home study training, and among them, it's NRI 3 to 1 over any other school.
(Summary of survey on request.) That's because you can't beat the training and you can't beat the value! For hundreds of dollars less than competing schools, NRI gives you


Other NRI training includes Computer Technology, Complete Communications Electronics.
and now includes training in video cassette and disc systems. Send for our free catalog and see for yourself why NRI works for you.

## Free Catalog <br> No Salesman Wili Call

Send today for our free 100page catalog which shows all the kits and equipment, complete lesson plans, and convenient time payment plans for courses to fit your needs and budget. Or explore the opportunities in other NRI home study courses like Microcomputers \& Microprocessors, CB and Mobile Radio, Aircraft and Marine Radio or Complete Communications. Send the postage-paid card today and get a head start on the state of the art. If card has been removed, write to:


NRI Schools
McGraw-Hill Continuing
Education Center
3939 Wisconsin Ave.
Washington, D.C. 20016


## What's news

continued from page 6
chise for three counties with a total population of 615,000. Jim Cardnell, Homosassa Springs, FL, has also purchased a threecounty block, with a population of about 685,000 , and William B. Terrel, Cincinnati businessman, has become the master franchiser of Hamilton County, with a population of around 877,000.

Each of the Tronics 2000 territorial franchisers will offer franchises to a limited number of qualified service centers in their territories, with programs in advertising, volume buying, technical support, and business management.
"The consumer-electronics service industry is one of the largest left unfranchised in the United States," says David Hagelin, President of Tronics 2000 and former publisher of Electronics Technician/ Dealer. "And today independent operations, rightly or wrongly, often mean consumer mistrust.'

The concept that Tronics 2000 has in mind, says Hagelin, is close to what Century 21 has done in the real estate field. "Franchisees will receive assistance in all phases of business management, including training, advertising, promotion, and individual consultation. Advertising will be local and regional, with a national identity symbolized by a bold blue logo that will identify the Tronics 2000 operation to the consumer wherever he goes in the United States."

## 3-dimensional television

The world's first public transmission of 3D television is claimed by 3D Television Systems of North Hollywood, CA. Subscri-
bers to the SelecTV television-cable system saw 3D films for the first time on December 19, 1980.
3D is well known in movies-though it has never become popular-and has been transmitted experimentally on TV. The process developed by 3D Television transfers three-dimensional films electronically to a master 3D videotape. Viewers of SelecTV on Channel 22 were able to see the 3D pictures by wearing stereo glasses similar to those used for 3D movies. Complimentary glasses were made available to SelecTV subscribers before the transmissions.
The system is described as "being able to portray objects coming out of the screen to within several inches of the viewers' eyes and then going deep back into the television screen."

## Test lab for nuclear fusion

The Lawrence Livermore National Laboratory, operated by the University of California, is constructing a Mirror Fusion Test Facility (MFTF) to explore the possibility of producing power by nuclear fusion. To that end, a contact has been let to RCA to produce 48 Neutral Beam projectors for the project.
Producing a plasma in the MFTF requires evacuating a 40 -foot diameter by 60 -foot long cylindrical fusion chamber to an almost perfect vacuum. Thereafter, 20 startup Neutral Beam injectors mounted around the wall of the chamber insert bursts of deuterium atoms to form the target plasma, which is confined in the center of the machine by powerful magnets. Once the plasma is formed, 24 high-powered Neutral


VIEWERS BEING TERRIFIED by the latest development in television

Beam injectors raise the plasma termperature to the more than the 100 -million degrees Celsius required for fusion.

Deuterium is the heavy isotope of hydrogen, in which a neutron as well as proton are contained in the nucleus. Deuterium is available in nearly unlimited quantities through extraction from sea water.
The 48 Neutral Beam injectors to be built by RCA include two spares of each type. Each consists of an arc chamber to ionize the deuterium working gas as well as an accelerator assembly to inject and focus the ion stream, which is later neutralized to form atoms, into the vacuum chamber. Twenty-two of the beam injectors are startup beam sources, rated at 20 -kilovolts accelerator voltage and 100-amperes current with pulse durations of 0.01 second. The remaining 26 devices are sustaining beam sources; they are rated at 80 kilovolts and 80 amperes, with pulse durations of 0.5 second.

Nuclear fusion is expected to provide power generation with improved safety, efficiency, and lower cost compared with present fission (splitting the atom) methods. For those reasons, many engineers believe fusion may be the basis for the nuclear power plants of the next century.

## CB installers are warned

The Electronic Industries Association (EIA) has issued a warning about the danger of electrocution while installing CB or TV antennas. (Through long and bitter experience, many amateurs have learned to keep away from power lines and icy roofs. However, recent casualty figures show that many CB and TV users still have to learn that lesson.)
The EIA recommends that anyone putting up an antenna obtain the free U.S Government Antenna Alert sheet, by writing to Antenna Alert Sheet, Consumer Product Safety Commission, Washington, DC 20207. The Association also reiterates some important points of that document:

Make sure the antenna is twice its
length away from any power or light lines. Remember that even insulated lines are dangerous. Insulation may be worn, or be cut through when struck by an antenna.
If the antenna is more than 30 feet high, or is to be erected in an area where light or power lines are closer than twice the antenna height, better use professional help.
The EIA warning and the government "Alert" confine themselves to electrical dangers. Probably more accidents-with TV antennas at least-consist of falling off roofs. The mast erected from the ground and fastened to the side of the building is many times safer for the non-professional installer.

R-E

## Sabtronics. An entire range of low-cost, top-quality instruments.


A. $2010 \mathrm{~A}-\$ 84.00^{*}$, B. $2015 \mathrm{~A}-\$ 99.00^{*}$, C. $8610 \mathrm{~A}-\$ 99.00^{*}$, D. $5020 \mathrm{~A}-\$ 85.00^{*}$, E. 8000 B $-\$ 199.00^{* *}$, F. $2035 \mathrm{~A}-\$ 79.00^{*}$, G. 2037A $-\$ 99.00^{*}$

Sabtronics revolutionized the market with the first low-cost, high-performance Digital Multimeter. Now we have an entire range of outstanding instruments in a reasonable range of prices. In fact, nobody can beat us in our price/performance ratio. And we can sell at a low price for some very good reasons. Our engineers design high performance products to be built at a low cost. And we refuse to stick on high mark ups. Plus we make sure your price stays low by selling directly to you. Because we sell so many instruments, we don't have to charge a high price. Naturally, we also offer all the helpful accessories you might want. And all our products are under warranty for good quality and high performance. In addition, you get from us the same quality aftersale service as any high priced instrument manufacturer. With Sabtronics instruments available, there's no need for you to spend a lot of money to do highly accurate testing and measuring.
2010A $31 / 2$ Digit LED DMM
2015A 31/2 Digit LCD DMM
8610A 600 MHz 8 -Digit Frequency Counter 8110A 100 MHz 8 -Digit Frequency Counter 5020A 0.2 to 200 KHz Function Generator 8000B 1 GHz 9-Digit Frequency Counter 8610B 600 MHz 9 -Digit Frequency Counter 2035A 3½ Digit LCD Handheld DMM 2037A 31⁄2 Digit LCD Handheld DMM
*price in kit form. Also available factory assembled, tested, and calibrated. Call us for prices.
** price fully assembled, tested, and calibrated.
Call us for more information:
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Making Performance Affordable

5709 N. 50th Street Tampa, FL 33610

## Error... Error... Error... Error...

## THE MISSING LINK

A serious omission was made in Part 2 of the "Pay TV Decoder" article that appeared in the February 1981 issue of Radio-Electronics (page 51). Someone apparently got so excited over the photo of Bo Derek that the first page of Part 1 was printed (again) instead of the first page of Part 2. What got left out was most of the conclusion of the "theory of operation" part of the article. The missing text appears below, and we apologize for any inconvenience this slip may have caused.

IN THE FIRST PART OF THIS ARTICLE WE discussed how pay-TV signals are scrambled and began talking about a device to decode them. Refer to Part 1 and Fig. 3 as we finish the discussion and describe the construction and connection of a decoder board.

Returning to IC1 for a moment, the phase-locked oscillator produces a square-wave output signal at pin 11 that is used to trigger a 74123 dual oneshot (IC2-a and -b). That one-shot produces the gating signal required to restore the sync and blanking pulses. The first section, IC2-a, has a period of approximately 30 microseconds. The trailing edge of its output pulse is used to trigger IC2-b, which has a period of approximately 12 microseconds-the width of the horizontal-blanking pulse. Because of the design of IC1, the $15.75-$ kHz output signal at pin 11 occurs somewhere in the middle of a scan line. An adjustable delay that allows the gating signal to begin at the proper time is provided by IC2-a. The actual gating signal is generated by IC2-b. and its width is set to match the width of the horizontal-blanking interval. The combination of R6-C9 is used to position the gating signal and the combination of R7-C10 is used to control its width. The gating signal appears at pin 13 of IC2-b, and its amplitude is controlled by R8. That signal is used to increase the IF gain of the TV receiver during the horizontal sync- and blank-ing-pulse intervals and thus correct the video waveform.

We have now discussed all of the circuitry required to reconstruct the
original video and audio signals. Now it is necessary to re-combine them into a useable TV signal. We have obtained a stable picture, but still have no sound. However, the intended use of our adapter is to provide a VHF signal for viewing on another, unmodified, TV set. Therefore, the video signal is taken from the video detector and applied to IC4, which is a complete video-modulator IC. The tank-circuit $\mathrm{L} 1-\mathrm{C} 21$ is used to set the video-carrier frequency. Resistor R21 controls the voltage at pins $2,3,4$, and 13 of IC4 which, in turn, controls the percentage of modulation of the video input at pin 12. The sound is generated by FM-ing a $4.5-\mathrm{MHz}$ oscillator using a transistor Photo of Bo Derek copyright 1979
by Orion Picture Company

## NOTE:

The legality of the use of pri-vately-owned devices to decode subscription TV broadcasts is currently the subject of much debate and pending litigation. The subscription companies have taken the position that decoding of broadcasts without payment is "theft of service" and the FCC has issued a notice to the effect that subscription-TV decoders are subject to FCC approval.

This article merely explains how such decoding devices are built and used, and you should obtain independent advice as to the propriety of its use depending upon your individual circumstances.

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Satellite Television Corp., the direct-to-home broadcasting subsidiary of Comsat, has revealed details of its plans to create a high-power pay-TV service that could be in operation by 1986. The three-channel service would offer major motion pictures, popular concerts, children's programs, sports, experimental theater and other shows, beamed directly into homes via a 2.5 -foot dish to be placed atop roofs nationwide. The satellites that will carry the signals will operate in the $17-\mathrm{GHz}$ band on the uplink and be received in the $12-\mathrm{GHz}$ range, with signals beamed via a high power carrier ( 1700 watts). The birds will be of the new PAM-D class, a refinement of the current breed now used by other Comsat subsidiaries.

Satellite TV Corp., which is applying to the FCC for permission to build the new pay-TV satellite system, envisions that subscribers will pay about $\$ 100$ for the reception antennas and then about $\$ 25$ per month for the decoder. It will also be possible to buy the receiving equipment, then pay about $\$ 18$ per month for the programming alone. STC engineers promise that the encoding/scrambling format for the video signal will be so complicated that the programs will be immune from piracy by unauthorized receivers.
STC wants to launch four separate satellites during 1986, each of which would cover a different time zone in the U.S. (with the Western satellite including spot beams for Alaska and Hawaii). The birds would be spaced about $20^{\circ}$ apart in orbital arcs at $115^{\circ}, 135^{\circ}, 155^{\circ}$ and $175^{\circ}$ west longitude. Two other satellites would be used as in-orbit spares in case one of the primary birds encounters technical troubles.

The entire project will cost at least $\$ 400$ million, STC estimates. In addition to the 2.5 -foot dishes atop home roofs, the company expects that apartments and other housing complexesespecially areas without cable TV-will be users of the new pay-TV service.

The FCC, which is currently examining a number of Direct Broadcast Satellite policy options, is expected to begin consideration of the STC plan within the year. In any case, STC admits that it will take at least three years to build and launch the system.

ABUNDANCE OF NEW EQUIPMENT

Lower-priced-and more versatile-satellite reception equipment continues to flood the market, as was evident at several recent industry meetings. Manufacturers serving both the hobbyist market and the low-cost professional user are offering a variety of new antennas and terminal devices. For example, such established companies as Microdyne Corp. (PO Box 7213, Ocala, FL 32672; phone 904-687-4633) are offering new 12-foot antennas, with EIRP contours between 36 and 33 dBW . Hughes has unveiled a new expandable dish, offering a basic 3.7 meter TVRO antenna that can be upgraded to 5 -meter diameter. Hughes says the 3.7 meter version provides $52 \%$ more gain than a standard 3 -meter dish, and that it can be expanded to the 5 -meter configuration without upgrading the foundation or mount structure. (Hughes Microwave Communications Products, PO Box 2999, Torrance, CA 90509; phone 213-517-6100.)

Downlink Inc., a new company, unveiled a modular satellite-TV system during the Consumer Electronics Show, with a promise that it intends to "become the Apple comupter of satellite TV"-a reference to the successful easy-to-use home-computer system. The Downlink package includes a control console with receiver mounted at the antenna, 12 -foot spherical antenna, $120^{\circ}$ low-noise amplifier, feedhorn and rotor with back assemblies, plus 100 feet of cables. (Downlink Inc., PO Box 33, Putnam, CT 06260; phone 203-928-7955.)

Chaparral Communications has introduced a new Feed Horn, designed to optimize the capabilities of parabolic antennas-providing an improvement of at least half a dB of system operation over a conventional rectangular horn. The $\$ 135$ feed horn consists of a standard WR220 waveguide flange and a front plate which is held in place with a set screw. The Chaparrel feed is not intended for use with spherical antennas because of their high focal length-to-diameter ratios (Chaparral, PO Box 832, Los Altos, CA 94022; phone 415-941-1555.)

Cable News Network, the all-news channel (Satcom I, Transponder 14) is carrying an electronic cable-TV guide on a sideband. The slow-scan video listings offer a program directory to what is being carried on cable-TV systems, including a rundown of that day's programs on other satellite-fed cable channels. The slow-scan sideband signals are intended to be used by cable systems on an otherwise blank channel.

Intelsat V, the $\$ 34$ million international satellite, completed its launch sequence and is now slated to go into operation in May as the primary Intelsat bird over the Atlantic Ocean. The big bird ( 4,300 pounds, two video transponders) is the first of nine new Intelsat satellites being built by Ford Aerospace.

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## Jetters

## SWEEP/FUNCTION GENERATOR

Thank you for the comprehensive review of our model 3020 Sweep/Function Generator in your October 1980 "Equipment Reports" section.

Actually, the instrument is even a bit more versatile than your review indicates. An internal gate with an adjustable duty cycle and repetition rate is provided for tone bursts. In addition, provision is made for external gating, as your review indicated. That feature, combined with those you listed, makes the B\&K-Precision model 3020 an exceptional stand-alone instrument. The only external signal requirement is for the amplitude-modulation function. GUS ROSE,
Dynascan Corporation

## THE BSR SYSTEM X-10

Mr. Steve Ciarcia's article on the BSR System X-10 ("Plug-In Modular Remote Control", September 1980 issue) was very good, but he glossed over a problem that has been a thorn in my side ever since I bought the system. According to Mr. Ciarcia: ". . . since . . . most homes derive
their 117-volt power from both sides of a 220 -volt line, sometimes there can be problems in obtaining consistent operation when receiver modules are used on both sides of the 117 -volt lines."

His solution? "Placement of the receivers could require some experimentation." C'mon, Steve-you can do better than that! There must be some way (or ways) to couple the $120-\mathrm{kHz}$ control signals from one side of the line to the other-capacitively, with a tuned circuit, or even perhaps an active circuit that would function as a repeater. There has to be a way-otherwise, I'm going to be stuck with a houseful of $X$-10 receivers and transmitters that can't talk to each other unless they're in the same room. Please help!
ANDREW BAIRD,
Princeton, NJ
With every article I write, I have to balance safety considerations against the interest that readers with a wide variety of sophistication may have in experimenting with new devices. Yes, there are ways to jump the signal across the two legs of the

AC line. The easiest way is to amplify the command-controller's output and pump a few watts into the line.

Unfortunately, there is no single solution. In some cases, it may only require a few capacitors between the lines. In others, you could use repeaters and still have problems. Also, there are occasions in winter (when 220-volt heaters are in operation) when nothing extra is required for a complete coupling. The answer is not an extra paragraph in a general article on the X-10.

I cannot apologize for evading this apparent limitation of the $\mathrm{X}-10$. To achieve consistent results, any suitable solution requires attaching circuitry directly across the 220-volt power line at its entrypoint to the house. Besides being risky, the power company usually frowns on such activities.

As to the limitations of the $\mathrm{X}-10$, aren't you expecting a lot from a $\$ 15$ remote-control device? Perhaps you should consider using a more conventional remote-control device in conjunction with the BSR system to fill in the gaps. In my own application, I have combined it with hard-wired relaycontrolled outlets for more predictable

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control. If you are interested in learning more about this, I refer you to a three-part article I wrote for the January, February, and March 1980 issues of Byte magazine. The title is: "A Computer Controlled Security System."
STEVE CIARCIA

## UNICORN-1

I am following with interest your series on building your own robot. However your readers may be interested to know that the manipulator arm wiring can be simplified by using two additional diodes. The circuit shown in the figure below (compare with Fig. 18 on page 58 of the September 1980 issue) performs the same functions, but

reduces the amount of wiring required on the arm and saves four positions on the terminal barrier strip. The diodes should be
rated at 50 volts and be able to handle the motor current.

Also, the network shown across the motor can profitably be used across all motors and solenoids to protect the switches from damage caused by arcing when the power is suddenly turned off. Without those, the life of the switches may be considerably shortened. If the current through a motor or solenoid is always in the same direction, a clamping diode, as shown on the "hand" solenoid can also be used.
It's easy to remember which way to connect a clamping diode: simply orient it so it doesn't short out whatever it is connected across!
GUY JUTRAS
Ottawa, Ontario, Canada

## VIDEO SIGNALS

Any government attempt to "stop unauthorized reception" of video signals is logically equivalent to prohibiting you from hearing the governor burp.
Incidentally, Officer Brown (Letters, October 1980), "laws" prohibiting radar detection are equally illegitimate in the U.S.A.
"Piracy?" Commissioner, if you don't wish to water your neighbor's lawn, then direct your hose elsewhere.
JACK D. DENNON
Warrenton, OR

## ON EINSTEINIAN IMPOSSIBILITIES

I am astonished that you would print such a letter as "Einsteinian Impossibilities," by A. H. Klotz even in your "April Fool'" issue.

In fairness to your other readers, you should have pointed out the obvious flaw in Mr. Klotz's argument:

The constant velocity of light is not predicted by Einstein's theory, as Mr. Klotz apparently believes. But the scientificallyproved fact that the velocity of light is constant, regardless of its source, is one of the physical world's anomalies that prompted Einstein to formulate his theory in the first place.

To argue with Einstein's theory, one must ask the question: "How can I account for the fact that the velocity of light from all sources (some of which are moving) is constant?"

If Mr. Klotz has new evidence about the speed of light, I am sure we would all like to hear about his experiments. JERRY MILLER,
Littleton, CO

## WIDE-RANGE AUDIO GENERATOR

In reference to the "Wide-Range Audio Generator" article in your May 1980 issue: It has 2 Zener diodes. Those are not 0.1volt Zeners as stated but rather both numbers are for 5.1-volt Zener diodes (1N5231 and 1N751).
I ordered all required parts, but had trouble finding a 0.1 -volt Zener diode. Fortunately the numbers given were in an ad in the back of the issue, so I checked both of them, and both turned out to be 5.1 -volt Zeners.

It sure looks like a good project and I'm looking forward to completing it. CRAIG LEWIS,
Kailua, HI
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# equipment reports 



CIRCLE 101 ON FREE INFORMATION CARD

THE CHOICE OF A NEW DIGITAL MULTIMETER IS influenced by many factors. Since most units within a certain price range will have almost identical accuracies, other features such as ease of operation, special ranges, and cabinet sizes may be more important. In the past it was common to select a portable unit to take advantage of its use in the field as well as on the bench. In most cases however it has been discovered that a hand-held unit is not the most convenient one for bench use.

Keithley Instruments, Inc. (28775 Aurora Road, Cleveland, OH 44139) has introduced its model 169, a digital multimeter designed for bench use. Since the model 169 uses six C-size batteries for power, it can also be taken into the field if the need arises.

The model 169 features a large ( 0.6 -inch) $31 / 2$-digit LCD display. The easy-to-understand front panel has color-coded pushbuttons. Units are displayed on the LCD to confirm the range and use selected. Since the front panel is larger than that of a hand-held unit, the pushbuttons are easier to operate and the cabinet (again due to the larger size) remains stable. One hand can be used to change ranges. A large bail-type carrying handle attached to the sides of the cabinet also serves as an adjustable stand to allow the user to position the meter at an angle. The cabinet has four feet attached to the underside. Those feet have provisions for test-lead storage. Test leads can be connected to the front panel with standard banana jacks. That means you will still be able to use your favorite special-function leads. The model 169 is supplied with standard test leads, including probes.

Input protection is provided for all functions
on the model 169. Those who forget to change the ranges and functions will be happy to know that the model 169 will withstand 1400 volts (peak) on the voltage ranges and 300 volts on the resistance ranges. A two-ampere fuse protects the current ranges and the fuse is accessible without removing the cabinet.

These specifications will provide the reader with a brief idea of the model 169 's versatility. Both AC and DC voltage is measured on five ranges from 200 mV up to 1,000 volts fullscale. Accuracy on the DC ranges is $0.25 \%$ of reading +1 digit. Accuracy on AC is $0.75 \%$ of reading +5 digits at frequencies under 1 kHz . Input impedance is 10 megohms shunted by less than 100 pF . Current can be measured on five scales from $200 \mu \mathrm{~A}$ to $2,000 \mathrm{~mA}$ full-scale on both AC and DC, although the DC readings are more accurate than AC. Resistance is measured on six ranges from 200 ohms to 20 megohms full-scale. Voltage resolution will vary from $100 \mu \mathrm{~V}$ to 1 V , depending upon the range, in either the AC or DC functions. Likewise current resolution is from .0001 to 1 mA and the resistance ranges offer resolutions of from 0.1 to 10 K ohms.
continued on page 26

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## EQUIPMENT REPORTS

continued from page 24
Overrange is indicated when the three least significant digits are blanked on the display. There is automatic polarity reversal, and a display indicator will signify the polarity of the test voltage or test current. The display will also indicate when there is less than five percent of battery life remaining. The estimated battery life is 1,000 hours when carbon-zinc batteries are used, and 2,000 hours with alkaline batteries.

The instruction manual also serves as a repair manual for those who like to maintain and calibrate their own equipment. It contains exploded views of the unit, complete parts lists, and an extra large two-page schematic diagram. In addition, there is an excellent discus-
sion of the instrument's theory of operation and how the instrument is used.

All circuitry is contained on a $41 / 2 \times 71 / 2$ inch printed-circuit board and, due to the relatively large size, troubleshooting is greatly simplified. The front panel is attached to the board with connectors that makes removal easy. The LCD display is fastened to the panel by two studs and it is connected to the main board by ribbon cable. The model 169 measures $31 / 2 \times 91 / 4 \times 10^{3} / 4$ inches and has a net weight of three pounds.

The unit has been tested and it performed well. All ranges equalled or exceeded the published specifications. If you are in the market for a large DMM that can also be used in the field, it may be worth your time to check out this versatile unit. The model 169 from Keithley sells for $\$ 169$. R-E


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## SPECIFICATIONS

ACCURACY DC Voltage AC VOLTAGE DC CURRENT AC CURRENT RESISTANCE

| RANGE |  | ACCURACY |
| :--- | :---: | :---: |
| $200 \mathrm{mv}, 2 \mathrm{~V}, 20 \mathrm{~V}, 200 \mathrm{~V}, 1000 \mathrm{~V}$ | $.5 \%$ |  |
| $200 \mathrm{mV}, 2 \mathrm{~V}, 20 \mathrm{~V}, 200 \mathrm{~V}, 750 \mathrm{~V}$ |  | $1 \%$ |
| $2 \mathrm{~mA}, 20 \mathrm{~mA}, 200 \mathrm{~mA}, 2000 \mathrm{~mA}, 10 \mathrm{~A}$ | $2 \%$ |  |
| $2 \mathrm{~mA}, 20 \mathrm{~mA}, 200 \mathrm{~mA}, 2000 \mathrm{~mA}, 10 \mathrm{~A}$ | $3 \%$ |  |
| $200 \Omega, 2 \mathrm{k} \Omega, 20 \mathrm{k} \Omega, 200 \mathrm{k} \Omega, 20 \mathrm{~m} \Omega$ | $.5 \%$ |  |

IT SEEMS THAT EVERY TIME WE TURN AROUND, someone has taken another step toward making prototyping less complicated, more instantaneous, and, admittedly, a little more fun. A P Products' Hobby Blox, are a very versatile series of breadboarding products that, working together, perform as a complete system for circuit building.
The generous use of color and the low prices of the several elements of the Hobby-Blox system can easily lead to the impression that it is a system intended strictly for beginners. That is not the case. Hobby-Blox perform on a par with any professional breadboard, in most applications. Its many unique elements permit the easy incorporation of circuit elements that would be difficult to accommodate with many breadboarding products.

Part of the secret of the flexibility of HobbyBlox lies in its unique carrier tray. In addition to its primary purpose of providing structural rigidity for the breadboard strips, the tray also has a number of molded-in features that contribute to the modularity and expandability of the system. The side rails, for example, feature a tongue on one edge and a mating groove on the opposite. There are also slots spaced along one edge to accommodate the blank panel, control panel, or speaker panel elements of the system.

The trays can be readily stacked side-to-side or, with tray extender clips, end-to-end. Also available is a right-angle, vertical tray pack that includes a smaller tray and adapter strip. Either large or small trays, however, may be vertically mounted using the adapter strip.

The standard tray is $6.3 \times 3.12$ inches and the shorter tray is $3.7 \times 3.12$ inches. The inside width, between the side rails, is 2.75 inches. That is the same as each of the solderless strip elements of the system except one-a between-the-trays, 6.3 -inch long, bus strip with two continuous rows of 60 connected, solderless tie points each.

There are five different types of solderless breadboarding elements designed to fit in the Hobby-Blox tray. Yellow terminal strips provide a row of 26 three-tie-point terminals. Red distribution strips each provide two rows of 26 connected tie points. Gray $3 \times 16$-inch terminal strips repeat arrays of 16 three-tie-point terminals (arranged as twin columns of eight each on either side of a 0.3-inch DIP-standard center spacing) three times across its widthperfect for placing three 7 -segment DIP displays in a row. Gold LED strips accept six discrete LED's, provide a solderless tie-point connection to each solderless LED socket lead,
continued on page 32

The Sinclair ZX80. A complete computeronly $\$ 199.95$ plus $\$ 5.00$ shipping.

Now, for just $\$ 199.95$, you can get a complete, powerful, full-function computer, matching or surpassing other personal computers costing several times more.

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## EQUIPMENT REPORTS

continued from page 26
and include a bus of 23 connected tie points for easy connection to power buses through limiting resistors. Blue discrete component strips provide 14 terminals with five connected, solderless tie points each. These strips use 0.2 -inch center spacing (versus DIP-standard 0.1-inch spacing used in the other elements) with adjacent rows staggered 0.1 inches.

Unique and welcome features on those Hob-by-Blox elements include raised molded lines to show the electrical connections between sol-
derless tie points, and a molded letter-andnumber indexing system for each hole.

The yellow terminal strips may be spaced for standard ( 0.3 -inch center) or LSI-wide (0.6inch center) DIP spacing by using a $0.3 \times$ 2.75 -inch spacing strip. Those include the same molded slots as the side rail, and can double as support strips.

The Hobby-Blox system includes a batteryholder pack that houses a standard 9 -volt battery (not included), plus twin three-tie-point terminals for power connections and a bindingpost strip with one black and two red binding posts. Each binding post is brought out to a solderless tie point.

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- Sweep-time magnifier effective for precise measurement; sweep time magnifying 10 times with one-touch operation.
- Trace rotation system for easily adjusting bright-line inclination caused by terrestrial magnetism.
- Z-axis input provided-possible to use as CRT display.
- $0.2 \mu \mathrm{~s} \sim 0.2 \mathrm{~s}$-wide sweep range setting.
- One-touch shifting waveform slopes.
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A P Products has made available two starter packs, one for IC's at $\$ 5.97$ and one for discrete components at $\$ 6.57$. Each starter pack includes a tray, breadboarding elements, and a book of 10 well-documented, easy-to-build projects. Prices for individual elements range from $\$ 1.29$ to $\$ 3.59$.
All-told, Hobby-Blox represents a very interesting and well planned approach to solderless breadboarding. Its features should appeal to beginner and professional alike. A P Products, 9450 Pineneedle Drive, Mentor, Ohio 44060.

R-E


ANYONE WHO HAS SPENT ANY AMOUNT OF TIME with an illegibly-marked group of three-legged semiconductors knows how frustrating and time-consuming a task it is to sort through them. Hickok (10514 Dupont Avenue, Cleveland, OH 44108 ) has automated the task with an amazing new handheld tester, the model 216, that performs in- or out-of-circuit pass/ fail tests. It also identifies NPN, PNP, or FET type; diode and SCR polarity; and the base or gate lead-all in a few seconds.

A device can be plugged into the front-panel transistor socket (graphics clearly identify which of the pins to use to test diodes), or a set of three colored leads may be plugged into the front-panel jacks and alligator-clipped to the device. The comprehensive instruction manual offered by Hickok specifies that for in-circuit testing, the maximum load between any two leads is 500 ohms, $0.2 \mu \mathrm{~F}$, or any impedance equivalent to 500 ohms at 1000 Hz .

All indications are made through six frontpanel LED's. There is a red LED marked FAIL and two green LED's marked PASS, one with the letter P and a diode symbol with the cathode at the left and the other with the letter N and a diode symbol with the cathode at the right. Each of the three terminals to which the device under test is connected has a red LED associated with it. That is true both for the socket on the front panel and the three banana jacks that the color-coded (red-black-yellow) test leads plug into.

A single switch controls operation of the model 216 , selecting the transistor test mode, the diode test mode, or off. After a few seconds (approximately four) of blinking through the test sequence, the LED's hold on a fixed pattern that identifies whether the device is PNP (P LED lights) or NPN ( N LED lights). The LED associated with the lead connected to the continued on page 36

# Light-torque rotary switches make the LM-3.5A DMM as easy to operate as it is to carry. 



On a benchtop or a belt, over a shoulder or in a tool kit, the LM 3.5A DMM, and its LCD counterpart, the LM-350, are ready to go when you are.

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Troubleshooters swear by it. Repairmen find the LM-3.5A works wonders on tvs, business machines, even cameras. Checks all quiescent AC and DC voltage values. Spots current drains. Measures the resistance of suspect components. Quickly and precisely.
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LM-3.5A at a glance.

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| AC Volts | 1 to 750,4 ranges |
| Kilohms | 1 to 10000,5 ranges |
| AC/DC Current | 1 mA to $1 \mathrm{~A}, 4$ ranges |
| Polarity Selection | Automatic |
| Readout | $0.3^{\prime \prime}$ Red LED |
| Size | $1.9^{\prime \prime} \mathrm{H} \times 2.7^{\prime \prime} \mathrm{W} \times 4.0^{\prime \prime} \mathrm{D}$ |
| Weight | 9.2 oz (batteries installed) |
| Power | 3 type AA rechargeable |
|  | Nicad batteries and charger |
| Price | $\$ 165.85$ |

LT-3 Digital Temp Meter. Featuring $0.1^{\circ}$ resolution and high accuracy, the $31 / 2$-digit, 2,000 count full scale LT-3 is indispensable for home or industry. Checks everything from thermostats to appliances. Even monitors critical operations like photoprocessing and electroplating.
The LT-3 can be supplied with any of eight thermistor and RID temp sensors to read ranges of $0-100^{\circ} \mathrm{C}$ $32-199.9^{\circ} \mathrm{F}$, or $0-199^{\circ} \mathrm{C}$ or F .
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Operator convenience is the key to our line of frequency and temperature meters, too. Pictured leff to right, SC-5 prescaler, FM-7 frequency meter, LED format LT-3 digital temp meter, and its LCD cousin, the LT-31. Top, the MLB-1 digital logic probe.

Hobbyists, radio and tv studios, phone companies and the military all depend on the versatile FM-7. Whether the job calls for calibrating fixed, variable frequency or voltage-controlled oscillators, checking flowmeters, highspeed photocell counters, or setting the IF or heterodyne frequency in communications equipment, the FM-7 is a standout performer

## SC-5 Prescaler. Top range booster.

 This $512-\mathrm{MHz}$, battery or AC line-operated prescaler was developed to extend the frequency range of the FM-7 from 60 to 512 MHz . Adapts to most other frequency meters, too.Get the word on us. We offer a full lineup of convenient, competitively priced products. From DMMs, frequency and temp meters to miniscopes and DPMs.

For further technical information or the names of your nearest distributors, contact Non-Linear Systems Inc., 533 Stevens Ave., Solana Beach, CA 92705. Telephone (714) 755-1134. TWX 910-322-1132.


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THE TEST EQUIPMENT SPECIALISTS TOLL FREE HOT LINE

## EQUIPMENT REPORTS

 continued from page 32base of the transistor also lights. If the device is bad, the FAIL LED flashes alternately with the others. If the device is an FET, the N indicates N -channel, the P indicates P-channel, and the lead LED signifies the gate. The model 216 will not give a correct indication on some low $\mathrm{g}_{\mathrm{m}}$ FET's because of their low on-resistance, though it will indicate the validity of any P-N junction in the diode test mode. Also, the model 216 cannot test where there is no junction, as in the gate lead of an IGFET.

In the diode test mode, only the two lead connections marked with a circled D are used, and a good device is indicated by the lighting of the appropriate PASS LED's, indicating the
polarity of the diode or junction. The instruction manual explains how to use that mode to test SCRs, triacs, bridges, and other devices. The model 216 will not test PUT's or tempera-ture-compensated Zeners.

The circuitry of the Hickok model 216 is a fascinating hybrid of analog and digital approaches to semiconductor testing. A 4022 octal counter/driver scans the tester through a sequence of tests, controlling a total of 12 analog switches. A bipolar oscillator's output is applied to two leads of the device under test, and its third lead is connected to the input of an op-amp configured as an inverting lowpass (DC-smoothing) filter. All three permutations of lead connections are tried, in sequence.

It can be shown that for the two-diode model of a semiconductor, a device with an open or shorted junction will always provide a zero out-

put from the op-amp for at least one of the three configurations of leads, but a good device never will.

The output of the first scan goes from the op-amp filter to an op-amp integrator, and the result of the three lead configurations (two always result in one polarity of voltage out of the filter, the third in the opposite polarity) is stored while the same three configurations are repeated.

Thanks to some very clever use of logic gates, a comparison between the integratorstored majority-vote voltage polarity and the second-scan output of the op-amp filter triggers three events: first, a four-second delay is introduced, inhibiting the next count of the 4022 and permitting adequate display time for the results of the tester's analysis; second, another op-amp circuit lights either the P or N LED depending upon the polarity of its input (intriguingly, that same op-amp resets the 4022 to light the FAIL LED in the case of a zero-voltage out of the filter); third, as that sequence of events has uniquely indentified the base lead, the approximate LED is lit.

The model 216 does not (nor is it intended to) provide any information on the gain or any other device characteristic. Those seeking quantitative tests must look elsewhere. But as a qualitative tester, it is one of the simplest, most foolproof, thorough and informative instruments we've encountered.

Power for the model 216 comes from a pair of 9 -volt batteries and a rechargeable battery option is available. All inputs are protected with back-to-back Zener diodes in case of residual voltages during in-circuit (powerdown) tests. Hickok offers a one-year limited warranty. The 18 -page instruction manual includes a complete parts list, a schematic, and a printed circuit diagram.

Manufacturer's suggested U. S. retail price for the model 216 is $\$ 135$. It's available through electronic distributors.

## Gold Line Model 1139 Base Station Power Supply

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An internal trimpot adjusts the output voltage from 13.5-15 volts. Voltage regulation is provided by a type-723 IC, and the pass transistor is a 2 N3055. A full-wave bridge consisting of four 1 N 5400 silicon diodes provides the DC rectification.

The husky transformer is capable of running at the full current limit without becoming hot. The current is limited by the power transistor which does become hot, even though adequately heat-sinked. A thermostatic switch mounted on the heat sink senses excessive heat dissipation and shuts down the power supply for several minutes when overloaded.

All circuit components (except the transformer and output transistor) are mounted on a phenolic circuit board. An internal fuse provides additional circuit protection. That fuse is soldered in place, but a replacement is rarely necessary.

Our test unit worked well, after shipping damage was repaired. The filter capacitor leads had broken loose during shipment and the power transformer was bent slightly on its frame (upright type " A " mount). Fortunately, component spacing proved to be adequate, and no secondary damage occurred from loose or bent components.

The model 1139 power supply cabinet is finished in a black-wrinkle enamel. Ventilation holes provide adequate cooling if the unit is operated within its specified limits. Rubber foot pads prevent the unit from scratching the surface of a desk or operating table. A frontpanel incandescent pilot lamp lights when the unit is on. It is connected across the DC output of the power supply, providing a bleed load when the unit is off. A pair of (red and black) press-release terminals on the rear apron of the supply securely grips power leads.

The supply weighs about $4^{1 / 2}$ pounds, and measures $41 / 2 \times 3 \times 6$ inches. Our unit came without any literature or instructions.

Because of the 3-amp nominal current rating of the 13.5 -volt supply, it proves to be particularly well suited for home operation of most mobile electronic equipment. The model 1139 3-amp base station power supply from Gold Line sells for $\$ 27.98$.

R-E

## Sony ICF-6700W General Coverage Receiver



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FOR QUITE SOME TIME, MANUFACTURERS HAVE made either very good or very bad shortwave receivers. There seemed to be nothing in the middle. But now the market is changing. Manufacturers seem to be recognizing the fact that there are a number of serious hobbyists who enjoy casual monitoring of the shortwave spectrum and expect dependable equipment with which to do it.

Sony has continued the trend with a series of

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## EQUIPMENT REPORTS continued from page 37

quality, hobby-class radio receivers. One of the better ones is the model ICF-6700W.

The military-styled cabinet measures $18 \times$ $71 / 4 \times 9$ inches. Selectable upper or lower sideband, AM, and CW modes are available from $.53-29.7 \mathrm{MHz}$ (the receiver skips $10.4-11.3$ MHz ) with FM coverage from 87.5-108 MHz .

A narrow/wide IF selectivity switch really helps separate closely-spaced stations. Additional audio bandpass shaping is possible with the separate bass and treble controls. The RF gain control may be adjusted to reduce strong signal overload on SSB.

The FM fidelity is very good, with the AM
sound quality superior to that found on most small home-entertainment radios. A multiplex output jack allows coupling of the FM IF signal to a stereo adaptor. A timer jack provides for external on/off control of the receiver by a suitable timing clock (unfortunately, neither the stereo adaptor or timing clock are available from Sony.) Separate recorder and headphone output jacks are also included. Audio output is 900 milliwatts (at $10 \%$ harmonic distortion).

A switchable AFC control allows FM locking within approximately $\pm 100 \mathrm{kHz}$ of the center frequency. An illuminated signalstrength meter doubles as a battery tester. The digital, five-character, LED frequency display works on all frequency ranges, and may be switched off to conserve battery power during portable operation of the receiver. It is quite accurate and our sample was within 1 kHz at


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Book 3: Half adders and full adders; subtractors: serial and parallel adders: processors and arithmetic logic untems. Book 4: Flip-flops: shift registers: asynchronous counters; ring, Johnson and exclusive-OR feedback counter, random access memorie
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ROM: address decoding: instruction sets: instruction decoding: control program instruction
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BASIC COURSE


Digital Computer Logic \& Electronics

## CONTENTS

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room temperature on all modes. The display is a little dim in bright ambient light, but that is probably intentional for lower current drain during battery operation.
The gating time of the display is a little slow, so it takes a fraction of a second for the display to catch up during tuning. That has a tendency to slow down the dial-searching process, because the receiver doesn't have a calibrated shortwave dial and the user is entirely dependent upon the digital readout.
Frequency stability of the receiver is excellent. Single sideband reception is quite acceptable, although dial setting is touchy. There is no fine tuning, so the initial setting process is rather delicate.
Minor pulling of the oscillator frequency occurs as a result of the normal AGC action when receiving strong CW and SSB signals. That effect is eliminated by reducing the RF gain. There is some hand-capacitance effect apparent on the front panel, but it is minor. The mechanical stability of the receiver is fair and tapping or pressing the cabinet will result in some frequency instability. No dial backlash was detectable.
The main tuning dial is a flywheel. It has a good rugged feel to it, and seems securely mounted. A linear-preselector dial is used for peaking the RF input to the receiver. It is poorly calibrated, so the user must depend upon an increase in background signal to know whether or not he has optimized receiver sensitivity. That may be a problem; if the preselector dial is peaked on the wrong frequency, image interference will be enhanced, and desired frequencies will be attenuated. Proper preselector setting may take a little getting used to. The receiver does not have a noise limiter and that may pose problems.
Antenna provisions include a built-in ferrite bar for AM broadcast (or push terminals for an external wire antenna), and a collapsible whip for FM and shortwave (or push terminals for an external shortwave antenna). There is no external FM antenna input.
One desirable feature is the automatic fre-quency-offset readout during upper and lower sideband monitoring. When the station is properly tuned in, the suppressed carrier frequency will be displayed. The offset is $\pm 2$ kHz , ideal for SSB reception.
The battery compartment is accessible from the top of the cabinet (six D-type batteries are used). With the lid flipped up, a world timezone chart and table of international broadcasting frequency allocations is displayed.
Carrying handles are mounted on the sides of the receiver, and slots are provided to accommodate shoulder straps. Adjustable plastic feet allow the receiver to be tilted for comfortable viewing and operation.
The model ICF-6700W may be operated from its own internal batteries, from an external 12-volt source (using a Sony battery cord to drop the voltage to 9 volts), or from AC lines ( $110,120,220$, or 240 volts AC, switchselectable). Power consumption ( AC ) is 7 watts.
The accompanying service manual is rather brief. It does provide basic user tips, but has very little theory; no schematic or block diagram of the receiver is included. No warranty policy was supplied with the unit that we evaluated, but there was a list of regional service centers.
The model ICF-6700W lists for $\$ 439.95$ and it is available from Sony dealers. From Sony Division, Sony Industries, 9 W. 57th Street, New York, NY 10019.

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# HI-FI NOISE FLLTER/Range Expander 

> Bring out the best in any recording with this combination noise filter/dynamic range expander.

THE ARTICLE, "NOISE REDUCTION TECHniques," that appeared in the January and February 1981 issues of RadioElectronics, presented block diagrams of commercially available dynamic range expanders and noise filters. That two-part article showed how, by improving the dynamic range of even the best recorded musical signals, expanders and noise filters restore much of the emotional impact that is lost during the recording process. This two-part article will describe the operation and construction of a combination dynamic range expander/noise filter called the ASRU (Audio Signal Restoration Unit).
This month, we will describe the basic operation of the ASRU and provide an in-depth description of how the expander portion of the circuitry works. Next month, we will discuss how the noise-filter circuitry of the ASRU works and provide the construction, installation, and operation details.

## The expander-how it works

Like the expanders discussed in the January 1981 issue of Radio-Electronics, the expander section of the ASRU makes the low-level signals softer and the loud signals louder, thus providing improved realism and reduced noise. The expansion curve of the circuit is shown in Fig. 1. Note that the total change in gain is about 8.5 dB ; the slope is very shallow. It requires over 40 dB of range to change from minimum to maximum gain, for an average expansion rate of about 1.2:1 (the ratio of output-level change to input-level change in dB ). The curve shown provides expansion without unnatural side-effects.


FIG. 1-EXPANSION CURVE of the ASRU shows a shallow slope.

A block diagram of the expander portion of the ASRU is shown in Fig. 2.

The first stage sums the right- and left-channel signals coming from the noise filter so that both channels can be controlled together, preventing the stereo image from changing due to variations in signal level in one channel or the other.

The control-voltage filter takes the output of the summing network and attentuates the high- and low-end frequencies to produce an audio signal that approximates the response of the
human ear (see Fig. 3). That response is shown by the well-known FletcherMunson curves (Fig. 4) that depict the sensitivity of the ear for equal perceived loudness at different frequencies. Note that, at most levels, the ear is significantly more sensitive to midrange frequencies than to high- or low-end ones. In fact, due to the resonance of the ear canal, the ear is most sensitive to sounds in the $4-\mathrm{kHz}$ range.

That midrange sensitivity accounts very strongly for our perception of the loudness of a sound and the controlvoltage filter is designed to take advantage of that fact.

The attenuation of both ends of the audio spectrum tends to reduce the effects of noises such as turntable rumble and FM multiplex "hiss."

Furthermore, the steep roll-off at low frequencies prevents low-frequency signals from causing rapid and unnaturalsounding gain changes. That is beneficial because sudden changes in gain during the period of a signal can result in harmonic distortion-something we


FIG. 2-BLOCK DIAGRAM of the expander portion of the ASRU. The first stage sums both channels to maintain stereo imaging.

## R-E TESTS IT

## LEN FELDMAN

## CONTRIBUTING HI-FI EDITOR

We tested a prototype of the Audio Signal Restoration Unit in our laboratory, using static signals as well as musical program material. As the author suggests, setting up the unit is a bit tricky. To some degree (unless the expander section is turned off altogether), there is some audible interaction between the various front-panel controls on the unit. We found that the best setting for the sensitivity control is such that medium or average loudness-level portions of the program source cause sequential extinguishing of the indicator LED's. The threshold control should be set so that in the absence of any signal, the lowest-level LED flashes only occasionally.
With the expander switch to the ON position, optimum setting of the expander-sensitivity control occurs when the right-hand LED flashes only intermittently. Of course, it is possible to use each section (noise reduction, dynamic filter and expander) as required, to suit program material, but we found that with the controls set as described above, we were able to improve reproduction of most program sources without having to make extensive readjustments every time we altered program material or content.
With the expander out of the circuit, and with the unit set for widest bandwidth (no dynamic filtering or noise reduction), overall frequency response of the unit measured flat within $\pm 0.75 \mathrm{~dB}$ from 20 Hz to 20 kHz . The unit has essentially unity gain, but that may be varied by means of the input sensitivity control. With 0.5 volt input, we measured a signal-tonoise ratio of 90 dB , IHF " $A$ "-weighted. With both the expander and the noise filter on, total harmonic distortion for a $1-\mathrm{kHz}$ input signal at the 0.5 -volt level measured $0.17 \%$. With the expander turned fully off (the threshold control at its minimum position) but the noise filter on, distortion decreased to less than $0.1 \%$ for the same test signal.

A series of composite spectrum analyzer sweep photos for the ex-pander/filter/noise-reduction unit is shown in Fig. 1. In both the upper and lower series of sweeps, the expander is on and degree of expansion is varied, as are the noise reduction and filtering action. Note that greater expansion occurs at the higher signallevel (upper traces) and that regardless of the level at which the tests were made, no expansion is evident at the low-bass frequencies.
Figure 2 shows the expander action alone (without any noise reduction or band-filtering action). With the expander turned off, response is flat from 20 Hz to 20 kHz ; but with the


COMPOSITE spectrum-analyzer sweep photos for the ASRU.

expander turned on, the degree of expansion for louder passages, less for moderate passages and, in the lower traces, even a bit of downward expansion for quietest passages.
The Audio Signal Restoration Unit operates with very few side effects once it is properly adjusted. By not allowing expansion to take place at the bass frequencies, the designer has overcome some of the pumping and breathing effects common to other linear expanders. The 1.2:1 ratio of expansion is quite moderate, compared with some other commercially available expanders, but nevertheless is sufficient to add a measure of realism to most program material that has been compressed during recording.
As for the variable-bandwidth filters: if used to excess, they can create some undesirable audible effects; but it is possible to benefit from them without suffering such effects if adjustment of threshold and bandwidth is carefully done while listening to program material. We did not find the indicator LED's to be as helpful in setting up the unit as the author had suggested; but we did find that, with a little practice, we were able to use the "ASRU" with just about any component system that is equipped with an ordinary tape-out/ tape-play monitor loop. The tapemonitor loop on the amplifier that is used to conect this unit is duplicated on the unit itself, so owners of cassette or open-reel tape decks need not worry about losing it.


FIG. 3-FREQUENCY RESPONSE of controlvoltage filter matches that of ear.


FIG. 4-FLETCHER-MUNSON curves show that the ear is most sensitive to midrange frequencies.
can do without.
The audio from the filter is passed through a precision full-wave rectifier that generates a current used to produce the control signal.

The logarithmic curve-shaping and attack/delay circuits convert that current into a control voltage that is approximately proportional to the logarithm of the current and that section of the expander provides attack and decay times that adjust themselves to the rate of change in signal strength.

Finally, the control voltage is supplied to the voltage-controlled attenuator/amplifier where it is used to modify the qualities of the original audio signal.

The ASRU's expander does not expand signals in the low-bass region as much as it does in others. There are two reasons for that.

First, consider Fig. 5-a, showing a warp or rumble (very-low-frequency) waveform along with a toneburst. As can be seen in Fig. 5-b, at the moment the toneburst is added, the level of the warp signal will increase because the expander will increase the gain and a "thump" will be evident, even though the warp noise alone was inaudible. Figure 5-c shows what happens when the ASRU is used-the "thump" doesn't occur because the action at very low frequencies is minimal.

Second, although the ear is relatively insensitive to very low frequenciesrefer to the Fletcher-Munson curves in


FIG. 5-WARP OR RUMBLE (thin line) with tone burst (thick line). The ASRU (c) eliminates thumps by not expanding such a signal.


FIG. 6-THE ASRU's gain vs. frequency response curves at varying control-voltages. Note that the low-bass region is not expanded.


FIG. 7-THE TRIPLE-OUTPUT voltage-controlled filter has two integrators for each channel.

Fig. 4 -once their level is above the threshold of hearing, a $2-\mathrm{dB}$ increase appears as great as a $5-\mathrm{dB}$ increase in the midrange area.

For both of those reasons, as well as to keep distortion to a minimum, the ASRU's expander does not expand the low-bass as much as it does the midrange. Figure 6 shows the ASRU's gain
vs. frequency response at varying con-trol-voltage levels. Note how well that matches the changes in gain sensitivity shown in Fig. 4.

The ASRU's shallow expansionslope, midrange-emphasized controlsignal and minimized low-bass expansion explain why it is so cleansounding, while allowing 8.5 dB of
effective expansion.

## The noise filter-how it works

The heart of the noise-reduction system is a triple-output, voltage-controlled filter, the block diagram of which is shown in Fig. 7. It is a statevariable filter, which means that certain of its characteristics can be modified while others are maintained.

The integrators process the signal for use by later stages of the noisereduction system (see Fig. 8). For sinewaves, the output is reduced by a factor of two ( 6 dB ) for every octave increase in input frequency. By varying the gain or time-constant of those integrators, or the amount of feedback around them, the corner frequency (the frequency at which the amount of attenuation reaches 3 dB ) can be changed without changing the shape of the filter.

Refer to Fig. 8 as we discuss the ASRU noise-reduction system.

If no signal is present, the control voltage sets the corner frequency of the triple-output filter to 1.2 kHz . Figure 9 shows the frequency response of each output of the triple-output filter with the corner frequency set at 1.2 kHz . The overall output of the noise filter is taken from the low-pass output via a buffer. Thus, with no input signal present, any noise will be greatly attenuated.

If a $5-\mathrm{kHz}$ tone is suddenly applied to the input, it will appear unattenuated at the high-pass output and will be greatly attenuated at the low-pass and bandpass outputs. The AC-DC converter connected to the high-pass output will provide a strong signal that will rapidly pass through the attack/delay element and cause the control voltage to increase. As the control voltage increases, the corner frequency of the filter will also increase until it exceeds 5 kHz .

Soon there will be a stronger signal in the bandpass section than in the high-pass section. That is converted to DC and will be fed back and reduce the control voltage. In the case of a steady tone, that action will serve as a feedback loop that forces the bandwidth of the filter to "catch" the input frequency, allowing it to go through the low-pass filter to the output, while the noise above that frequency is filtered out.

Music, of course, is more than just simple tones. The ASRU noise filter will track the highest significant frequency of a complex signal. During a transient-a short, but intense, increase in high-frequency energy-the corner frequency will overshoot slightly. That is desirable, since transients mask noise very well.

If the signal is extremely strong,

## PARTS LIST

All resistors $1 / 4$ watt, $5 \%$ unless otherwise specified
R1, R9, R10, R101, R109, R110, R221, R521, $-100,000$ ohms
R2-R5, R102-R105, R211, R213, R214, R305, R306, R309, R310, R405, R406, R409, R410, R519, R533, R535- 10,000 ohms
R6, R106, R230, R507, R508, R520, R530-20,000 ohms
R7, R12, R107, R112-200 ohms
R8, R14, R15, R108, R114, R115, R205
R501, R502, R529-2200 ohms
R11, R111, R218- 36,000 ohms
R13, R113, R201, R208, R209, R212, R215, R220, R222, R229, R513, R522, R528-4700 ohms
R16, R116, R202, R203, R210, R50310,000 ohms, $30 \%$, slide potentiometer, linear taper
R17, R117, R506, R537-3900 ohms
R204- 2700 ohms
R206, R207- 1.5 megohms
R216- 68,000 ohms
R217, R509-3300 ohms
R219-820 ohms
R224, R307, R308, R312, R407, R408, R412, R514, R531, R532, R5341000 ohms
R226, R516-270 ohms
R228- 150 ohms
R231, R302, R303, R402, R403270,000 ohms
R232, R304, R404-1200 ohms
R223, R510-6800 ohms

R225, R233, R313, R413, R504560 ohms
R227, R527-120 ohms
R301, R401, R526, R536-1500 ohms
R311, R411, R525-12,000 ohms
R505-470,000 ohms
R511-910,000 ohms
R512, R515, R517-22,000 ohms
R518, R523- 47,000 ohms
R524-47 ohms
R601- 1.5 ohms

## Capacitors

C1, C3, C5, C101, C103, C105, C201,
C202, C204, C506- $0.01 \mu \mathrm{~F}, 5 \%$
Mylar
C2, C6, C7, C102, C106, C107, C503,
C504- $3.3 \mu \mathrm{~F}, 35$ volts, electrolytic
C4, C104, C205, C212-C214-0.022 $\mu \mathrm{F}$, $10 \%$, Mylar
C203- $0.001 \mu \mathrm{~F}, 10 \%$, Mylar
C206, C208- $0.0033 \mu \mathrm{~F}, 10 \%$, Mylar
C207, C209-680 pF ceramic disc
$\mathrm{C} 210, \mathrm{C} 211, \mathrm{C} 301, \mathrm{C} 401-10 \mu \mathrm{~F}$,
25 volts, electrolytic
C215, C505- $1 \mu \mathrm{~F}, 35$ volts, electrolytic
C216, C302, C307, C402, C407, C502$0.1 \mu \mathrm{~F}, 5 \%$, Mylar
C303, C304, C403, C404, C501100 pF , ceramic disc
C305, C306, C405, C406-22 $\mu \mathrm{F}$, 16 volts, electroytic
C601, C602- $1000 \mu \mathrm{~F}, 25$ volts, electrolytic
C603-C607-0.1 $\mu \mathrm{F}$, ceramic disc


FIG. 8-BLOCK DIAGRAM of the ASRU's noise reduction system. If no signal is present, the corner frequency of the filter is set to $1.2 \mathbf{k H z}$.


FIG. 9-THE TRIPLE-OUTPUT filter has low-pass, high-pass, and bandpass outputs. The scale here is $\log -\log$.

## Semiconductors

D201-D204, D206-D210, D501-D513,
D515-D517-1N4148
D205-3.3-volt Zener
D514-4.7-volt Zener
D601-D604-1N4001
LED201-LED204, LED501, LED502-mini-LED (TL209 or equivalent)
Q201-Q203, Q501, Q502-2N3904
Q204-Q206-2N4250
IC1, IC2, IC4, IC7, IC9, IC10, IC11RC4136 quad op-amp
IC3, IC6-4049 CMOS hex inverter
IC5, IC8-739 dual audio preamplifier
IC12, IC13-78L12A 12-volt positive voltage regulator
L201, L202- 6.8 mH coil
T1-13.5 VAC, 350 mA , wall-plug transformer (Dormeyer PS14204 or equivalent)
J1-J4, J101-J104-RCA-type phono jacks
S1, S2-DPDT toggle or slide switch
Miscellaneous: 12 -conductor ribbon cable, IC sockets, chassis and end panels, solder, wire, hardware, etc.

The following are available from Symmetric Sound Systems, 912 Knobcone Place, Loveland, CO 80537: Complete kit (ASRU) \$110.00; PC boards (ASRUPC), $\$ 18.00$. Write for information on assembled units. No other parts or different combinations are available. End panels are unfinished. All prices include UPS shipping within U.S. Colorado residents add $3 \%$ tax.

## EXPANDER-ONLY KIT

For those requiring only the expander portion of the ASRU, a kit, somewhat different from the one described here, is available from Symmetric Sound Systems. That kit, the EX-1, is priced at $\$ 60.00$. A bare PC board, the SSS7, is also available for $\$ 11.00$. See parts list for ordering information. A schematic, parts list, and a diagram for laying out your own EX-1 PC board can be obtained from the above company if a self-addressed, legalsize, stamped envelope ( 28 cents) is sent along with the request.
the clamp in the bandpass section will allow the bandwidth to extend all the way to 25 kHz . The attack/decay circuitry is designed so the bandwidth of the filter can be expanded rapidly, but takes longer to decrease than it did to increase. Because of the large amount of feedback used to control the bandwidth, that nonlinear response does not affect the steady-state (constantlevel) response, but becomes very important in the case of transients.

As pointed out in the "Noise Reduction Techniques" article in the February 1981 Radio-Electronics, one of the advantages of a filter/expander combination is that each section can be adjusted to keep side-effects to a continued on page 110

NEW DEVELOPMENTS IN



#### Abstract

With us since radio's early days, jamming stands as the biggest obstacle in the way of fair use of the shortwave broadcast-bands. Here's a look at the current situation.


## STANLEY LEINWOLL

WHAT WILL PROBABLY TURN OUT TO BE the most important shortwave broadcasting conference ever held has been scheduled by the Administrative Council of the International Telecommunication Union.

Acting on the recommendation of WARC-79, (World Administrative Radio Conference-1979), the Council has set January 1983 for a shortwave broadcasting WARC that is to establish technical standards and procedures related to planning the use of the shortwave broadcast spectrum. A second session of the BC-WARC is scheduled for October 1984. At that session an attempt will be made to plan the efficient and equitable use of the bands allocated to shortwave broadcasting.

The Conference has its work cut out for it, with a number of major obstacles to overcome before any serious effort at planning can succeed. One of the problems overhanging the Conference like a dark and ominous cloud is jamming. Indeed, many of the world's most competent shortwave broadcasting experts feel that rational planning of high-frequency broadcasting is impossible as long as jamming continues.

To understand fully why jamming
and technically feasible high-frequency broadcast planning are judged by many observers to be incompatible, it is necessary to provide some information about jamming, and to describe attempts by jammed broadcasters to overcome it.
Jamming is the deliberate transmission of raucous, irritating noise and other interference on a frequency in order to hamper or utterly destroy the programs of another broadcaster operating on the same frequency.
Jamming doesn't only interfere with the target broadcast; it also degrades the transmissions of broadcasters operating on adjacent frequencies because of its broadband characteristics. Consequently, for each frequency jammed, three are adversely affected, as a rule: the one being jammed, plus the frequency on either side of that one.

During the height of the Cold War, virtually every major Western broadcaster transmitting to the Communist world was jammed. In recent years, however, there has been a decrease in jamming; but it is still a serious blight on the shortwave spectrum. At the present time the USSR and some of its satellite countries are responsible for
most of the jamming being observed. Soviet jamming is currently being directed principally toward the broadcasts of Radio Free Europe, Radio Liberty, The Voice of Israel, and the People's Republic of China. In addition, the People's Republic of China jams some Soviet programs beamed to China.

To accomplish their task, the USSR and its satellites have developed a highly complicated and very sophisticated jamming network, consisting of several thousand jammers at hundreds of different locations throughout eastern Europe. It is estimated that it takes about five thousand technicians and administrators to operate the jamming system at a cost far exceeding that of the broadcasts being jammed. Furthermore, the original cost of setting up such a jamming system probably exceeded a quarter of a billion dollars.

There are two types of jammers: local, and sky-wave. Local jammers operate primarily in and around large population centers, usually those with a population of a quarter of a million or more persons. They are generally located so that they overlook the region to be jammed. Local jamming, which is often incorrectly referred to as


FIG. 1-LOCAL JAMMING consists primarily of a direct wave and a reflected wave. The ground wave is fairly unimportant in this type of jamming.
tried to overcome its effects in a number of ways. Some of those techniques have failed while others have been highly successful. In order to assess fully the impact of jamming on planning efforts at WARC-83/84, the major anti-jamming techniques will be discussed.

Those anti-jamming techniques include using high-power transmitters of up to 1000 kW , and high-gain, highly directional curtain antennas, by means of which effective radiated powers of more than 100 megawatts can be achieved. That brute force technique produces very high signal-levels, on the order of five to ten millivolts-per-meter, delivered to the target areas. Signals of that order of magnitude put considerable strain on the jammers and increase areas in which desired signal strengths are above the jammer noise. That enables listeners to receive the programs.

One of the best methods of countering the effects of jamming is generally known as saturation- or barragebroadcasting, in which as many transmitters as possible-each on a different frequency-are massed simultaneously to carry a particular program. Satura-
ground-wave jamming, consists principally of a direct wave and a reflected wave, as shown in Fig. 1. It is evident, from Fig. 1, that ground-wave jamming plays a relatively minor role in the effectiveness of that type of jammer.

The effective range of local jammers depends on the height of the jamming antenna. Although the average range is about 20 to 30 kilometers from the antenna tower, it is obvious that the higher the antenna, the greater the distance the jamming signal will travel. Most local jamming antennas are placed on tall buildings, church steeples, or on hills or mountains overlooking the target.

Depending on the size of the area they must cover, and the number of people in it, local jamming stations have between fifteen and fifty jamming transmitters. Those are of relatively low power, which ranges from 5 kW to 20 kW each.

Sky-wave jammers are used to propagate the jamming signals great distances via the ionosphere. They can cover considerably larger areas than the local jammers, and their principal mission is to blanket areas lying in the rural and suburban parts of a target area that lie beyond the range of the local jammers. That is shown in Fig. 2.

Sky-wave jamming transmitters are of much higher power than local jammers, being of the order of 50 kW to 100 kW each. The antenna systems used are more sophisticated (rhombics and log-periodic antennas, compared with low-gain broadband dipoles employed with local jamming transmitters.)

Jamming transmitters are modulated in two ways: white noise is produced electronically and covers most of the


FIG. 2-SKY-WAVE JAMMING uses the ionosphere to propagate jamming signals over great distances.
audio spectrum. It is referred to as noise jamming and is very effective.
In Mayak jamming, distorted program material is transmitted in lieu of white noise. The word mayak means "beacon," which is the name of one of the Soviet domestic home-service programs. It is not unusual to find three or four simultaneous Mayak transmissions, each slightly out of phase with the orders, each distorted, operating on one frequency. They, too, are highly effective.

Over the years, broadcasters whose transmissions were being jammed have
tion-programming has been quite successful in putting pressure on the jamming system to the point where some of the frequencies being used to transmit a program are either thinly covered by the jammers, or not covered at all.

It is obvious that even in population centers with their own local jamming networks, if more than fifteen transmitters are used to carry a program and the jammer complex has fewer than fifteen jammers, some frequencies will be clear of local jamming.

In the past, when broadcasts of BBC, VOA, Deutsche Welle, etc.


FIG. 3-DURING THE TWILIGHT PERIOD only local jamming is effective because key-wave jamming signals are not reflected by the lonosphere.
were jammed, efforts were made to coordinate programming among them so that the maximum number of frequencies were being used to carry jammed transmissions. That method, when used, was highly successful.

Perhaps the most effective method of all for overcoming jamming is the
use of a basic shortwave radio propagation technique usually referred to as twilight immunity. During the daylight hours the ionosphere is able to propagate higher frequencies than at night. That is so because radiation from the sun produces ions and free electrons in the ionosphere. The range
of frequencies the ionosphere can reflect is proportional to the number of those particles. At night, radiation from the sun is cut off and free electrons and ions begin to re-combine, resulting in a less dense ionosphere, which is capable of supporting only the lower frequencies.
It follows, therefore, that a transmitter to the west of a target area will enjoy a period of several hours in the late afternoon (twilight) when the path between the transmitter and the target is in daylight, but the target area itself is in darkness. That is shown in Fig. 3. During that twilight period as many high-frequency transmissions as possible are scheduled. Examination of Fig. 3 shows that jammers operating via the sky-wave mode are relatively ineffective because attempts to use the higher frequencies are generally fruitless, since those frequencies are not reflected by the ionosphere. During twilight immunity-periods only local jammers are effective. Since there are literally thousands of cities and towns that do not have local jammers, a high degree of effectiveness can be achieved by using the saturation technique.

It is clear at this point that attempts by WARC-83/84 to develop technical standards or to plan the rational use of the spectrum will be severely hampered by continued jamming.

Preliminary planning for the broadcasting WARC has included discus-

## A HISTORY OF JAMMING

The Germans are generally credited by radio historians as being the first to use jamming techniques. As early as 1915, they transmitted random characters to disrupt a radioteletype circuit between France and Russia, which were allies during World War I.
In the 1920's, before radio broadcasting was regulated, some broadcasters deliberately transmitted on frequencies being used by competing broadcasters in an effort to drown out their programs. Although some of the interference caused during those early days was accidental, there is no doubt that much of it was intentional.

The first case of political jamming occurred in the mid-1930's before the German-Austrian anschluss. The government of Chancellor Dolfuss of Austria jammed some Nazi-German broadcasts to Austria that were critical of that country. The Nazis themselves were quick to recognize the effectiveness of jamming to keep out unfavorable comment, and the Spanish, French, Russians, Japanese, and Italians soon followed suit.

Jamming increased in intensity and effectiveness during World War II, being used as a military weapon, both to keep out unwanted broadcasts and to disrupt military circuits. The Germans jammed
broadcasts of the BBC extensively. and the escape to the open seas of the German warships Scharnhorst, Gneisenau, and Prinz Eugen, under the nose of British artillery. was possible because the Germans effectively jammed British radar installations overlooking the English Channel.

With the advent of the Cold War after the conclusion of World War II in 1945, jamming flourished to an extent that had not been dreamed of before. The Russians jammed programs in the Russian language that were beamed to the USSR from Franco Spain, and the Spanish, in turn, jammed Spanish-language broadcasts emanating from the Soviet Union.
In early 1948, however, a jamming effort was begun by the Russians that dwarfed anything that had been attempted before. In February of that year a dozen or so Russian transmitters were used to jam the Russian-language programs of the Voice of America. The Russian-language broadcasts of the BBC were jammed shortly thereafter, and by 1950, over 450 such jamming transmitters were in operation.

Although we have no way of knowing for certain, it appears that the decision to launch a massive jamming campaign was twopronged: first, it was a method of
keeping control of the information monopoly within the USSR, where total censorship of news from external sources was the policy. Second, jamming could be used milltarily, as had been demonstrated in World War II, and by operating large numbers of jammers the military jamming-machine was kept well oiled.

By the end of 1951, most of the other countries in the Communist orbit had commenced jamming operations of their own against Western broadcasts; and in the beginning of 1952, over 1,000 jamming transmitters were in continual operation.

By 1956, between 2,500 and 3,000 jamming transmitters were in operation against most major Western broadcasters, with particular attention to Radio Free Europe and Radio Liberty, which had begun broadcasting in 1951 and 1953 respectively.

On November 24, 1956 the first break in the electronic curtain occurred when jamming directed against RFE Polish-language broadcasts abruptly ended after a series of riots in the Polish city of Poznan, and the coming to power of a new Polish leader, Ladislaw Gomulka.
The Polish press had been complaining vociferously about jam-
ming, and there is considerable evidence to indicate that the people of Poland resented it. That is supported by the fact that during the first hours of the Poznan riots the local jamming station in that city was destroyed.
The hiatus in jamming against RFE Polish-language programs lasted 14 years. In 1970, following food riots in the north of Poland, jamming against RFE Polish programs was hastily resumed. Apparently caught off-guard, Polish authorities ordered that transmitters being used by Radio Warsaw in its external shortwave broadcasting service be redeployed and operated as jamming transmitters. Until jamming transmitters became available, Radio Warsaw international service was sharply curtailed.
Additional major breaks in the jamming pattern began in June 1963, when jamming directed against BBC and VOA broadcasting in the languages of the USSR, including Russian, were discontinued shortly after the conclusion of an atomic test-ban treaty. That marked the first time in 15 years that those programs were unjammed, and were another indica-
tion that a thaw in the Cold War had occurred. In July 1963, Romania stopped jamming all Western broadcasts; in February 1964, Hungary followed suit.
Two months later, Czechoslovakia stopped most jamming of BBC and Voice of America programs, but continued jamming RFE.
However, jamming can be turned on and off at the discretion of the Communist bloc; therefore, the situation proved $t$ be temporary. On August 21, $1 \times 3,200,000$ Warsaw Pact troops ir raded Czechoslovakia and within hours massive jamming of VOA, BBC and Deutsche Welle was resumed. it continued until 1973, when it was again discontinued.
The jamming transmitters no longer being used against BBC and VOA broadcasts were not taken out of service. Relations between the USSR and the People's Republic of China had worsened in 1973, and many of the jamming transmitters were rescheduled against Peking transmissions in Russia to the Soviet Union.
In addition, the Soviets had launched a major jamming effort against the Voice of Israel, whose broadcasts to the Soviet Union
called for a more liberal emigration policy toward Soviet Jews-a position not greeted with enthusiasm by the Soviet Politburo.
At the present time, all Radio Liberty programs beamed to the Soviet Union are jammed. In addition, Radio Free Europe programs in Bulgarian, Czechoslovak, and Polish are jammed. Radio Free Europe programs in Hungarian and Romanian are not. Deutsche Welle programs in Bulgarian are jammed and Voice of Israel broadcasts to the USSR in Russian, Hebrew, and Yiddish are jammed, as are People's Republic of China transmissions to the USSR in the Russian language.

On August 20, 1980, the Russians resumed jamming of the Voice of America, BBC, and Deutsche Welle. The resumption of jamming, after seven years, was generally thought to be due to growing labor unrest in Poland, and the Russians' fear that it could spread to the Soviet Union. BBC and Deutsche Welle programs in Russian were affected. Voice of America broadcasts in Russian, Ukrainian, Uzbek, Armenian, Latvian, Lithuanian, and Estonian were hit by noise jammers and Mayak jammers.
sions of power limitations, limiting the number of frequencies per transmission, protection ratios, and the gradual introduction of single sideband to the broadcasting service.
Evidently, rational planning in the face of jamming is a paradox. In addition, jamming against broadcasters such as the BBC and Voice of America can resume at any time, since the USSR has demonstrated in the past that it can turn jamming on and off like a faucet.
The United States is eager for the BC-WARC to succeed because this country firmly believes in the rational, equitable use of the high-frequency
broadcast spectrum. It is a certainty, therefore, that U.S. planning for WARC-83/84 will emphasize the technical standards and planning necessary to a successful conference. However, it would be naive to assume that the ugly specter of jamming does not loom over the Conference, or that it will not be a major impediment to its successful conclusion.

At WARC-79 the United States expressed its grave concern about jamming by entering a formal reservation when it signed the Final Acts of the Conference. This reservation states:
"The administration of the United
States of America, calling attention
to the fact that some of its broadcasting in the high-frequency bands allocated to the broadcasting service is subject to willful harmful interference by administrations that are signatory to these Final Acts, and that such interference is incompatible with the rational and equitable use of these bands, declares that for as long as this interference exists, it reserves the right with respect to such interference to take necessary and appropriate actions to protect its broadcasting interests. In so doing, however, it intends to respect the rights, to the extent practicable, of administrations operating in accordance with these Final Acts."R-E

## New software systems to spur office automation

Lack of integration is possibly the greatest weakness in the present rapid automation of office systems. Too often many of the benefits of partial automation are not realized. For example, a word processor may be obtained, but remain isolated from existing automated facilities, such as machine dictation, telephone, and facsimile.

By the mid-1980's, reports Frost \& Sullivan, Inc., international business research reporters, that piecemeal approach to business automation will be replaced by general systems offering a broad and integrated range of facilities: text editing, report formatting, teleconferencing, mail calendar, on-line data access, statistical
analysis, dictation, telephone services, document and information retrieval, dictionary, and other services.

Hardware components for such systems, says F \& S, are already available. Software is the primary bottleneck. Three fundamental challenges must be faced by software for future automated office systems: adaptability to changing user needs, suitability to extensive customization, and conformability to existing office practices (rather than vice versa).

## New fusion-reactor concept

Hope for an earlier solution of the problem of nuclear fusion is being held out by scientists of the University of Wisconsin. Their new conceptual nuclear reactor de-sign-which is yet to be tried out experi-
mentally-is called WITAMIR (Wisconsin Tandem Mirror). They believe that WITAMIR has many advantages over TOKAMAK, the Russian design up to now considered the most promising.
The new concept calls for a magnetic confinement vessel as long as a football field and roughly ten yards in diameter. Magnets would be positioned along the length of the chamber, and several extra coils placed at each end to reduce plasma leakage.
It is those extra coils, which create electrostatic and magnetic potentials that hold the plasma in the long central tube, that give the new concept an advantage over earlier ideas. "It is the first tandem mirror design that could produce electricity cheaper than a Takomak, and be easier to build and maintain," according to Gerald Kuchinski, leader of the University of Wisconsin's fusion-research program.

THIS IS THE FINAL INSTALLMENT IN OUR series on electronics and automobiles. This month we'll take a look at how electronics are helping carmakers de-sign-in self-diagnostics that aid the service and repair operations every car eventually requires.
An interesting concept in auto service is being developed at the General Motors Service Development Center at Warren, MI. When you buy a new car, the selling dealer installs a small, inexpensive radio transmitter that sends out a signal with your personal code.
When you reach the dealership and drive up to the service garage, a receiver reads the code. The garage door is opened and the code is passed to a computer where your service records are stored.

By the time your car is in the door, the service technician knows who you are and has a complete service record of the car, including a flag on any recalls. The computer terminal even prints out your repair order, saving another aggravating delay.

That personalized code, by the way, would be your vehicle's VIN (Vehicle Identification Number). New Federal requirements are increasing it from 13 digits to 17. There may be an added benefit in transmitting that code: it would make police recovery of stolen vehicles much easier. Unfortunately, it would also make computerized ticketing for such offenses as speeding a technological breeze.

## Check engine

A new telltale signal (idiot light) is showing up on car dashboards. It reads CHECK ENGINE, or something very close to it (Fig. 1). James G. Vorhes, General Motors Vice President in charge of Consumer Relations and Service, explains why it's there:
"The Computer Command Control system (see Fig. 2 -editor) has been on most of the GM gasoline-engine automobiles in California since the start of the 1980 model year and will be on nearly all cars sold in this country in the 1981 model year. It's the largest new use of computer controls in the history of the industry and we will be prepared for it.
"One of the most exciting features of the system is its ability to automatically diagnose the cause of any malfunction. We have designed-in an extremely high degree of reliability. Like all systems, it can malfunction. But, unlike other systems, it will help diagnose itself.


Electronic self-diagnostics are taking the guesswork out of auto repair, and that should make you, and your mechanic, a lot happier.



FIG. 1-A NEW TELLTALE SIGNAL informs motorists of a malfunction. It's part of GM's new Computer Command Control system.


FIG. 2-KEY FEATURES of GM's Computer Command Control system. It is offered on nearly all of their 1981 cars.
"Mechanics will not have to become electronic engineers to work on the system. In fact, there will be no need for expensive computer testers. All a mechanic will need is an ordinary dwell meter, a test light, some jumper wires, a vacuum source, a tachometer, and a digital voltmeter. Most shops already have this equipment. Any qualified technician who can work on current engines will be able to fix one equipped with the system."

The check engine light is there to aid both the motorist and the mechanic.

When it lights, indicating a malfunction, the motorist will still be able to drive the car in for service. Once there, the Computer Command Control system tells the technician which system may have the problem by flashing a code. Then the technician goes through a simple factory-provided diagnostic routine to determine whether the component, a wire, or a connection is at the root of the problem.

## Self-test at Ford

'Self-test in its various forms is
going to reduce diagnostics to the very simple goal of determining which part is bad. We, the industry, are headed that way.'
That's the forecast of Walter Doelt, the fellow at Ford in charge of this area.
Radio-Electronics was privileged to enjoy an exclusive interview with Mr. Doelt dealing with a wide range of subjects relating to the role of electronics in designing a car for serviceability. Ford's commitment to microprocessors is very strong.

## "They provide the best economic

 approach to providing the widest degree of freedom in control strategy, and to very flexible control systems. The key is mandated performance requirements versus drivability."Mandated performance requirements include emission and fuel economy goals, both in terms of government regulations and corporate marketing and design objectives.

Mr. Doelt continued:
"When implemented with a microprocessor, you can freeze the basic design early in the design cycle, then work the calibration in software later.'
Ford has been working to reduce the component count in their various onboard systems (engine control, dashboard, etc). Systems that used seven or eight IC's in 1978 were down to four in 1979, two in 1980, and should be down to one in the near future. Since the majority of failure modes are in the interconnections of IC's, single-IC microprocessors are very popular.

The 6800 series ( 6809 , etc.) is used for Ford's advanced Electronic Message Center dashboard. The 8048/8049 series is used for engine control where continuous timing is not required. The company also uses a number of custom circuits in integrated control systems (ignition, carburetion, fuel metering, and carburetor feedback) in their top-of-the-line models. Motorola, Intel, Toshiba, and Texas Instruments are among Ford's suppliers.

For example, 1980 models sold in California included a new feedback carburetor design using an Intel 8048; earlier designs used a number of discrete analog IC's and devices. By incorporating the 8048 , Ford was able to include a limited self-test feature.

## Lead times

One important point about new developments is that it takes time to incorporate them. For example, 1981 model year cars were introduced in October of 1980. The pilot runs of those cars began in July 1980. The control units to go into those pilot production vehicles had to be built by late March 1980. So the microprocessors had to start as silicon slices way back in December of 1979 , with design requirements frozen even earlier. But each
year brings more advancements.

## STAR performance

One of the most intriguing aspects of the Ford self-test system is the way the under-hood microprocessor "talks" to the service technician: it uses a digital code of $1 / 2$-second pulses, $1 / 2$-second apart, with two seconds between digits.

That pulse timing was decided on to let any service technician read the code with just a VOM. But there's a nice alternative called STAR (Self-Test Automatic Readout) shown in Fig. 3. That handheld, calculator-sized gadget translates the pulses and displays the result on a two-digit LED readout. The self-test sequence can be initiated from a front-panel pushbutton. A readout of "11," for example, translates to "test complete, everything normal."

The STAR costs about \$70. They're being built for Ford by the Hickok Electrical Instrument Company. Hickok also sells an inexpensive DMM to Ford for their service technicians.

One feature of the Ford self-test system is that it's designed so all systems test the same way, no matter which model is being examined. There are also a number of universally-applicable manual test operations.

For example, a testing unit will read the voltages at various sensors through a selector switch. The switch setting, selected sensor and anticipated voltages are the same from vehicle to vehicle. Another example is that any vehicle's self-test sequence will, at a given point, call for an EGR (Exhaust Gas Recirculator) flow of $30 \%$, then check the position sensor on the throttle body. At other points, hydrocarbon and carbon monoxide exhaust tests will be performed, both at idle and, with the engine unloaded, at some higher RPM.

Testing on demand in that way is well within the capabilities of both engine and electronic technology. Just beyond is the goal of full-time testing, where a car monitors its own performance periodically during normal driving.

## Noise and interference

The biggest problem EMI used to cause in a car was noisy radio reception. But today, with multiple microcomputer systems and miles of harness wire on board, anti-EMI measures are a high priority in design.

Communication links between modules are designed for very low impedance. Modules are housed in metal enclosures, with both the top and bottom grounded. Measurements from analog sensors are performed ratiometrically to make them less sensitive to power-supply variations. Sensor wires are very carefully routed, especially around ignition areas. Diode and capacitor buffering is used on all I/O lines. Roll-off filters block higher and


FIG. 3-SELF-TEST AUTOMATIC READOUT, or STAR, translates a diagnostic pulse string into an easily read fault code.


FIG. 4-THE VIN (Vehicle Identification Number) can be written as a bar code that may be read by a portable data terminal.
lower frequency signals. And the buzzer has become obsolete.

Buzzers created tremendous noise problems, generating harmonic-filled spikes with hundreds of volts in them. Now they're being replaced with electronic tone and chime generators that, surprisingly, cost about the same

Ford and others are taking a serious look at fiber optics as a possible answer to many noise problems but, in Walter Doelt's opinion, they won't be incorporated until there is some significant improvement in the cost of material, the cost of terminations, and the performance of optoelectronic links.

Another area being explored is the digitization of position sensors. That permits both noise reduction and the reduction of wiring through time-division multiplexing.

In the near future, non-volatile RAM will be used more and more, especially since it permits alterable test criteria as improvements are developed. Also, the lower cost of momory is going to permit designers an enormous increase in program lineage, with 10 K chunks of ROM replacing each 1 K available today.

One important question is what happens in the fail-fail mode? What if the
computer crashes so badly that it can't even yell for help? The answer is that all systems are designed so that in any event of computer failure, the car keeps going. Those "baseline conditions" may keep the engine running, but there's no way the driver can fail to notice that there is a problem. The car bucks, surges, gurgles, and wheezes like a sick hippo. And yes, that "reduced drivability" is also designed-in.

## GM developing the TOUCH

The theory is that explaining a probblem is the biggest part of solving it, or, as expressed in a recent GM press release, ".. once a symptom is accurately described, there is one set of most probable causes." Either way, the idea has led to a system being developed by GM called TOUCH (Touch-Operated Universal Communications Helper).

TOUCH is a computer system that asks a customer questions about his problem, and uses each answer to formulate either a new question or a "hunch" as to what the problem might be. And it can even be used when the repair facility is closed.
The customer could, for example, drop his car off at night. TOUCH might look very much like a 24 -hour banking terminal. If the car is making a noise, TOUCH might ask where it's coming from and display a picture or diagram of the car to help the customer describe the location. It could ask what the noise sounds like, prompting the customer with either descriptive words, recorded sounds, or a sound-effects generator. Then it might ask if the noise happens all the time, or just when climbing a hill, coasting to a stop, turning a corner, or whatever.
Then, if the customer has described a squealing sound that comes from the front of the car during turns for example, TOUCH would print out a repair order telling the technician to check the power-steering pump belt for slack. You put your keys in an envelope and TOUCH takes them, gives you a receipt, and tell you when your car should be ready.

## More future goodies

That 17-digit VIN we mentioned earlier appears in more places this year, and it will appear on still more in the future. GM is translating it into a bar code so it can be read by wands and portable data terminals (Fig. 4). In addition to helping identify cars being recalled for various reasons, the service history of an individual car, and even the identification of its owner, the VIN is important in recovering stolen cars. As it appears on more and more major components-such as engine blocks, frames, and glass-it makes it easier for law enforcement operations to locate continued on page 88

# ANTI-MECATILE FEEDBAGK FOR HI-FI ANPLIFIERS 


#### Abstract

High levels of negative feedback can actually add to dynamic distortion. Here's a look at a totally new amplifier that uses power MOSFET's and eliminates conventional negative-feedback circuits.


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WHEN NEGATIVE FEEDBACK WAS FIRST used in an audio amplifier, back in the early days of vacuum tubes, it seemed as though it would be the panacea for all audio-distortion problems. Add 20 dB of negative feedback to any old amplifier having a harmonic distortion level of $10 \%$ and presto, the distortion drops to $1.0 \%$. Add another 20 dB to that, for a total of 40 dB , and distortion is reduced by another whole order of magnitude, to $0.1 \%$. Of course, with each addition you lose 20 dB of gain or so, but that's easily made up by adding extra stages of amplification that are not all that expensive.

The age of solid-state electronics made the application of high levels of feedback even more attractive to design engineers. After all, the transistor is not the most linear of amplifying devices (and is therefore better suited to switching applications than it is to linear audio-amplification service), and feedback could be used to cover a multitude of sins. Unfortunately, as engineers found out much later, it could also introduce new sonic aberrations to audio amplifiers-defects that were not apparent from static single-tone bench measurements but were definitely audible when the amplifiers were called upon to amplify real-world music signals. By now, all of us have read about transient intermodulation distortion, and other forms of dynamic distortion, that are aggravated through the use of inordinately high amounts of negative feedback.

It is not surprising, therefore, that many manufacturers of audio equipment, both here and abroad, have been addressing the problems associated with negative feedback and finding
ways to eliminate those problems. Some companies, such as Lux Audio and Onkyo have come up with double feedback loops, the first of which is of lower -than-usual magnitude (typically, 30 to 40 dB instead of the higher 60 to 80 dB commonly encountered in solid-state amplifier designs), while the second, generally referred to as a servo-DC feedback loop, addresses the problem of feedback in the infrasonic region. That helps to stabilize DC-amplifier designs, reduce DC drift, and act as a sub-


FIG. 1-MONAURAL power amplifier, the model TA-N900 from Sony does not use negative feedback.


FIG. 2-HIGHLY-MAGNIFIED view of the new MOSFET chip.
sonic filter all at once.
Still other companies, such as Sony Corporation, have taken a different approach. At the recently held Tokyo Audio Fair, they introduced a new highquality monaural power amplifier, called the model TA-N-900, pictured in Fig. 1. That amplifier uses an entirely new circuit that eliminates the conventional negative feedback circuit used in other amplifiers. One of the developments that makes it possible is the new 2SK173 power MOSFET recently developed by Sony. Figure 2 shows a highly magnified view of the chip used in the construction of that power MOSFET while in Fig. 3 we see a cross-sectional diagram.
Similar to a standard FET, a power MOSFET has a high input impedance and low storage time. It is characterized as having a wide area of safe operation and high reliability. The double diffusion type power MOSFET developed by Sony also has a very high gain and a high maximum rated voltage of 210 volts. As can be seen in Fig. 3, the drain of the MOSFET is at the same potential as the case so that there is no increase in capacitance between the drain and heat sink (when used as a source follower) and no degradation in frequency response.
A simplified partial schematic diagram of the amplifier circuitry is shown in Fig. 4. The first predriver stage consists of a differential-input double-cascode bootstrap circuit, using junction FET's and bipolar transistors. It has been designed so that high-frequency distortion caused by non-linearity of the FET's is reduced to minimum limits and also to achieve thermal stability through the use of a cascode-connected


FIG. 3-CROSS-SECTIONAL diagram of the 2SK173 power MOSFET recently developed by Sony.
current-mirror load and finally to improve power-supply ripple-rejection characteristics.

The second stage consists of a cascode bootstrap inversion-amplifier using a bipolar transistor. It provides good linearity and, like the first stage, features high power-supply ripple-rejection characteristics. The final stage of the prepower stages is a complementary
emitter follower single-ended push-pull output circuit. All stages up to that point have light loads and are isolated from the speaker load and the power stage, so that there can be virtually no problems caused by reactive output loads. As a result, negative feedback can be used around those stages without affecting power-stage performance.
The power-output stage itself is a
source follower single-ended push-pull output circuit using four of the new power MOSFET's connected in parallel. The driver stage has a distortionreduction circuit that compensates for the nonlinear characteristics of the output devices. The power MOSFET's operate in the Class A mode.

As can be seen in Fig. 4, there is no negative-feedback loop from that power stage to the pre-power/driver stages. Despite the lack of an overall loop-feedback circuit, the amplifier is able to achive remarkably low distortion figures: less than $0.05 \%$ THD at 200 watts output into 8 -ohm loads and less than $0.05 \%$ IM distortion for the same output. The 200 watts rating applies to all load impedances from 8 ohms down to 2 ohms. In the direct-coupled mode, frequency response remains flat from DC to $100 \mathrm{kHz},-3 \mathrm{~dB}$. Damping factor, normally one of the first parameters to suffer in the absence of high amounts of loop feedback, remains a high 50 (referred to 8 ohms) and slew rate is an impressive 150 volts-per-microsecond.

## Pulse-locked power supply

Since a power amplifier is called upon to transfer high levels of currents into a


FIG. 4-THERE is no negative-feedback loop in the model TA-N900, as can be seen in this simplified schematic of the amplifier circuitry.


FIG. 5-SCHEMATIC DIAGRAM of the power supply. A diode bridge is used to rectify the incoming line voltage.


FIG. 6-INTERNAL VIEW OF THE model TAN900 amplifier.


FIG. 7-TRANSFER CURVE for the 2SK173 power MOSFET.
speaker load, the power supply of such an amplifier plays an important role in the overall design. In the model TAN900, Sony chose to use a pulse-locked power supply that offers extremely good regulation, low output impedance, low hum and noise levels, as well as other advantages. A diagram of the power-supply circuit is shown in Fig. 5. The circuit rectifies incoming line


FIG. 8-TRANSFER CURVE for the 2SJ54 (Pchannel) power MOSFET.


FIG. 9-INPUT and output waveforms for a 625kHz signal.
voltage directly, by means of bridgeconnected diodes. The section identified as the $20-\mathrm{kHz}$ power oscillator generates a $20-\mathrm{kHz}$ squarewave signal using four high-power switching transistors in an oscillator circuit that supplies the needed power to the converter transformer.
Because that transformer handles a high frequency signal of 20 kHz , it can
use a ferrite core that has low high-frequency losses. Compared with transformers that can handle conventional power supplies (operating at 50 or 60 Hz ), this transformer can be constructed with fewer windings that, in turn, results in an extremely low internal impedance. The secondary rectifying circuit shown in Fig. 5 rectifies the $20-\mathrm{kHz}$ output, using high-speed diodes, to convert the voltage into positive and negative DC. Finally, the pulse-width control circuits shown in Fig. 5 sense the positive and negative DC potentials that are applied to the final stage of the power amplifier and control the pulse width of the $20-\mathrm{kHz}$ signal to maintain the output voltage at a constant level for both positive and negative supplies. The filter capacitors are $22,000 \mu \mathrm{~F}$ units that are incorporated at the output of the power-supply circuitry to insure that the amplifier can handle high-level transient signals without difficulty.
An independent series-type constantvoltage supply is provided for low-level stages (including the driver stage), completely separate from the pulse-locked power supply so that there can be no interference from the power-output stage to the low signal-level stages.
Referring once more to Fig. 4, there is a protection circuit that senses and monitors the DC voltage of the power amplifier, the temperature of the power MOSFET's, and the load impedance. That protection circuit ${ }^{2}$ will cut off the output should either the connected load (speaker) or the power amplifier approach operational limits.

The amplifier is equipped with a speaker-impedance switch that allows the speaker to be driven in the most efficient and stable manner possible, regardless of whether is has an impedance of 8.4 , or 2 ohms. The impedance switch simply controls the voltages produced by the pulse-locked power supply
continued on page 87


TEMPTED TO BUY THAT GRAB BAG LOADed with unidentified IC's? Go aheadthose IC's have a signature that will tell you what pins are probably outputs. From there on it's easy-a few voltage measurements, some current measurements, and you should know what you've got. You should even be able to determine if the device is defective. So grab your trusty ohmmeter and get ready to record your first IC signature.

An IC signature is an array of resistance readings derived from the IC and displayed in an organized way. The $\times 100$ range of an ohmmeter is used. (Be sure you know which ohmmeter lead is positive; some ohmmeters change polarity when switching from volts to ohms.)

The signature is obtained by recording the resistances between all terminal pairs of the IC. Use the form shown in Fig. 1. Connect the ohmmeter's positive lead to pin 1, and move the negative lead sequentially through the remaining pins. Record the measured resistances across the top row of the signature chart. A resistance measurement of over several
hundred thousand ohms does not convey very much useful information, so there is no need to record it-put a dash through the box instead.
Move the positive lead to pin 2 and fill in the second row of the chart by moving the negative lead to pin $1,3,4, \ldots$, etc. Continue in the same manner until every row of the signature chart is completed. If this is done properly, you should have as many rows in your chart as there are pins on the IC.

The steps that follow show how to use the completed signature to identify your IC.

Step 1: Examine the chart and circle each terminal-to-terminal resistor-you can tell which ones those are because each purely resistive connection between two terminals reads approximately the same in both directions.

For example: In Fig. 2 there are 12 circled boxes, 6 above the diagonal and 6 below. The circled number in Row 5, Column 3 has its mirror image on the opposite side of the diagonal in Row 3, Column 5. The resistance is 7 K ohms in both directions and it is therefore a ter-
minal-to-terminal resistance. That is noted to the right of the chart (Fig. 2), along with the other resistance values and identified as step one. The remaining terminal pairs show grossly different resistance measurements in opposite directions, indicating the presence of one or possibly several semiconductor junctions in the path.

It is highly unlikely that a TTL IC, or for that matter any linear IC, would contain 6 identically valued terminal-toterminal resistances. (Maybe the IC is RTL or DTL?)

Step 2: Disregard all circled boxes and scan the signature to locate the row with the lowest resistance readings-Row 4 in this case. That uniquely identifies pin 4 as the substrate connection of the IC or, in other words, the most negative terminal of the IC.

Scan across Row 4 for the lowest uncircled reading-in this case it is the 750 -ohms reading in Column 11. That tells us that pin 11 is the $\mathrm{V}_{\mathrm{CC}}$ terminal of the IC. Record those numbers in the place provided at the right of the chart-Step 2, Fig. 2.


FIG. 1-THIS FORM is used to record all resistances between terminal pairs of the unknown IC.

The other uncircled low-resistance readings in the ground row usually identify transistor collectors; i.e., output terminals. That is an important clue to be used later.
Step 3: Before proceeding to the identification of other terminals we measure $\mathrm{I}_{\mathrm{CC}}$. Apply a low voltage, say 3.6 volts (RTL supply voltage), to the IC through a 100 mA milliammeter. The positive voltage goes to the $\mathrm{V}_{\mathrm{CC}}$ terminal (in this case pin 11) and the return connects to the IC substrate (in this case pin 4).
To protect the IC and the equipment, place a 120 -ohm resistor in series with the current meter. A dead short in the IC will only draw 30 milliamps. Remove the resistor and re-connect the current meter only when it is clearly safe to do so. Most standard TTL gates draw between 2 to 4 mA . Thus, a quad NAND or NOR would draw 12 to 15 mA . In the case at hand, there was no current flow at all.

DTL or TTL would have shown some current-so again the evidence suggests RTL.

A third clue: If there is a normal current flow, raise the voltage to 5 volts, measure, and record $\mathrm{I}_{\mathrm{CC}}$ in the space provided at the right of the signature chart.

Step 4: Remove the milliammeter and apply the selected voltage directly between the $\mathrm{V}_{\mathrm{CC}}$ and ground pins. Measure volts-to-ground, mA-to- $\mathrm{V}_{\mathrm{CC}}$ (through a 330 -ohm resistor) and mA-to-ground for each pin of the IC. Record the measured values in the rows at the bottom of the signature chart.

The "volts-to-ground" row generally identifies all inputs and outputs. Voltages from about 2.2 volts up to the applied voltage indicate outputs in the high state (for a logic chip). Thus, pins 3, 5, 8 , and 14 are likely candidates for output terminals. (You will recall in step 2 that


FIG. 2-COMPLETED SIGNATURE CHART for the unknown IC. It turned out to be an quad, 2-input NOR gate.


FIG. 3-SIGNATURE CHART for a TTL 7400 IC. All but a few TTL IC's have this typical two row signature.
those are the same terminals that were suggested as outputs by their low readings in Row 4.)

A voltage less than 0.2 , but greater than zero, usually indicates logic outputs in the low state. None of those appear in Fig. 2.

Now is the time to remove and reapply power to the IC. Do that several times, each time comparing the voltage at each suspected output to its original recorded value. Often a flip-flop will reveal itself by changing the state of one or more of its outputs. A simple gate will never change state in response to that little trick. The IC in Fig. 2 did not change state so I assumed it was not a flip-flop.

Voltages from about 1.8 down to 0.8 usually indicate TTL or DTL inputs. The fact that there are no such voltages in the "volts-to-ground" row of Fig. 2 was certainly a surprise to me, but it did lead to a pretty solid conclusion: If the IC is not defective, then it is not TTL or DTL.

Currents in the low state should read 10 to 20 mA when measured between the output and $\mathrm{V}_{\mathrm{CC}}$. Currents in the high state can read anywhere from 2 to 30 mA when measured between the output and ground if the IC is TTL. As an example of a typical TTL signature, Fig. 3 shows the signature chart for a 7400 TTL IC.

Input currents for RTL, DTL and TTL fall between 0.8 mA and 2.0 mA . In Fig. 2 all the probable inputs draw 1.4 mA referenced to $\mathrm{V}_{\mathrm{CC}}$ and nothing referenced to ground. That verifies that they are inputs and shows they are active (draw current) when the input is pulled high. RTL is active-high. DTL and TTL are active-low. Since their appear to be twice as many inputs as outputs, the chart suggests that our IC is a quad gate of some sort. It is reasonable to conclude that pins $1,2,6,7,9,10,12$ and 13 are inputs.

The bottom row of the chart shows the outputs provide only 1 mA to ground despite the fact that the voltage mea-
sured at those terminals is 3.6. That suggests an internal pull-up resistor connected to the output terminal (see Fig. 4). If that is so, and the device is a quad gate (which seems very likely), there should be four identical resistors to $\mathrm{V}_{\mathrm{CC}}$-one from each output. And that implies we should read twice the pull-up resistor value between any two outputs. In that case, the circled 7 K values in the signature point to 3.5 K pull-ups in each output. With a $\mathrm{V}_{\mathrm{CC}}$ of 3.6 volts applied, grounding any output through the current meter should cause a current flow of just about 1 mA . And that's what we got! List the outputs and inputs on the signature chart.

Step 5: The symmetrical pattern of resistances in the signature and the strong evidence for four independent outputs with logic-level voltages pretty much rules out any linear IC. Resistive pull-ups could be DTL, but DTL inputs are active-low and our IC is active-high. After reviewing all the evidence I felt there was absolutely no doubt that this device was RTL. That conclusion was recorded in Fig. 2.

Step 6: We now manipulate the inputs and observe the output responses to determine what kind of logic device we have.

With $\mathrm{V}_{\mathrm{CC}}$ applied, we connect a voltmeter from ground to a terminal thought to be an output. Ground the inputs one at a time, noting the change, if any, in the metered output. If that output does not change state for any grounded input, repeat the procedure, this time connecting one input at a time to $\mathrm{V}_{\mathrm{CC}}$ instead of ground. In this example it happened that pin 3 went low when either pin 1 or pin 2 was pulled high (to $\mathrm{V}_{\mathrm{CC}}$ ). None of the other outputs responded to changes in pin 1 or pin 2 . This indicates that pins 1 and 2 are inputs to one gate whose output appears on pin 3. That procedure is continued until all inputs and outputs are related in some way. Truth tables can be consulted to identify the gates. This device furned out to be a quad, 2-input NOR gate.

The relationships between the inputs and outputs and the conclusion as to the type of device I was dealing with are listed in Fig. 2 as step 6.

Had the device not responded at all to any of the above techniques, I would have tried exercising two, or even three, inputs at a time and I would have begun to search for a possible "enable" or "inhibit" input. The more complicated devices require a little ingenuity and some intelligent guesswork.

Step 7: Use the results of step 6 to draw the schematic of the IC. At that point the device could be used in the average hobby project without needing to know anymore about it. But, if you feel compelled to assign a number to your IC, its time to consult the IC data books. That's what I did.

Step 8: It took quite a while to locate a


FIG. 4-LOGIC input and output circuits. Use these circuits along with your resistance measurements to determine the logic family of the unknown IC.

## WHAT MAKES THE IC SIGNATURE POSSIBLE

Practically all IC outputs, linear or digital, are formed from transistor collectors. All NPN collectors are imbedded in a $P$-type substrate that is designated ground ( -V for linear IC 's). As shown in the accompanying diagram, the collector and substrate form a P-N junction that, like any other diode, conducts well in one direction and poorly in the other.

Connecting an ohmmeter from substrate to collector in the forward direction (positive lead to substrate) will cause the ohmmeter to indicate between 500 and 900 ohms. Other diodes in the same IC will read between 950 and 1300 ohms. Actual resistance values will vary with the type of ohmmeter and the degree of doping in the IC, but the IC outputs will always give the lowest readings.

Thus it is possible to locate every output terminal on an IC. The row containing all those low-resistance readings will be the ground $(-\mathrm{V})$ row.

In every IC there are usually several transistors whose collectors are connected to $\mathrm{V}_{\mathrm{CC}}$ either directly or through

some resistance, When reading forwara resistance from substrate to $\mathrm{V}_{\mathrm{CC}}(+\mathrm{V}$ in linear IC's), that multiplicity of paths will give a lower reading than any other terminal on the IC.
Thus it is possible to identify the $\mathrm{V}_{\mathrm{CC}}$ terminal.

## A LOGIC-FAMILY TREE

Mention is made in this article of the RTL (Pesistor-Transistor Logic), DTL (Diode-Transistor Logic) and TTL (Tran-sistor-Transistor Logic) families. Of the three, TTL is the only one that is still in common use, but a look at its predecessors is worthwhile. (Refer to Fig. 4.)

As advances in technology have made it possible to construct more complicated devices on a silicon chip, we have been able to take advantage of their sophistication to create faster and more elaborate logic families.

All three of those logic-families IC's work by causing their output transistors to go into saturation (a condition where no amplification takes place - only conduction) but differ in the way input signals are processed to bring about that state.

RTL was the first IC logic-family to find widespread use. Each input line going to the output transistor contains a resistor. Its purpose is to reduce the amount of current consumed by the device and to isolate the logic-gate inputs. The input voltage passed through the resistors drives the output stage into saturation, making the collector voltage of the output transistor drop and causing the output to go "low."

The resistors, though, slow down the switching speed of RTL devices because they increase the time needed to charge and discharge the input capacitance of the output transistor.

Typically, RTL has a switching speed on the order of 50 nanoseconds and operates from a 3.6 -volt supply.

The next step in IC evolution was DTL. That family substitutes diodes for the resistors used in RTL. The diodes provide better isolation at the inputs and, because of their low forward resistance, make it possible for DTL circuits to switch more rapidly than their RTL equilvalents.

DTL has a typical switching speed of 25 nanoseconds and requires a fourvolt supply.

Finally, TTL uses multi-emitter transistors in the input stage. The base-collector junction of those transistors is never fully off, meaning that a state of saturation can be reached considerably more quickly than with either RTL or TTL.

Switching speeds for simple TTL IC's are frequently under 10 nanoseconds. TTL uses a five-volt supply.

While it is still possible to find RTL and DTL IC's on the surplus market, the TTL family is now the dominant one. Its two most common forms are standard TTL and "LS" (Low-power Schottky) TTL, the latter being even faster and having a lower power consumption, at a small sacrifice in drive capability.


FIG. 5-LINEAR DEVICES can also be identified using the signature chart. This is the signature of a 741 IC.


FIG. 6-PIN-OUT for the 741 IC is shown in this schematic diagram. There are 1 K resistors between each offset-null terminal and -V.

Motorola IC book containing RTL data sheets. The electronics department at the local college was good enough to let me look through their copy. If you need that kind of assistance, let me urge that you make the local college your first stop. I wish I had-it would have saved a lot of time.
The Motorola book had 256 pages of RTL data-whew! Fortunately the plastic-case style of my IC eliminated two of the three RTL sections. The index of the remaining section listed only two IC's that were quad 2 -input NOR gates. The collector pull-up resistors of the first IC type were nominally 640 ohms. The collector resistors in the second IC type were nominally 3.6 K . Bingo! (We guessed 3.5K in step 4-not bad!).
The device is without a doubt an MC $717 \mathrm{P} / 817 \mathrm{P}$ and all the information on that data sheet applies to this IC. I am unable to differentiate further between the 717 P and its higher-performance counterpart the 817P. Since the safer move is to assume that the more restricted temperature range applies, I declared the device to be a Motorola RTL IC, type MC717P.

Those techniques work on linear devices as well as digitals.

For example: A signature of an 8-pin DIP 741 op-amp is shown in Fig. 5. Note that the low-resistance row still identifies the substrate, -V , for an opamp , and also that the lowest reading in that row identifies $+V$. The only other uncircled reading in row 4 is 950 ohms in column 6 , identifying pin 6 as the output terminal.

Figure 6 shows the 741 schematic with the 8 -pin DIP pin-out. Note the nominal 1 K resistors from each offsetnull terminal to -V . The circled resistances in Fig. 5 illustrate some interesting facts about IC resistors: They do not always read the same in both direc-tions- 750 ohms one way, 850 ohms the other (pins 1 and 4, and 5 and 4), and they may deviate quite a bit from nominal (1K). Nevertheless the offset-null-terminals are clearly identified.
The op-amp inputs are almost impossible to identify from a signature, but the information already obtained is enough to identify the IC in the data books.

I highly recommend that all those interested in identifying IC's read Olson and Zevnik's excellent article in the January 1980 issue of Radio-Electronics, "How to Identify Unmarked IC's." The article is filled with useful suggestions for getting advance information from the PC board.

The smart tinkerer uses all the clues he can get. But when there aren't any advance clues-when there is no PC board-when there are no marking to go by-then the only alternative is the IC signature.


This up-to-date version of an old favorite makes an amusing conversation piece, and a great project!

NOEL NYMAN


## THE

STATE-OF-THE-ART OF DOING NOTHING

A POPULAR CONSTRUCTION PROJECT some years ago was the "do-nothing" box: one or more neon lamps that flash one-at-a-time either randomly or in sequence. The basic circuit is a relaxation oscillator based on the neon-lamp characteristic of firing at about 65volts. Figure 1 shows a one-lamp flasher. The capacitor charges at a rate determined by the R-C constant. When the neon trigger voltage is reached the lamp flashes, discharging the capacitor and the cycle repeats. Figure 2 shows circuits with more lamps.

To get the lamps to fire in sequence requires careful mismatching of the lamps. Because neon lamps characteristics change as they age, the sequential firing may deteriorate after a while. For more information on neon flashers, see the October 10, 1958 engineering issue of Electronics magazine.

The popularity of that circuit was probably due to the bright display that "moves" with no moving parts (remember that was twenty years ago). The parts were cheap and easy to obtain and the power supply was simple. Because of the low current required you could even run it from batteries (expensive ones).

Integrated circuits and LED's have upgraded the "do-nothing" box considerably. A simple sequential circuit


FIG. 1-ONE-LAMP flasher fires when the neon trigger-voltage is reached.

Although that is a clever modern way of doing nothing, the Erasable Programmable Read-Only Memory (EPROM) has made it obsolete also. In case you're unfamiliar with them, EPROM's are field-programmable IC memories that retain their stored information even with circuit power turned off. The programmed information appears on output lines when the appropriate logic levels are placed on address lines. The EPROM used for


FIG. 2-MULTILAMP flasher circuit. The neon lamps must be carefully mismatched if they are to fire in sequence.
of ten LED's requires only two IC's as shown in Fig. 3. That circuit can use inexpensive batteries, fires sequentially, and the clock frequency is easily changed.
this project can store up to 2,048 eightbit words ( 16,384 bits) if it's a 2716 , or up to 1024 eight-bit words $(8,192$ bits) if it's a 2758 -which is really half of a 2716 .


FIG. 3-THIS simple "do-nothing" circuit will light ten LED's in sequence and requires just two IC's and a few discrete components.
eighth selects whether the line or circle is displayed.
I used the least-significant data bit, $\mathrm{D} \emptyset$, to select the display. Logic-1's at the other seven data bits will light the corresponding numbered LED's. As an example, Table 1 shows the program for lighting each LED in turn clockwise around the circle. The 1 in $\mathrm{D} \emptyset$ selects the circle. The choice of 1 or $\emptyset$ for the circle is arbitrary and is determined by how you wire the circuit. Table 2 shows a more complex program. The first seven steps circle the display counterclockwise. In step eight, D $\emptyset$ goes to $\emptyset$ and the 1 at D2 lights LED2 in the line display. In the remaining steps the dot moves to LED1, then lights the line from right to left leaving the LED's on as it moves.

Most of us don't have EPROMprogramming equipment. For a small fee, EPROM's can be programmed by various sources. Check with your local parts supplier or write to EPROM manufacturers. Also, some of the ad-

TABLE 1

| $D 7$ | $D 6$ | $D 5$ | $D 4$ | $D 3$ | $D 2$ | $D 1$ | $D \emptyset$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | 0 | $\emptyset$ | 1 | $\emptyset$ | 1 |
| 0 | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 0 | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | 1 | 0 | $\emptyset$ | $\emptyset$ | 0 | 1 |
| $\emptyset$ | 1 | $\emptyset$ | 0 | 0 | $\emptyset$ | $\emptyset$ | 1 |
| 1 | $\emptyset$ | $\emptyset$ | 0 | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 |

## PARTS LIST

All resistors $1 / 4$ watt, $5 \%$
R1-R4- 10000 ohms
R5-470,000 ohms
R6- 100,000 ohms, trimmer potentiometer
R7-R13-270 ohms
R14-R16-4700 ohms
R17-390 ohms

## Capacitors

C1-50 $\mu$ F , electrolytic
$\mathrm{C} 2-0.01 \mu \mathrm{~F}$ ceramic disc

## Semiconductors

IC1-4013B dual-D flip-flop
IC2-4011B quad 2 -input NAND gate
IC3-4069B hex inverter
IC4-4040B ripple counter
IC5-2716 or 2758 EPROM*
IC6-7404 hex inverter
IC7-7805 five-volt positive voltage regulator
Q1, Q4-2N4400 or other small-signal NPN-type
Q2, Q3-2N4402 or other small-signal PNP-type
LED1-LED13-jumbo red LED
S1-S3-momentary push-button switch (N.O.)

Miscellaneous: perforated construction board, IC sockets, wire, solder, etc.
*A pre-programmed 2758 EPROM showing several circle and line variations can be obtained postage paid for $\$ 25$ from: Noel Nyman, MAB, P.O. Box 88868 , Seattle, WA 98188

TABLE 2

| D7 | D6 | D5 | D4 | D3 | $D 2$ | $D 1$ | $D \emptyset$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 1 |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ | $\emptyset$ |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | $\emptyset$ |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 1 | $\emptyset$ |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 1 | 1 | $\emptyset$ |
| $\emptyset$ | $\emptyset$ | $\emptyset$ | 1 | 1 | 1 | 1 | $\emptyset$ |
| $\emptyset$ | $\emptyset$ | 1 | 1 | 1 | 1 | 1 | $\emptyset$ |
| $\emptyset$ | 1 | 1 | 1 | 1 | 1 | 1 | $\emptyset$ |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | $\emptyset$ |

Using an EPROM to control a "donothing" box allows for more than just sequential operation. LED's can be made to circle clockwise, counterclockwise, or to alternate direction. The LED's can stay on as they circle. Opposite LED's can appear to rotate around the circle. Using a straight line, other displays can be generated. A dot can bounce from one end to the other. Lighting pairs from the ends converging on the center is another variation.

The original idea for a PROM (nonerasable EPROM) "do-nothing" circuit was suggested by Todd Kitajo of Almac Stroum Electronics, Seattle, WA. His design used a PROM to control two straight-line displays, one red and one green. I used LED's arranged in a line and a circle. The circle is made up of seven LED's, $51^{\circ}$ apart. A sevenLED line bisects the circle. The circle and line share LED6. Seven EPROM data bits control the LED's and the
vertisers in the back pages of RadioElectronics can program EPROM's.
Your program will have to be submitted in a form that programming machines can read. Punched paper tape and punched cards are common media, although mark-sense cards may be used by some sources if you don't have access to punch equipment. Check with the programming service for format specifications. In some cases, you may need to write your program in hexadecimal. Each hex digit represents four data bits, so two hex digits will represent the data at each EPROM address. A conversion table is shown in Table 3, and Table 4 shows the hexadecimal equivalent of the program from Table 2.

Check your program before submitting it! One bit with the wrong value will throw the entire sequence off. An EPROM pre-programmed with variations of both circle and line displays is


FIG. 4-SCHEMATIC DIAGRAM for the state-of-the-art "do-nothing" box. The line and circle displays share LED6.
available (see parts list).
Figure 4 shows the schematic and Fig. 5 the 2716/2758 pinout. Inverters IC3-a and IC3-b form a clock oscillator A NAND gate, IC2-c, gates the clock pulses to counter IC4. As IC4 counts, its binary outputs address each memory location of EPROM IC5 in sequence. The seven highest data-outputs of IC5 control LED's. Six pairs of LED's are controlled by inverters (TTL for high current output) and LED6 is controlled with a transistor (using a seventh inverter would waste five inverters of another IC). The 270 -ohm resistors limit LED current to about 20 mA . The D $\emptyset$ IC5 output controls Q2 and Q3 through inverters. A logic-1 from $D \emptyset$ will cause Q3 to conduct and the LED's in the circle will light. A logic $\emptyset$ will light the line LED's using Q2.

Although an on-off switch isn't necessary, I decided to use latches to get three operating modes: RUN, one-cycle, and reset. Pushing the ONE-CYCLE switch sets latch IC1-a, allows clock pulses through IC2-c, and switches Q1 on through IC2-a.

| 6 1 2 3 | Øøø Ф申Ф1 0010 $\emptyset 011$ | $\begin{aligned} & 8 \\ & 9 \\ & \text { A } \\ & \text { B } \end{aligned}$ | $\begin{aligned} & 1 \phi \phi \varnothing \\ & 1 \phi \phi 1 \\ & 1 \phi 1 \phi \\ & 1 \phi 11 \end{aligned}$ |
| :---: | :---: | :---: | :---: |
| 4 | ¢100 | C | 1106 |
| 5 | 0101 | D | 1101 |
| 6 | $\emptyset 110$ | E | 1110 |
| 7 | 0111 | F | 1111 |


| TABLE 4 |  |  |
| :--- | :--- | :---: |
| 81 | $\emptyset 2$ |  |
| 41 |  |  |
| 21 | $\emptyset 6$ |  |
| 11 |  |  |
| $\emptyset 9$ |  |  |
| $\emptyset E$ |  |  |
| $\emptyset 5$ |  |  |
| $\emptyset E$ | 1E |  |
| $\emptyset 3$ | $7 E$ |  |
| $\emptyset 4$ |  |  |

Transistor Q1 controls all display current and shuts off the display when the RESET pushbutton is depressed. When counter IC4 reaches a count of 512, all addresses of IC5 will have been displayed and pin 14 of IC4 will go high. That resets latch IC1-a and


FIG. 5-PINOUT for the EPROM. Either a 2716 or a 2758 may be used.

IC4 through IC3-d and IC2-d. If the RUN switch is depressed, IC1-b will latch and the display will operate as before. The logic-1 appears at pin 14 of IC4 and resets IC4 to $\emptyset$, but since it does not reset latch IC1-b, the cycling repeats.
continued on page 87

## इull sinl Unicorn

A LOT OF THOUGHT WAS GIVEN TO HOW Unicorn-1 could be controlled remotely. A number of schemes were consideredultrasonics (not reliable enough and not enough range); infra-red (the same, but more so), and, of course, radio. A system was even devised using model-airplane $\mathrm{R} / \mathrm{C}$ equipment, but that proved to be expensive and not easily expandable to computer-control.
The system finally chosen was inspired by one used by amateur radio operators for VHF and UHF repeater control and its principles are probably familiar to most Radio-Electronics readers from at least one other source-the telephone company.
Before getting into the actual construction of the robot's R/C system, it might be a good idea to fill you in on this scheme, so you have an idea of the direction we're headed in.
The heart of the system is the DTMF (Dual Tone Multi-Frequency) systemalso known as Touch-Tone. A 16-key pad (shown in Fig. 65)-or a matrix of


FIG. 65-A 16-KEY Touch-Tone pad similar to this one was used in the prototype to modulate an FM transmitter.
switches providing the equivalent func-tion-is used to instruct a DTMF generator IC, in this case a ICM7206JPE, to produce a pair of tones unique to the key pressed.
That tone-pair modulates an inexpensive, low-power FM transmitter operat-
ing in the FM-broadcast band. The signal is received by a standard FM broadcastband receiver located in the robot and the tone-pairs are decoded to generate a one-out-of-sixteen control signal. That control signal is fed to a relay-driver board to energize the coils of the appropriate relays (as described in Part 7 of this series) and operate the robot's motors and solenoids.
This method will lend itself particularly well to computer control. The 16-key pad is arranged as a 4 -row by 4 -column switch matrix where each row generates its own tone, as does each column. The result, if the rows and columns are considered together (lined up in one row) is the equivalent of a computer byte-the standard 8 -bit word.

A computer can output, through a parallel port, an eight-bit binary number that can represent those same switch closures. That byte can be used in place of the keypad to cause the tones to be generated, thus allowing a computer program to direct the robot's actions.

Alternatively, if the robot carries an on-board computer, the output of its parallel port can easily be translated into control signals for the relays.
Several installments will be required to describe the control system in detail. This one will talk about FM transmitters and the relay-driver board. The next will talk about the Touch-Tone encoding and decoding circuits, and their interfacing to the others.

Finally, we'll talk about computer interfacing and a little about programming as it pertains to robot control.

## FM transmitter

This transmitter can actually be used for two purposes, although not simultaneously. In essence, it is what's commonly called an FM wireless mike. Usually it is used to transmit voices or music on an unused frequency of the FM broadcast band for personal entertainment purposes.

In that mode, using the robot's built-in amplifier and speaker (see Part 5), the robot can talk to persons in its vicinitywith a little help from the operator. In

ROBOT
Part 8-Last month we began to look at a remote-control system for the Unicorn-1 robot. In this part we will continue with that system by describing our control scheme, a simple FM transmitter, and a relay-driver board.

## JAMES A. GUPTON, JR.

fact, if the robot carries a second wireless mike, operating on a different frequency, a two-way conversation can be carried out.

However, that is secondary to our main purpose-actually controlling the robot. (Come to think of it, though, the control tones could also be fed to the robot's amp and speaker, making him sound a little like good old R2-D2.)

A schematic for a wireless mike is shown in Fig. 66. No foil pattern is given, since the circuit can be easily constructed on perforated construction board. Suitable FM transmitters are also available from a number of companies who advertise in Radio-Electronics.

## Transmitter construction

If you build your own transmitter, it can be constructed on a piece of perfo-

## PARTS LIST-FM TRANSMITTER

All resistors $1 / 4$ watt, 5\%
R1, R2-1 megohm
R3, R6, R9-8200 ohms
R4- 330 ohms
R5-470,000 ohms
R8, R11- 15,000 ohms
R12-3900 ohms
R13- 220 ohms
R14 (optional)-390,000 ohms
Capacitors
C1, C3, C4, C6-5 $\mu \mathrm{F}$, tantalum
$\mathrm{C} 2-0.1 \mu \mathrm{~F}$, ceramic disc
C5- $10 \mu \mathrm{~F}$, tantalum
C7, C11-0.01 $\mu$ F, ceramic disc
C9-5-15 pF, variable (E.F. Johnson 274 -0035-005 or equivalent)
$\mathrm{C} 10-7 \mathrm{pF}$ (approx.), ceramic disc
L1-see text
L2-see text
Miscellaneous: construction board, highimpedance microphone, solder, wire, etc.
ic, and most of the ones available as kits, are intended to be modulated by a highimpedance microphone. (If you intend to use a crystal mike, be sure to include resistor R14.)

If you are going to use the transmitter only with the Touch-Tone pad for control purposes, the first two stages-Q1 and Q2 - can be omitted, and the output of the tone-generator IC applied to the base of Q3, since its output level is much higher than that of a microphone, and not as much amplification is needed. In fact, you probably will have to add several hundred kilohms of resistance to attenuate the tones so they do not overdrive the transmitter and cause distortion.

Best results with the homebrew transmitter were obtained when tantalum capacitors were used where values of five and ten $\mu \mathrm{F}$ were needed. The tuning capacitor, C9, should have a value such that, when it is parallelled with C 10 , the


FIG. 66-SCHEMATIC DIAGRAM for the FM transmitter. Value of dropping resistor R14 may range from several hundred kilohms to two megohms or more.


FIG. 67-PROTOTYPE TRANSMITTER built on a small piece of board. Any construction technique may be used.
rated construction board or on a prototyping board. A prototype transmitter, built on a piece of board about $1 \times 4$ inches, is shown in Fig. 67. Wire-wrap or point-to-point wiring techniques can be used. Keep the leads as short as possi-ble-lead length begins to get critical at these frequencies $(80-108 \mathrm{MHz})$.

The transmitter shown in the schemat-
total capacitance does not exceed 22 pF . A good place to look for something to use as C 9 is in a junked portable FM radio.

Coil L1 is made using eight turns of No. 16 copper wire. Its outside diameter is $1 / 4$-inch and the total length of the coil is 0.6 inches. Coil L2 consists of 12 turns of No. 30 wire (wire-wrap wire will do nicely) closely wound around a quarterwatt resistor of the highest value you have on hand (it should be at least 100 K ). The ends of that coil can be soldered to the resistor leads, which, of course, then become the leads of the coil.

The antenna lead is soldered to the third turn of L1, counting from the 12 volt end of the coil. The antenna itself can be either a fancy telescoping type, or simply a piece of stiff wire. Since we are deliberately not trying to obtain maxi-
mum efficiency, the length of the antenna is not critical-about ten inches seems to work well.

Locate the antenna right at the transmitter, which can be mounted inside the command console if you like. It is not necessary to feed the antenna with coaxial cable-it can be connected directly to the output of the transmitter. What is important, though, is that the antenna be insulated from the case containing the transmitter, if that case is metallic, to prevent it from shorting out to ground.

The frequency of the transmitter can be affected by the antenna. It should be as rigid as possible and, more important, because of capacitance effects, it should be as far away from possible contact with your body as possible. Keep that in mind when you are tuning the transmitter, especially if the transmitter and antenna are mounted on the case containing the keypad and tone encoder.

The best section of the FM band for your use is probably the bottom-around 88 MHz . Tune your receiver to a clear spot in that area and turn up the volume so you can hear some background hiss. Then, using an insulated-or plasticscrewdriver, slowly adjust C9, or its equivalent, if you assembled a kit, until the hiss is blanked out. That will indicate that you are receiving your transmitter's carrier. Be patient-the tuning process is critical. It may also be necessary for you to stretch or compress L1 slightly to get into the right portion of the band.Before you fire up the transmitter, you should be aware of the FCC regulations governing the use of such devices. Those regulations may be summarized as follows:

- The use of such devices for personal surveillance is illegal!
- The range of such devices is limited to 100 feet. Do not attempt to extend that range through the use of higher power or more efficient antenna systems-use only what you need! Improve your receiver, if necessary.
- Do not attempt to use the transmitter below 88 MHz or above 108 MHz . The former may interfere with commercial two-way radio services; the latter with aircraft communications. Do not use the transmitter anywhere near commercial airlanes!
To be safe, make sure the signal begins to fade out about 90 feet from the transmitter. If it is too strong at that range, shorten the antenna or reduce the input power. That will not only keep you out of trouble, but will ensure that you can clearly observe-and control-the robot's actions before it does something to embarrass you.*

In the next part of this series we'll go into detail on connecting the tone genera-

[^1]

FIG. 68-SIMPLIFIED SCHEMATIC of the relay-driver board. Only one section is shown as all others are the same.


FIG. 70-PARTS PLACEMENT diagram for the relay-driver board. If 2 N 2222 transistors are not handy, almost any other type may be used.
tor to the transmitter, and the tone decoder to the receiver (and to the robot).

## Relay-driver board

A portion of the relay-driver board circuit is shown in Fig. 68. There is really very little more to it than that-the same circuit, for all intents and purposes, is repeated 32 times.

A foil pattern for the relay-driver board is shown in Fig. 69, and the partsplacement diagram in Fig. 70. Don't be put off by its complexity, though. Initially, we'll use only half of that boardeach relay will have its own output from the ULN-2813A driver IC and its own transistor. The balance of the board is reserved for future use-primarily when it becomes necessary for the robot, when it gets its on-board computer, to respond to stimuli from its environment. It can also serve to provide other control functions if a more sophisticated control system is used, and suggestions for that will appear in a future part of this series.

Jacks J1 and J2 will be used to connect the relay-driver board to the decoder board.

Here's how the circuit works: The ULN-2813A is an inverting octal driver. What that means is that it has eight identical sections; and when a logic-high signal (about five volts) from the decoder board is applied to the input of one of the sections, the output of that section goes to a logic-low state (zero volts, or ground) and will act as a ground for any voltage that is applied to it.

When an output of the IC goes "low," it causes its associated transistor to be saturated. That allows five volts to pass from the collector and out the emitter to the coil of the relay assigned to that transistor, causing the relay contacts to close. That's all there is to it.

Although IC pin and function assignments are arbitrary, Table 1 shows a suggested arrangement for use with a 16-key Touch-Tone keypad.

Because we are limited to 16 on/off control signals, several of the robot's original functions temporarily have had to be eliminated or combined. For example, we can no longer beep the horn, and both end-effectors now operate simultaneously.

Regaining those lost functions will be easy under computer control and later we'll present a couple of ideas for some simple logic circuits that will allow the 16 radio-control channels to provide more than 16 functions.

Construction of the board is straightforward. The 2 N 2222 transistors were used because they were handy. As Fig. 71 shows, almost any transistor can be used-you can see four different types there. If you have PNP-say, 2N2907instead of NPN transistors, the only change that has to be made is to insert the transistors in the board backwards-the emitter goes where the collector would


FIG. 69-FOIL PATTERN for the relay-driver board. Only half the board will be used at this time, with the rest reserved for expansion.

| TABLE 1 |  |  |
| :---: | :---: | :---: |
| IC No., Pin No. | Key No. | Function |
|  | $\begin{aligned} & 1 \\ & 2 \\ & 3 \\ & 3 \\ & \text { A } \\ & 4 \\ & 5 \\ & 6 \\ & \text { B } \\ & 7 \\ & 8 \\ & 9 \\ & \text { C } \\ & \hline 0 \\ & \hline \\ & \hline \end{aligned}$ | Both wheels, forward <br> Both wheels, reverse <br> Left wheel, forward <br> Left wheel, reverse <br> Right wheel, forward <br> Left \& right arm solenoids <br> Body rotate, right <br> Body rotate, left <br> Left shoulder, up <br> Left shoulder, down <br> Right shoulder, up <br> Right shoulder, down <br> Left arm, up <br> Left arm, down <br> Right arm, up <br> Right arm, down |


| TABLE 2 |  |  |
| :---: | :--- | :--- |
| Transistor | Relay No. | Finger No. |
|  |  |  |
| Q1 | RY3, RY5 | R, 6 |
| Q2 | RY4, RY6 | M, 3 |
| Q3 | RY3 | R |
| Q4 | RY4 | M |
| Q5 | RY5 | 6 |
| Q6 | RY19, | 18,4 |
|  | RY20 |  |
| Q7 | RY1 | 20 |
| Q8 | RY2 | 17 |
| Q17 | RY9 | P |
| Q18 | RY10 | L |
| Q19 | RY11 | 7 |
| Q20 | RY12 | 2 |
| Q21 | RY15 | N |
| Q22 | RY16 | 8 |
| Q23 | RY17 | 5 |
| Q24 | RY18 | 1 |

be, and vice-versa. The resistor placement
Installation can stay as shown.

## PARTS LIST-RELAY-DRIVER BOARD

All resistors $1 / 4$ watt, $5 \%$
R1-R32-470 ohms
R33-R64-220 ohms
Semiconductors
IC1, IC2-ULN-2813A inverting octal driver (Sprague)
Q1-Q32-2N2222 or equivalent NPN-type: 2N2907 or equivalent PNP-type (see text for details)
J1, J2-20-pin, double row, header con nector (AP Products AP923862-R or equivalent)
Miscellaneous: PC board, 22/44-finger prototyping board (Radio Shack 276-154 or equivalent), two $22 / 44$-pin sockets (Radio Shack 276-1551 or equivalent), 18 -pin IC sockets, mating connectors for J 1 and J2, 20-conductor ribbon cable, hardware, wire, etc
A PC board for the relay driver board is available from PPG Electronics Co., Inc, 14663 Lanark St., Van Nuys, CA 91402. (213) 988-3525. Price is $\$ 9.95$ plus $\$ 1.00$ for shipping and handling. CA residents add 6\% tax. MC and Visa accepted.
completed, it is piggy-backed onto a $22 /$ 44 -finger board like the one that was used for the relay board, using $1 / 4$-inch spacers. Make positively sure that the two boards are electrically isolated from each other.

The emitter (output) of each 2N2222 transistor is connected to a finger on the piggy-back board. You can use wire-wrap wire for that. The most straightforward way is to use the same finger number (or letter) as that which is connected to pin 16 of the appropriate relay on the relay board (See Table 2)


FIG. 71-THE RELAY-DRIVER BOARD is mounted parallel to the relay board, seen reflected in the mirror.

A $22 / 44-$ pin edge connector is mounted parallel to the one for the relay board (refer to Fig. 71) and, assuming that you have followed the wiring scheme described above, connections are made between like-numbered pins on the driv-er-board socket and the relay-board sock-
continued on page 82

Here's how the picture and sound are transferred from the disc to your TV set in the new laser videodisc system.

MUCH HAS BEEN WRITTEN ABOUT THE makeup of the three videodisc systems currently vying for acceptance in the marketplace. There are laser optical discs favored by Magnavox, Pioneer, and others, capacitance-pickup grooveless discs, proposed by JVC, Matsushita, and General Electric, and capaci-tance-pickup groove-type discs developed by RCA. In the case of the optical and capacitance videodisc systems so far proposed, we have seen many diagrams of how tiny "pits" or dark and light spots in the disc tracks will be carrying the encoded information. But little has appeared in the literature as to just what that information is, how it is encoded, and what the circuitry needed to decode it will be like, once you get past the question of pickup format.

The Magnavox Consumer Electronics Company has prepared an excellent booklet about their model VH-8000 Videodisc Player in order to familiarize technicians and others with their laser disc system as well as with playback circuitry. While the booklet deals with all aspects of the optical-laser videodisc player's operation, our concern here will be only with the signal-processing circuits that handle the signals picked up by arrays of photo-sensitive diodes that catch the laser-beam reflections from the surface of the spinning disc.

To understand the circuit descriptions that follow, we must know just what sort of signals are contained in the


LEN FELDMAN
laser-disc tracks themselves. The intelligence that is encoded on the videodisc is the combination of three different FM signals:

1. An 8.1-MHz FM signal modulated with composite video, including chroma.
2. A $2.3-\mathrm{MHz}$ FM signal modulated with Channel-I sound.
3. A $2.8-\mathrm{MHz}$ FM signal modulated with Channel-II sound.

Figure 1 shows those three signals in the frequency spectrum. Each sound carrier (the videodise contains two independent sound channels and thus has the capability, among other things, for stereo sound) has a maximum FM deviation of $\pm 100 \mathrm{kHz}$. The $8.1-\mathrm{MHz}$ video carrier has a deviation of 1.7 MHz (from sync tip to peak white) with its bandpass extending from below 4 MHz to above 12 MHz to include all sidebands. Each of the sound signals modulates the 8.1MHz video FM to create the resultant signal that becomes encoded on the videodisc, as shown in Fig. 2.

## Signal processing

A simplified block diagram of the sig-nal-processing circuitry needed to produce the video and audio signals from the laser disc is shown in Fig. 3. The total composite FM signal from the photosensitive diodes and preamplifier module is applied to a high frequency amplifier/splitter. Frequency tuned networks separate the sound-FM signals from the video FM. The sound FM is applied to two frequency-sensitive


FIG. 1-AUDIO AND VIDEO signals as recorded on videodisc. One video sideband-note sync pulses and video ramp wareform-is shown at 8.1 MHz.

b
SOUND FM SIGNAL


FIG. 2-RESULTANT SIGNAL created when both of the sound signals modulate the 8.1 MHz video-FM.
stages: a $2.8-\mathrm{MHz}$ sound demodulator and a $2.3-\mathrm{MHz}$ sound demodulator. Those stages serve as ordinary FM detectors and retrieve the audio signals from their respective FM carriers. The two resulting audio signals are applied to an electronic-switch network that applies either one or both of them to the RF modulator of the player. Front-panel switches on the videodisc player determine which audio signals are heard and used.

The $8.1-\mathrm{MHz}$ video-FM signal is applied to Video Demodulator I that extracts
the composite video signal from the carrier. That composite signal is amplified by a video amplifier and applied to the RF modulator. Since each frame of the 54,000 frames contained on a single side of the disc is numbered, and the player is capable of displaying that number on the TV screen, the picture-number information is stripped from the composite video signal by the clipper/decoder circuit. Here, the picture-number information is decoded and converted to a pic-ture-number video signal. That signal is also amplified by the video amplifier and


FIG. 3-BLOCK DIAGRAM of the circuitry used to derive audio and video from the videodisc.


FIG. 4-DETAILED block diagram of the splitter and high-frequency signal processor. The color separator removes the color burst from the video.
applied separately to the RF modulator. The RF modulator places the audio and video onto the required RF frequencies for TV Channel 3, or Channel 4, selectable by the user. The output from the RF modulator is connected to antenna
terminals on the viewer's TV set through an antenna-switch box.

The video circuitry creates a varying DC voltage that is proportional to the burst amplitude. That voltage is called the emphasis-control voltage and is ap-


FIG. 5-SOUND SIGNAL processing modules. The sound-FM signals are processed by these two
circuits.


FIG. 6-DROPOUT DETECTOR activates signal restoration circuitry when it detects missing information.
plied to a high-frequency amplifier to emphasize high frequencies when operating near the inner diameter of the videodisc (where such high frequencies would tend to be more attenuated). The control is necessary because the "pits" on the surface of the disc are more closely spaced at the inner diameter of the disc.

The player is also designed to compensate for minor dropouts of informa-


FIG. 7-VIDEO-FM and delayed video-FM are supplied to the demodulator circuit as shown in this block diagram.
tion from the videodisc. A dropout, in that case, is defined as an area on the videodisc that has incorrect encoding or no encoding at all. Such a loss of encoding could be caused by physical damage to the disc after it has been used, or by manufacturing imperfections. The drop-out-correction circuitry built into the Magnavox player can compensate for the loss of up to one complete horizontal line on the TV screen.

The $8.1-\mathrm{MHz}$ video-FM signal is applied to a dropout-detector circuit. If a bad spot on the videodisc is encountered, the $8.1-\mathrm{MHz}$ signal will be absent and the dropout detector will sense that absence.

The $8.1-\mathrm{MHz}$ signal is also applied through a 64 -microsecond delay line, to Video Demodulator II. Since 64 microseconds corresponds to the sweep time of a single horizontal line, when a dropout is encountered, the dropout detector activates an electronic switch that then applies the delayed, previous horizontal line in place of the one that has dropped out. The result on the video screen of the viewer's TV set is two horizontal scan lines with the same video information. In other words, the dropout has been filled in by repeating the previous line of video information.

## A more detailed look

A more detailed block diagram of the high-frequency signal processor and splitter is shown in Fig. 4. The entire composite FM signal is applied to an RF amplifier whose gain control, R3002, is used to set the correct output level. The output of the RF amplifier is applied to the sound FM amplifier. The input of the sound FM amplifier is tuned to pass only the $2.3-\mathrm{MHz}$ and $2.8-\mathrm{MHz}$ sound carriers, both of which are then present at the output of the amplifier.

The color-separator circuit shown in Fig. 4 removes the color burst from the video signal by gating the burst keyer with the horizontal burst-gate pulse. The amplitude detector creates a DC voltage proportional to the burst amplitude. As that DC voltage decreases, the high-frequency response of the video amplifier increases. The net effect is to boost the high frequencies towards the inner section of the videodisc. The video FM amplifier boosts the $8.1-\mathrm{MHz}$ video-FM signal. The 2.8 MHz trap removes any remaining Channel-II audio carrier at that point. Any remaining Channel-I ( $2.3-\mathrm{MHz}$ ) sound carrier is trapped out later in the signal path.

## Sound signal processing

The sound-FM signals are processed by two circuit modules, whose block diagrams are shown in Fig. 5. The sound-FM signal is applied to the 2.3MHz FM demodulator on the Sound continued on page 83

## When analyzing simple circuits, be careful not to overlook the obvious. EARL "DOC" SAVAGE, K4SDS, HOBBY EDITOR

WELL, THE ELECTRONS HAVE SETTLED from Light Mystery No. 2 (see Hobby Corner, August 1980 issue). And those electrons were flying all over the place from California to Connecticut and everywhere in between.

Let's take another look at Don Francois' puzzle. He constructed the two circuits shown in Fig. 1 and found that the bulb brightness in one was greater than in the other.


Things got tough when he measured the same voltage across each bulb. And, to make matters worse, Don measured the same current in each circuit.

That puzzle can be attacked by breaking it down into three questions:

1. How can the voltages and currents be the same in the circuits shown in Figs. 1-a and 1-b?
2. Are the bulbs brighter in either circuit?
3. Why is the brightness different?

The answer to No. 1, of course, is that the voltages and currents cannot be equal if the brightenesses are different. All the responses that addressed that question were correct though in some cases the reasoning was fuzzy.

The voltage waveforms in the two circuits are shown in Fig. 2. The fallacy of

the "equal" measurements lies in the fact that common AC meters are designed to measure sinewaves (Fig. 2-a). If the waveform is not sinusoidal, the meter gives an inaccurate reading. A DC meter also gives a false reading on pulsating DC (Fig. 2-b).

We must assume that Don used a typical multimeter. If so, his readings in the diode circuit would be wrong whether he measured AC or DC. Actually, as Fig. 2 shows, the voltages and currents are not the same and therefore, the power (and brightness) differs.

Getting on to the question of which is brighter-That is where I got a surprise. Two readers "proved" that the bulbs in both circuits are equally bright in spite of Don's report to the contrary. And, almost half of the responses were wrong!

One sure way to find the answer is to hook up the circuits and observe the difference in brightness. Quite obviously many of you did not do that.

In fact, it appears that none of you who wrote tried the circuits. Come on, guys and gals; trying it out (experimenting) is the only way you can be sure that your reasoning is correct. Logic and computation are fine if you have all the facts and don't overlook something. Even some of the right answers were based on incomplete facts.

Sometimes things are not as simple as they first appear. Everyone remembered that the diode "cuts out" half of the AC sinewave. But did you remember that the two bulbs in series effectively divide that 117 volts AC between them? Did you remember to take into account that the circuit resistances of both circuits are not the same? What else did you overlook? The way to tackle any electronics problem is to use a combination of reasoning and experimentation.

Those of you who answered the question correctly used a great variety of logical and/or mathematical proofs. Three sharp readers threw in some calculus (that almost threw me-it has been a long time). Let me hasten to say that calculus is not necessary to prove the answer. However you did it, congratulations to you-unless, of course, it was an accident.

I have given a lot of thought to making this report on Light Mystery No. 2. Hav-
ing stirred up the dust with what has been said so far, I'm going to stop short of giving you the answer. You may think that's "dirty pool" but, actually, the answers and the facts are above. Have fun!

A very special thanks to Don Francois for sharing that mystery with us. If you have an interesting electronic puzzle, send it along. Perhaps you too can make the electrons fly.

## Speedometer

Henry Milowski of Welland, Ontario sent me a very interesting letter. It seems that the speedometer on his exercise bicycle broke and he saved some bucks by making a replacement. The basis of his design can be used to measure the rate of any rotating object.

Henry used an old DC motor as a generator that is turned by the bike tire (Fig. 3). His motor came from a discarded tape deck. The output voltage is proportional to the speed and that output is measured by the meter (a miles-per-hour scale was added). The LED protects the meter if the tire turns backwards and it also adds to the appearance.


You may need to change the value of the resistor depending upon the meter and motor you use. Not only will the motor be different, but the size of the shaft "wheel" pressing on the tire will determine the output range. I suggest that you start with a 10 K or 20 K pot in the circuit.
Thanks Henry for sharing your project with us.

## Fundy

Noting Henry's Canadian address, reminds me of my summer trip around the Bay of Fundy. I guess I saw about as much as one can see there in a couple of weeks. I was duly amazed. Those 20 - and 30 -foot tides are all but unbelievable-
continued on page 72

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## HOBBY CORNER

continued from page 70
even when you are standing there looking at them. And while you're watching that awesome amount of water flow up and back every few hours, you cannot help but think how much electrical power it could generate.

You stand, watch, and dream just as countless others have done before you. Some day that potential will be harnessed. That will be something to see! (Any new ideas?)

## Another clock

If there is a piece of digital equipment that has universal appeal and application, it is the clock.

Larry Neel of Cincinnati, OH has raised a question that brings forth some interesting thoughts. Larry wants a clock that normally displays local time but could give Coordinated Universal Time (GMT for you old-timers) at the push of a switch. I certainly agree that it would be better than using two clocks for hams, SWL's, and others.

I haven't had a chance to work on this problem yet but it doesn't appear to be very difficult. Why not a switchable circuit between the clock IC and the display? That circuit would add five, six, or whatever hours needed to the time before it is displayed. Or perhaps it would be easier to switch between two continuously running clock IC's.

Have any of you built a clock like that? If so, send me the info and I'll pass it along.

## Reader question

E. M. Shanley of Novato, CA is looking for a control circuit to operate a piece of apparatus by radio control. The radio part is not causing him a problem-it's what comes after the receiver.

What E.M. needs is a circuit to close and latch a relay on the first pulse, and then open (unlatch) the same relay on the second pulse.

Can anyone offer any help?

## Flash anyone?

Frank Eatherton of Alameda, CA is a photographer and is looking for a reliable circuit for a wireless "slave" flash control. Frank says that he is not only tired of tripping over wires, but that it can be costly when he does so and topples over a flash or camera!

What is needed is a small unit to actuate additional flash units when the main one fires. Such a unit could operate by radio or light-control.

If you have developed a flash-control circuit, or any type of wireless control that could be adapted for flash use, send along a schematic with description of the device. Perhaps we can save Frank and other readers from damaging any more camera equipment.

# UNIVOTT'S DT810 dicitil multimeter 

## The unique space age digital multimeter with transistor gain (hFE) measurement capability should be the only multimeter you own.

Ora Electronics has offered in the past many fine Digital Multimeters (D.M.M.'S). We still sell the famous D.M.M.'S such as Beckman, Fluke, Hickok, and others. We have always followed the advance in technology used in D.M.M.'S, and we always wanted to supply our many good customers with the most Ideal Multimeter, at a price they can afford. In the past we had to sell good, but expensive Multimeters, expensive but "fair" Multimeters, and plain "cheap" Multimeters.

## WE FOUND IT!

Several months ago, a famous Test Equipment Manufacturer, walked in to our headquarters with a Prototype of a Digital Multimeter. We were very impressed it had almost everything we wanted plus a bonus, the only question remaining was "how expensive is it?" When we heard the answer, a big smile appeared on our faces. After several improvements we are proud to offer it. After you read the features (and price)! am sure you are going to order one or more, of these fine D.M.M.'S that we call the "UniVolt".

## LCD DISPLAY.

The unit has a 3.5 Digit liquid crystal display. The sharp digits are 14 mm high and have a viewing angle of $140^{\circ}$.

## HIGH ACCURACY.

The basic D.C. accuracy of the UniVolt is $0.5 \%$ of reading +1 digit, which makes it one of the more accurate instruments in its class. The input impedance is very high, 10 mega- Ohms $(10,000,000)$ Ohms, which helps in measurements of low voltage and high frequency signals.

## MEASUREMENT RANGES.

The UniVolt has D.C. voltage range of 100 uv to 1000 V in five steps, A.C. voltage range of 100 mV to 1000 V , current measurement range of 100 mA to $10 \mathrm{~A}(\mathrm{DC})$ and resistance range of 1 to $2,000,000$ Ohms.

## CONTINUITY \& DIODE TEST.

A fast and accurate continuity test mode utilizes a built-in buzzer to indicate continuity. The same mode is used to check diodes and their approximate forward voltage.

## EASE OF OPERATION.

The UniVolt is small, it measures $61 / 2^{\prime \prime} \times$
 including battery! It utilizes push buttons, for easy one-hand operation and the front panel has a unique color coding for reduced errors.


## OVERLOAD PROTECTION

The unit has an extensive overload protection on all ranges. On D.C. current ranges it uses a .5A GMA type fuse. A spare fuse is supplied with the unit at no extra cost.

## MAINTENANCE FREE

The heart of the UniVolt Multimeter is a 40 pin L.S.I. chip; the Intersil ICL710G. This space ages chip has proven to be one of the most sophisticated and reliable micro-electronic circuit in use, it is supported by minimum amount of external parts, which are over specified to insure failure safe instrument. Of course, Ora Electronics stands by this instrument and guarantees it for one year (See specific warranty information).
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## OTHER FEATURES

It uses one 9 volt carbon battery (included), which last approximately 200 hours of continuous use. Its sampling time is 0.4 seconds, operating temperatures of $30^{\circ} \mathrm{F}$ to $104^{\circ} \mathrm{F}$, and operating humidity of less than $80 \%$ R.H.

## BONUS!!

We left the best to the end. The UniVolt DT810 has something unique. It has a transistor gain ( hFE ) measurement mode! This unique feature enables you to measure hFE values of 0-1000 of either P.N.P. or N.P.N. transitors.

## SPECIAL PRICE

We had originally decided to sell the unit for $\$ 119.95$, but in order to promote the new advancement in D.M.M. design, represented by the UniVolt, for a limited time only you can buy this incredible unit for only $\$ 99.95$ including: standard red \& black test leads, a fresh 9 v carbon battery, a spare 0.5A GMA type fuse and an instruction manual.


FREE CASE
We have worked long on the UniVolt project and we hate to see scratches or bad looking units. So we decided to go all the way, when you buy the UniVolt DT-810 Multimeter (and for a limited time only!) we will give you absolutely free a hard vinyl leatherette, carrying case, with felt padding and a compartment for your test leads. The regular selling price for this case mode CC-01 is $\$ 8.00$.

## ACCESSORIES AVAILABLE.

The only two accessories available are: UP11, hFE probe with special plug and 3 color codes alligator clip, and the UP-12 I.C. clip adaptor, which will help you hook your multimeter to any I.C. pins. (You can buy both probes for only $\$ 6.00$, but only when you purchase the UniVolt DT-810 now.)

## ORDER NOW!

It's very easy to order your UniVolt DT-810 multimeter. Send $\$ 99.95$ (California residents add $6 \%$ sales tax) plus $\$ 2.50$ delivery charge to the address below, if you want the optional accessories, please add $\$ 6.00$ (California residents add $6 \%$ sales tax). A cashier check or money order will help speed your order. Credit card holders (master card or visa) can call our toll free number (800) 423-5336, in California it's (800) 382-3663. C.O.D. orders will be accepted, but you must pay by cash or money order and a C.O.D. charge of $\$ 1.40$ will be added. If you decided to buy another brand of Multimeter, please call us too, we carry many other types of multimeters and test equipment at low prices.


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## new poroduots

## More information on new products is available. Use the Free Information Card inside the back cover.

ELECTRONIC DESIGN LAB, model CA-16, features a variable-function generator: frequency 1 Hz to 100 kHz variable rate, 2 -volts output level, sinewave distortion less than $1 \%$, squarewave TTL and CMOS logic levels, and triangle and sinewave for linear operation. Power supplies include 6 regulated voltages, current-limited and short-circuit proof. Noise: 20 mV under load; There is a +5 -volt supply -1 -amp maximum load and $\mathrm{a}+12,+15,-5,-12,-15,750 \mathrm{~mA}$ maximum load.


CIRCLE 151 ON FREE INFORMATION CARD
Also featured are four LED logic-level indicators for TTL and CMOS logic, four (HI and LO) debounce toggle switches for TTL and CMOS logic, and two CMOS-to-TTL logic-level converters. Everything is at the user's fingertips and the components can be used over and over again. The model CA-16 has been designed for engineers and technicians for testing new prototypes, modifications, circuit changes, and ideas, and is recommended as a teaching aid for students just starting in electronics. Price is $\$ 99.95$.-Cascade Labs, 4156 South Alder Ave., Freeland, WA 98249.

PROGRAMMABLE ELECTRONIC THERMOSTAT, the Comfort Zone, is the only electronic thermostat on today's market that offers triple setback, or the selection of three temperature periods in a 24 -hour sequence. In addition, each day in a seven-day sequence can be programmed separately. A full week's program can be entered at one time and will repeat indefinitely until changed.


## CIRCLE 152 ON FREE INFORMATION CARD

Comfort Zone has been designed for easy operation: When programming, the user rotates easy-to-use thumbwheel until the desired time
and temperature settings appear on display. Pushing a key enters the time and temperature settings into the unit's memory, and an audio tone tells the user that the information has been properly entered. The previous day's program can be repeated at the touch of a key. When not being programmed, the time, temperature, and the day of the week are on continuous display.

Comfort Zone is fully automatic, but can be used manually for special periods of heating or cooling. The user may also arrange for 10-, 20-, or 30 -minute bursts of heat or cold without disrupting the pre-selected program. The device is solid-state, and no additional wiring, transformers, or relays are necessary for installation. A battery protects the microcomputer's memory in the event of power failure. Comfort Zone's suggested list price is less than $\$ 200.00-\mathrm{PCI}$, Incorporated, 1145 Sonora Court, Sunnyvale CA, 94086

FUSE BLOCK, the Snaptrack fuse block series $6 B F$, features neon indicator lamps that light when a 3AG fuse blows. (The 3AG fuse is very


CIRCLE 153 ON FREE INFORMATION CARD
small and difficult to find when blown.) The 6BFseries fuse blocks are available in models carrying two or three fuses. Connections are made via screw terminals. The fuse clips, terminal blocks, and lamps are wave-soldered to epoxy-glass printed-circuit boards that snap-in mount in Snaptrack mounting channel; the modules mount easily there, along with relay sockets, edge connectors, power supplies, and other RDI control modules. Snaptrack requires only two fasteners per foot of track instead of two fasteners per fuse block. Rating is 10 amperes, 300 volts, with $120-$ volt neon lamp; the block handles 3AG fuse ( 1.25 $\times .25 \mathrm{in}$.) Price: three fuse modules, $\$ 4.35$ each in 100 lots.-RDI, 525 Randy Road, Carol Stream, IL, 60187.

WIDEBAND UHF/VHF AMPLIFIERS, are a line of low-cost, high-performance, prepacked hybrid IC wideband amplifiers for CATV, MATV, and similar applications that has been introduced by Amperex Electronic Corporation. They operate from a 12 -volt supply voltage and cover a 40 - to 860 MHz frequency range. The line is produced as thin-film circuits on ceramic substrates and consists of $1-, 2-$, and 3 -stage types, with gain ranging from 12 dB for the single-stage amplifier to 28 dB for the 3 -stage type.
The line includes five devices: ATF445/OM 345 has 12 dB gain and $99 \mathrm{~dB} \mu \mathrm{~V}$ output; ATF442/ OM350 has 18 dB gain and $100 \mathrm{~dB} \mu \mathrm{~V}$ output; ATF443/OM360 has 23 dB gain, $105 \mathrm{~dB} \mu \mathrm{~V}$ out-


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put; ATF444/OM361 has 28 dB gain, $105 \mathrm{~dB} \mu \mathrm{~V}$ output, and ATF446/OM370 has 28 dB gain, 113 $\mathrm{dB} \mu \mathrm{V}$ output. Higher gain is achieved by cascading two amplifiers. The hybrid UHF/VHF wideband amplifiers are priced as low as $\$ 3.05$ each (1000 qty) for the AFT445.-Amperex Electronic Corporation, Providence Turnpike/P.O. Box 98 , Slatersville, RI 02876.

VID-KADDY offers an attractive means of storing the cords and cables at the back of your VCR unit. It is a small, simple storage case that holds the remote pause control device and cable. All that will then be visible will be the compact black


## CIRCLE 155 ON FREE INFORMATION CARD

box, but the remote control can still be used easily at any time. Vid-Kaddy can be attached with foam tape to the back of the VCR unit in a matter of seconds. Price is $\$ 6.95$.-Video Specialties, P.O. Box 244, Fraser, MI 48026.

HOBBYIST'S ELECTRONIC COURSE, designed for individuals with little or no previous education in electronics, is arranged in seven easy-tounderstand units. The course includes the relationships among voltage, current, resistance and power, magnetism, DC and AC, transformers, inductance, reactance, and the proper use of a voltmeter, ammeter, and ohmeter. In addition, the fundamentals of rectifiers, filters, power supplies, and oscillators are covered, as well as many types of digital circuits and an explanation of how a digital computer works.


CIRCLE 50 ON FREE INFORMATION CARD
Finally, the course provides an overview of the major electronics hobbies-experimentation and construction, shortwave listening, amateur radio, R/C, personal computing, and high-fidelity audio. Twenty-six optional experiments will provide thorough hands-on experience with electronic devices. They will, however, require that the student purchase the Heathkit/Zenith model ET3100 Experimenter/Trainer Kit. The Hobbyist's Course is priced at $\$ 54.95$; the Trainer Kit is $\$ 74.95$.-Heathkit/Zenith Educational Systems, Dept. 350-470, Benton Harbor, MI 49022. R-E

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[^2]
## Harmonics? Here's a simple and inexpensive way to get rid of them. HERB FRIEDMAN, COMMUNICATIONS EDITOR

there are virtually no new frontiers left as far as the radio spectrum is concerned. It's no longer a matter of when devices would be introduced that could efficiently generate and amplify transmissions on the next higher slice of the spectrum but rather, it's where do we put everyone who wants to use the radio spectrum.

Until recently, no one really cared where spurious signals from transmitters were going. Early spark-gap telegraph operators actually depended on their signals taking up hundreds of kilohertz with sideband-splash, clicks, pops, and crackles. How else was anyone to know they were on the air? When the interference got too bad (because more than two stations were trying to operate within a halfmegahertz of each other) the old Department of Commerce simply forced the CW operators to change from an AC high-voltage supply to DC. That created a thunderous crash of silence in which many stations could operate at the same time.

As equipment improved, so did the rules tightening control of interference. But the harmonics of broadcasting stations still fell in the VHF marine band, and FM stations couldn't have cared less about their harmonics because simply no one was operating at 170 MHz or higher. Radio amateurs had even less of a problem because, until TV came along, their harmonics generally fell within their own bands.

But today there's someone operating on virtually every frequency, and the general rule is that if your station is causing interference to the reception of any other station because of transmitter characteristics such as harmonics or spurious signals, you either fix the problem or you
shut down. It is never the other station's problem. (Even amateur and CB transmitters now have strict limitations on harmonic and spurious output.)

Because of the need to squash all harmonics and spurious signals, many deepnotch filters that were formerly considered oddball, have now become commonplace in VHF and UHF transmitters. A good example of what to expect is the harmonic filter in Heathkit's model VF7401 two meter scanning transceiver.


FIG. 1
Figure 1 shows a schematic of this filter. Components L306, C326, C327, C328, L307, C334 form a conventional low-pass filter used for harmonic suppression. The problem is that the L-C filter doesn't provide adequate second harmonic suppression. To cascade standard L-C filters might cause more harm than good because RF tends to seek a free path around the coils and capacitors, and really loss-free L-C filters are fairly expensive.

One way to get an inexpensive, effective filter is to short all, or actually most, of the RF to ground through a tuned line. That is precisely what's done in the $V F$ 7401 by a $1 / 4$-wavelength transmissionline stub, shown in Fig. 1 just ahead of
antenna connector J301.
The stub is an electrical $1 / 4$-wavelength of shielded cable (coax) coiled within the transceiver. The center conductor is connected to the transmitter output at the junction of L307 and C334. The other end of the stub is shorted, with both the center conductor and shield connected to ground.

If you recall, $\mathrm{a}^{1 / 4}$-wavelength stub is an impedance inverter. If one end is shorted, the other end appears as a high impedance. So, at the operating frequency, the stub actually does not exist in the circuit because it appears as a high impedance in parallel with the 50 -ohm transmitter load (the antenna).

At the second harmonic frequency, however, the stub is an electrical $1 / 2$-wavelength. Half-wavelength stubs are impedance repeaters, often called matching transformers. At the operating frequency, what appears at one end also appears at the other. Thus if one end is a shortcircuit, the other end appears as a shortcircuit and the stub in Fig. 1 is actually a low-impedance path to ground for the second harmonic.
Just because we can coil up a stub made of coax cable within a small cabinet doesn't mean we must always use flexible cable. For higher-power transmitters, rigid transmission line can be used. A practical example of what to expect is the filter used by FM stations whose harmonics interfere with UHF communications. Imagine if you will, an FM station with an assigned frequency of 91.3 MHz . Its fifth harmonic falls at 456.5 MHz , a frequency in the UHF spectrum used by police and fire departments, as well as land transportation systems such as taxi services. Now the fifth harmonic might not sound like much, but if the FM station is running from 10 to 20 kW on its assigned frequency, and its antenna is about one mile from, and on a direct line with, the antenna of a taxi service, it can create all sorts of reception problems for taxis on the fringes of their service area. In an actual instance, the received fifth harmonic was $1.8 \mu \mathrm{~V}$ at the antenna terminals of a UHF receiver. It "broke" the squelch continually and just about destroyed radio communications for the taxis.

It might appear to be a contest between mismatched opponents, but remember the rule: If your gear creates interference it's your problem. The taxi service does continued on page 78

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## COMMUNICATIONS CORNER

 continued from page 76not shut down or change frequency (the FCC did not permit a change). But the FM station must eliminate the interference or shut down. The fact that the FCC might have made a mistake in assigning a UHF channel located close to the FM station, and directly on the fifth harmonic, has nothing whatsoever to do with it. The fifth harmonic goes or else!
This problem can be handled by a stub serving as a harmonic filter. Figure 2 shows how it's done. The transmission line is a 50.5 ohm 3 -inch rigid line that resembles a large pipe. The center conductor is copper tubing, approximately $5 / 3$-inch in diameter. A filter, of the same material and construction as the transmission line, is soldered to the line and its shield and the center conductor are shortcircuited at the end by a solid air-tight cap so that there are no breaks for the RF to escape. What we end up with is a grounded, fully shielded, stub. The effective attenuation is about 60 to 80 dB .
As the spectrum reaches saturation on a continuing basis, the need for, and use of, deep-notch filters will become more commonplace. Keep this in mind the next time you're working on a transmitter of any sort. The little piece of extra transmission line just might be the filter that allows the rig to stay on the air. R-E

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|  |  |  |
| (over 100.000 characters of storage on each onel) | Yes | Ye |
| Telephone Communications available connect to large |  |  |
|  |  |  |
| computers/electronic mail etc | c Yes | Yes |
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destroyers, etc! STAR WARS: Fly your space fighter into the Death Star to destroy it! But watch out,
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## radio procurts

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up on the antenna, giving an extra boost in performance. The antenna may even be mounted at ground-level-there's no need for tower or mast mounting to achieve excellence. The Sigma IV has an RF safety factor of 2,000 watts. Cost is \$89.95.-Avanti Research and Development, Inc., 340 Stewart Avenue, Addision, IL 60101.

TRANSCEIVER, OMNI-C Series, model 546, covers bands from 160 through 10 meters and has all nine high-frequency bands, with crystals included for seven of the nine bands. It features a threemode, two-range offset-tuning capability with a choice of offset tuning for the receiver, transmitter, or combined transceiver. Switching is provid-


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ed for selecting the standard 8 -pole 2.4 kHz SSB filter, the optional 1.8 kHz 8 -pole SSB filter, the optional $250-\mathrm{Hz}$ or $500-\mathrm{Hz} 8$-pole CW filters, cascading them for 16 poles of filtering, or placing them in the signal path along with 450 and 150 Hz active audio filters. The OMNI-C also has "hang" AGC and a standard noise blanker. Price is \$1189.-TEN-TEC, Inc., Highway 411 E., Sevierville, TN 37862.

TRANSMITTER KIT, model 1750 Meter Transmitter, for the 160 to 190 kHz experimenter's band. Operation at one-watt input power and with a 50 -foot maximum antenna length is allowed by the FCC with no license required. The transmitter has two parts: The main assembly contains the frequency generator, power supply, and control panel, and is located at the operating

position. The antenna-tuning assembly mounts at the base of the antenna. The transmitter is for CW operation but can easily be converted to AM. Wiring of the kit takes about an hour using simple tools. Complete assembly and operating instructions are included. Suggested retail price is \$145.-Palomar Engineers, Box 455, Escondido, CA 92025.

MAGNET MOUNT, model $K$-350, comprises a heavy-duty, professional-grade magnet base with pre-assembled RG-58U cable and fittings to accommodate virtually all standard base-loaded or low-profile whips. The assembly will accommodate either $5 / 3-24$ threadbase-loaded whips or, with an adapter (furnished), $11 / 0-18$ thread. Cable


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length is 12 feet, ample for rooftop vehicle installations, and is furnished with a pre-assembled PL-259 standard RF connector. The suggested retail price is $\$ 28.75$.-The Antenna Specialists Co., Professional Division, 1234 Euclid Ave., Cleveland, OH 44106.

AUTOMATIC ANTENNA TUNER, Auto Track model AT2500 is for use with amateur, commercial, and government communications systems. It can handle power in excess of 2500 PEP over a frequency range continuous from 3 to 30 MHz ,


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and the average automatic tune-up time is 15 seconds. Front-panel switch positions permit use continued on page 82

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## RADIO PRODUCTS

continued from page 80
of three coaxial antenna outputs, one long-wire antenna, and one coaxial tuner bypass. The impedance is $10-300$ ohms. A direct-reading SWR meter on the front panel is calibrated from 1.1 to infinity.

The front-panel power meter displays RMS power with continuous (CW) carrier and automatically displays the peak power when in the SSB mode in ranges of 0-250 W and 0-2500 W.

The price of the model AT2500 is $\$ 698.00$. -Bell Industries, J. W. Miller Division, 19070 Reyes Ave., P.O. Box 5825, Compton, CA 90224.

VARIABLE FILTER, the Varifilter can be set to maximize one signal (peaking) or to minimize an interfering signal (notching); it works with CW (Morse), single-sideband, and AM signals, without ringing, oscillation, or instability. The bandwidth is variable from under 30 Hz to over 1 kHz .


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The Varifilter has its own internal power supply, which is switchable from 115 to 230 VAC, and can also run from 12 to 18 VDC. Each unit has a tuning eye that lets the operator see when he has filtered the signal he wants to.

The Varifilter is priced at $\$ 139.95$.-Kantronics, 1202 E .23 rd Street, Lawrence, KS 66044.

## UNICORN-

continued from page 66
et. Table 2 will also help you with that.
Because it is possible to insert the boards into the sockets backwards, we recommend that you use a marker pen or nail polish to indicate the finger-1 edge of the board and the pin-1 end of the socket. If the boards are removed and then replaced, lining up the marks will prevent embarrassing accidents.

## Things to come

Because the radio/computer-control section of the robot involves so many parts, it is impossible to present everything in one section and make the transition from a cable-controlled robot to a radio-controlled one in a single jump.

The circuit described here, though, can be checked out by disconnecting the switches in the command console from the 12 -volt supply, and providing them with five volts, instead. The motor and solenoid wiring inside the robot, which has served us well, can now be connected to the relays via the pins on the relayboard socket. Refer to Tables 1 and 3 in Part 7. The five-volts from the commandconsole switches can now be supplied, via the existing umbilical cable.

Next month we'll present the radiocontrol tone-encoding and decoding circuits, and Unicorn-1 will be able to cut its apron strings.

R-E

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## LASER VIDEODISC

continued from page 69

Demodulator I module. The output of that demodulator is the left-channel audio signal in the case of a stereo-encoded disc. That left-audio signal is applied to the an electronic switch on the Sound Demodulator I module. The 2.8MHz FM demodulator on the other sound module also receives the soundFM signal and the demodulated rightchannel audio signal is also applied to an electronic switch. Each switch is controlled by a DC voltage to mute the respective sound channel. The switches are shown in their normal position, with neither channel muted. In addition to passing through the electronic switch and out to the left-audio output jack, the left-audio signal passes through pin 7 of the module and into the adder circuit on the second-channel module. The right-channel audio signal is also applied to that adder circuit so that the output of the adder is an $\mathrm{L}+\mathrm{R}$ audio

## Dropout detection circuit

The $8.1-\mathrm{MHz}$ video-FM signal from the high-frequency processor module is applied to the $8.1-\mathrm{MHz}$ video FM amplifier through pin 3 on the Dropout Detector Module, as shown in Fig. 6. The trap removes any residual $2.3-\mathrm{MHz}$ signal present at that point from the sound carrier. The amplifier supplies two outputs: one at pin 15 and one at pin 17. The latter output passes through a 64 -microsecond delay line and is therefore delayed by the time it takes for one horizontal-scan line to occur.
The undelayed video FM is applied to the Dropout Detector. The output of that stage is a voltage at pin 7. If the high-frequency signal drops out momentarily, the pin- 7 output goes to 5 volts DC. The drop-out adjust potentiometer R3024 sets the actual amount of time the high frequency must be absent before the detector responds. That output at pin 7 is called the High Frequency Identification voltage because the DCvoltage level is an identification of


FIG. 8-A COMPOSITE video signal is supplied to the reference control and the video processor modules.
signal. That sum signal is applied to the RF modulator to modulate the soundcarrier portion of the VHF (Channel 3 or 4) signal.

If the left channel is muted at the front panel of the player, the voltage at pin 15 of Sound Demodulator 1 module goes high (to 5 volts) and switches the electronic switch to position B. The left-audio signal now goes nowhere, while the right-audio signal is passed to the left-audio out jack and, through pin 7 , to the adder. Thus, the right-channel audio signal is present at both adder inputs and at both rear-panel audio-output jacks.
whether or not high frequency is present. That identification voltage is used later on in other circuits.

The video-FM and the delayed videoFM signals are each applied to a demodulator module as shown in Fig. 7. Video Demodulator I module receives the undelayed video-FM signal and applies it to the demodulator stage at pin 15. The composite video output of the demodulator is present at a test point on pin 1 and is applied to a video amplifier. That amplifier has a frequency response out of 4.2 MHz . The gain of the amplifier is controlled by continued on page 108


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# service ollinic 

- The time constant is the key to recognizing faults in oscillator circuits.

AN OLD PROBLEM HAS COME UP AGAINincorrect time-constants in critical TV sweep-circuits. All circuits are critical, of course, but the two sweep oscillators are even more so. If the basic problem is recongized when you begin the diagnosis, it's a lot easier all around.

The symptom is that the oscillator is running, but at the wrong frequency. Vertical oscillators cause the display of two pictures, or an overlapped picture which can look almost like the first. If you see two complete pictures, but only half-height, the vertical oscillator is running at half-frequency -30 Hz . If you see two overlapped pictures, the oscillator's running at double frequency, or 120 Hz . In all cases, that is due to an incorrect time-constant in the frequency-controlling circuitry. In the multivibrator types so popular a while ago, that was an $\mathrm{R}-\mathrm{C}$ time constant and resistor drift or capacitor leakage threw it off. In some of the new sets, however, oscillator transformers are coming back.

In the case that triggered this column, a Quasar TS-938, a two-winding transformer is used. The primary winding goes to the collector, and secondary to the base of the transistor. There is a resonating capacitor and some resistors across the secondary. That is shown in Fig. 1. The

time-constant involves all of the components: the secondary winding, the capacitor, and the resistors. All of those should be checked.

The symptom here was two complete

## JACK DARR, SERVICE EDITOR



FIG. 2
pictures. The oscillator frequency was checked on a scope and it was 30 Hz , thus the oscillator was running slow. That is a valuable clue. A low frequency tells us that one of the values is far too large so a component must have increased in value. We have three possibilities: inductance, capacitance, and resistance. It is quite * impossible for an inductor to go up in value. The same is true for a paper capacitor. So, we can eliminate those. That leaves us with resistance. And indeed it's quite possible for a resistor to increase in value. There are three resistors used: R13 ( 47 K ohms); R14 ( 10 K ohms); and the 500 -ohm hold control. Checking those out showed that R13 had gone up to more than 100 K ohms. After replacing it, the problem cleared up. You can see the reasoning that eliminated most of the components, and left us with the only ones that could produce the indicated fault. Incidentally, the DC voltages around the stage were very close to normal so, in this case they weren't of much help. But the picture on the screen plus confirmation with the scope gave us the needed information very quickly.

In similar cases with older sets, a slightly leaky capacitor in the feedback loop (output-stage plate back through an R-C network to the input grid) can upset the time constant and cause off-frequency operation. In a great percentage of those cases, the culprit was the last capacitor in the loop going to the input grid. Here, it affected the oscillator time-constant. By the same reasoning, a leaky coupling capacitor from the sync-separator to that grid can also upset things.

And the same thing can happen in the horizontal-oscillator stages too. In the cathode-coupled (emitter-coupled) circuit, where a tapped coil is used in the grid-cathode (base-emitter) circuit, there will be resonating capacitors across the coil. There is also a coupling capacitor from the top of the coil to the input. All of those capacitors are critical. Even very small leakage in any of them can cause trouble.
The horizontal AFC circuit can cause off-frequency operation, in some cases. That can be mistaken for a time-constant problem in the oscillator. Luckily, there is a very simple test for it: Just kill the AFC, and see if you can make the oscillator run on-frequency with the hold control. If so, the oscillator is OK and you have to deal with the AFC.
You do have some time-constant circuits in the horizontal AFC, by the way. Note the filtering circuitry between the AFC diodes and the oscillator base in Fig. 2. You can get a real headache if some of those parts have gone off-value.
There is an electrolytic capacitor and that can go down in value. One function of the filter network is to remove all sync signals from the AFC-control voltage. Failure will allow the sync to get into the oscillator control, and the result will be a "mysterious" horizontal jitter. Here again, don't rely on DC-voltage readings exclusively. The scope will show you whether there is any unwanted signal on the oscillator base, and that is a key clue.

Service questions are on page 86

# Radio- 

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## service questions

## NO HIGH VOLTAGE

This GE 15XB has no high voltage. All of the DC voltage supplies are OK. There is no collector voltage on Q253 (horizontal output); l've checked all of the things I can find, no result. Any ideas?-C.C. Rio Grande, PR.

With no collector voltage on the horizontal output transistor, you'll never get high voltage! The fuse doesn't blow, so something is open. Go to the +131 -volt source and start checking from there back to the collector of Q253. You'll find it somewhere in there.
(Feedback-R269 was open. That is 10 -ohm 2 -watt resistor from the +131 volt source to the flyback winding. I thought I was checking R269 but it was R266!)

## INTERMITTENT VERTICAL

Complaint was no vertical sweep in this RCA EC338W. There was just a horizontal line. Fiddling with the vertical size and vertical linearity controls brought the sweep back and the raster filled the screen. Deciding that something in there is heat-sensitive, I went to work. I found nothing conclusive; sometimes it worked, sometimes it wouldn't. But I did use up
three cans of freeze spray! Finally got down to looking for odd things. There was some discoloration on the PC board near the height control. For lack of a better idea, I sprayed that area and the vertical sweep came back! Scraping the PC board clean and replacing the PC conductors from the height control to the pins with wires fixed everything.
Thanks to David A. Day of Apalachicola, FL for that one.

## BREAKER TRIPS

I've got one in this RCA CTC-59XD that I can't pin down. The circuit breaker trips in three minutes. Seems to be in the highvoltage section. If I pull the MAG001 hori-zontal-oscillator module, the breaker stays in. Replaced MAG001 and trace SCR; no help. Where do I go from here?V.S. Dalton, PA.

If the breaker doesn't trip for three minutes, it looks as if something is heating up and breaking down. If the breaker holds with no drive to the high-voltage section, I'll take that idea as OK. While the breaker is still holding, try spraying coolant on things around the high-voltage section. That might give you some help. We've had problems in some of those sets with driver transformer T403 breaking down.
(Feedback-Thanks! It was T403. I sprayed it with coolant and that worked. Replaced it and things are fine.) R-E


STATE-OF-THE-ART/ DO-NOTHING continued from page 62

Although CMOS gates are used for four IC's, the EPROM current requirements are high and the "do-nothing" box needs a regulated 5 -volts at about 150 mA . I used a "battery eliminator" DC supply of the type used for calculators and cassette recorders. Those are often available inexpensively as discontinued items at surplus or discount outlets. Their regulation is poor and a 6 -volt eliminator at 200 mA delivers closer to 8 volts and works well here. Batteries are an impractical power source for 150 mA .

Construction is straightforward. Point-to-point or wire-wrap will work best. A printed circuit board would be very complex because of the IC pinouts. You could substitute TTL IC's for the CMOS ones with some component changes and an increase in powersupply current. Use a small heat sink on the voltage regulator. I mounted the LED's on the circuit board and replaced the box cover with red plastic. You can mount the LED's separately or use transistor drivers and incandescent lamps instead. Whatever type of display you use, you will have a truly (for now) state-of-the-art "do-nothing" box. R-E

## REDUCING DISTORTION

continued from page 54
so that the amplifier produces its rated power output for each of the selected impedances.

## Layout and construction

Figure 6 shows the internal layout of the newly developed amplifier. The previously described power-supply section of the amplifier accounts for approximately half of the total cubic space inside the unit. Also visible in the photo is a highly efficient cooling system that uses a heat pipe with a related blower fan. A linear torque motor similar to the type used for turntables and tape decks, is used to drive the fan for silent operation.

Since the power stage of this amplifier does not have a negative-feedback loop, the quality of the materials used in all components-and the quality of their performance-has a close relationship to tonal quality, according to the designers of the new amplifier. The power block and the heat-dissipation fins, for example, are of extremely rugged construction. Boron nitride has been used as the material for the transistor insulators. That material has a thermal conductivity rate of the same
order as metal, thereby resulting in outstanding heat conductivity and dissipation for the power devices.

Because the amplifier is monophonic, there is, of course, no risk of crosstalk between channels. Nevertheless, any interference between the first stages and the output stages will still impair tone quality. Therefore, the layout has been planned so that the signal travels the shortest possible distance from input to output.

More technically oriented readers may be interested in additional details concerning the two new power MOSFET's that play such an important part of this new amplifier design.

In Figs. 7 and 8 we have reproduced transfer curves of the 2 SK 173 (N-channel) and its complementary 2SJ54 (Pchannel) MOSFET's while in Fig. 9 we show the input waveform and output signal obtained for a squarewave having a repetition rate of 1600 nanoseconds (equivalent to a frequency of $625,000 \mathrm{~Hz}$ )!

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## CAR ELECTRONICS

continued from page 51
and shut down "chop shop" operations. It can even be used to keep the parts department from ordering the wrong part for your particular engine.
Service facilities are also increasingly dependent on thoroughly detailed engineering data from the manufacturers. Diagnostic and repair manuals are becoming much more graphic and their language more readable. But the more information that's available on each car, and the more cars, the harder it gets to store all that paper. Microfiche and other space-saving methods are being used, and a number of computer software packages are being developed.

Tools are also getting smarter. GM is developing an Automotive Service Analyzer, an advanced version of a cooling-system tester first developed in 1976. In its eventual form, it will be capable of diagnosing problems with air conditioning, cooling, electrical, and automatic transmission systems, and then will give repair instructions in English on an alphanumeric display.
The magic, of course, is performed by a built-in microprocessor. The tester is attached to the vehicle and the mechanic dials up the tests he wants to perform. The Automotive Service Analyzer walks him through a test procedure and gives him a diagnosis of the problem.
When development of the Automotive Service Analyzer is complete, GM will continue its policy of making technological advances like this available to qualified manufacturers of service tools and equipment, free of charge. That's how GM's SOS (Service Order Scheduling) system-developed by GM Service Research and Delco Electronics Division-became available through several commercial computer-system companies. SOS controls the work flow in a large service organization by scheduling and routing the flow of repair orders. That helps reduce shop inefficiency saving everybody time and money.

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100 W CLASS A

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Dynamic Bias Class "A circuit design makes this unit unique in its class. Crystal clear, 100 watts power output will satisfy the most picky fans. A perfect combination with the TA-1020 low T.I.M. ste-
reo pre-amp.
Specifications:
Output power: 100W RMS into 8 -ohm
125W RMS into 4 -ohm
Frequency response: 10 Hz - 100 KHz
T.H.D.: less than 0.008\%

S/N ratio: better than 80 dB
Input sensitivity: IV max
Power supply: $\pm 40 \mathrm{~V}$ @ 5 amp


TA-1000 KIT $\$ 51.95$ Power transformer $\$ 18.00$ each

REGULATED VARIABLE D.C. POWER SUPPLY KIT

Uses UA723 I.C. and 2N3055 power transistor as regulator. Output voltages can be adjusted from or 30V at an internal resistance of less than 0.005 ohm, ripple and noise less than 1 MV ; with built on board LED and audible overload indicator. Kit comes with P.C. board; all meters for voltage and amp: a professional look metal cabinet and instructions

Model TR-88A OM5V D.C. 3 amp Model TR-88B O 30 V D.C. 2 amp

$\$ 59.50$
per kit
WHISTLE ACTIVATED SWITCH BOARD

All boards are pre-assembled and tested. Your whistle to is FET condenser microphone from a distance, as far as 30 feet away (sensitivity can be easily adjusted) will turn the switch on, then latched you whistle to it again then it turns oft. Ideal for remote control toys, electrical appliance such as lights, coffee pots, TV. Hi-FI, radio or other projects. Unit works on 9V D.C

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\begin{aligned}
& \text { Model } 968 \\
& \$ 4.50 \text { each }
\end{aligned}
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## SUB MINI SIZE FET

 CONDENSER MICROPHONE

## Specification

Sensitivity: $-65 \mathrm{~dB} \pm 3 \mathrm{~d}$ FEQ. Response: $50 \mathrm{~Hz} \quad 8 \mathrm{KHz}$ Output Impedance: 1 K ohm max Polar Pattern: Omni-directional Power Supply: $1.5 \mathrm{~V} \quad 10 \mathrm{~V}$ D.C. Sound Pressure Level: Max. 120 dB EM4RP $\$ 2.50$ ea. or 2 for $\$ 4.50$ 9NEW MARK III
$=\begin{aligned} & \text { 9Steps 4 Colors }\end{aligned}$
LED VU
Stereo level indicator kit with arc-shape display panell!!! This Mark III LED level indicator is a new design PC board with an arc-shape 4 colors LED display (change color from red, yellow, green and the peak output indicated by rose). The power range is very large, from -300 B to +5 dB . The Mark III indicator is applicable to 1 watt - 200 watts amplifier operating voltage is $3 \mathrm{~V}-9 \mathrm{~V}$ D at max 400 MA . The circuit uses 10 LEDs per channel. It is very easy to connect to the amplifier. Just hook up with the speaker output!

IN KIT FORM $\$ 18.50$

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Pre assembled units. All you need is to hook up the speaker and the volume control. Supply voltage from $9 \sim$ 15V D.C. measures only $2^{\prime \prime} \times 31 / 2^{\prime \prime}$, making it good for portable or discrete applications. Comes with hook up data.


BUY 2 FOR $\$ 4.99$

MARK IV 15 STEPS LED POWER LEVEL INDICATOR KIT

## This new stereo level indicator kit consists of 364 -

 color LED ( 15 per channel) to indicate the sound level output of your amplifier from $-36 \mathrm{~dB} \sim+3 \mathrm{~dB}$. Comes with a well-designed silk screen printed plastic panel and has a selector switch to allow floating or gradual output indicating. Power supply is $6 \sim$ 12V D.C. with THG on board input sensitivity controls. This unit can work with any amplifier from 1 W to 200W!Kit includes 70 pcs. driver transistors, 38 pcs. matched 4 -color LED, all other electronic components, PC board and front panel

## =1:浣 <br> MARK V 15 STEPS LED POWER OUTPUT INDICATOR KIT

All functions same as Mark IV but this is with heavy duty aluminum front plate and case. Can be easily slot into the front panel of your auto, truck or boat. Operates on 12 V DC


BATTERY POWERED FLUORESCENT LANTERN
MODEL 888 R
FEATURES
Circuitry: designed for operation by high efficient, high power silicon transistor which enable illumination maintain in a standard level even the battery supply drops to a certain low voltage. $9^{\prime \prime} .6 \mathrm{~W}$ cool/daylight miniature fluorescent tube.
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$\$ 10.50 \mathrm{EA}$
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It works in 12 V DC as well! Kit includes 1 PC SANYO STK-043 stereo power amp. IC LM 1458 as pre amp, all other electronic parts. PC Board, all contro pots and special heat
sink for hybrid. Power sink for hybrid. Power cluded. It produces ultra cluded. It produces ultra hi-f output up to 60 watts ( 30 watts per channel) yet gives out
less than $0.1 \%$ total harless than $0.1 \%$ total harmonic distortion betw
100 Mz and 10 KHz .
5W AUDIO AMP KIT
2 LM 380 with Volume Control er Suply $6 \quad 18 \mathrm{~V} D$ ONLY $\$ 6.00$ EACH

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All meters white face with black
Tivoe MU-52E scales. Plastic cover

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ASSEMBLED! NOT A KIT!
Features: • 4 digits $0.5^{\prime \prime}$ LED Displays - 12 hours real time format - 24 hours alarm audio output - 59 min. countdown timer - 10 min . snooze control.
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SANYO HYBRID AUDIO
POWER AMPLIFIER I.C.


## PROFESSIONAL FM

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Tuning voltage $+1 \mathrm{~V} \sim+28 \mathrm{~V} D \mathrm{C}$. Input impedance 75 OHM. I.F. band width $7 \sim 16 \mathrm{MHZ}$. Noise figure 11.5 dB MAX. Size $25 \%^{\prime \prime} \times 11_{4}^{\prime \prime} \times 3 / 4^{\prime \prime}$ Supply voltage 15 V DC. Sound I.F. $=58.0 \mathrm{MHZ}$. Video I.F. $=62.5 \mathrm{MHZ}$


All units are brand new from Sanyo. MODEL 115-B-405A $\$ 35.00 \mathrm{EACH}$

Kit includes the Ultra Sonic Transducers, 2 PC Boards for transmitter and receiver. All electronic parts and instructions. Easy to build and a lot of uses such as counter. Unit operates by 9.12 DC. $\$ 15.50$
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Touch On Touch Off uses 7473 I.C. and
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1 WATT AUDIO AMP
All parts are pre-assembled on a
mini PC Board. Supply Voltage 6
mini PC Board. Supply Voltage 6
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## CB EQUIPMENT

LIMITED quantity of Motorola 40-channel remote CB radios w/o mikes. These are a closeout of new D84Y18B806A radios made for 78-79 Lincoln/ Mercury/Ford automobiles. Shipping is prepaid and orders received will be shipped within 24 hours. Send cashier's check or money order for $\$ 70$ to GUY GRIDER CB, 220 SE 89th, Oklahoma City, OK 73149

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The CT-90 is the most versatile, feature packed counter available for less than $\$ 300.00$ ! Advanced design features include, three selectable gate times, nine digits, gate indicator and a unique display hold function which holds the displayed count after the input signal is removed Also, a 10 mHz TCXO time base is used which enables easy zero beat calibration checks against WWV. Optionally, an internal nicad battery pack,external time base input and Micropower high stability crystal oven time base are available. The CT-90, performance you can count on!

SPECIFICATIONS:
Range: $\quad 20 \mathrm{~Hz}$ to 600 MHz
Sensitivity: Less than 10 MV to 150 MHz
Less than 50 MV to 500 MHz
0.1 Hz ( 10 MHz range) $1.0 \mathrm{~Hz}(60 \mathrm{MHz}$ range) 10.0 Hz ( 600 MHz range)

Display: $\quad 9$ digits $0.4^{\prime \prime}$ LED
Time base: Standard- $10.000 \mathrm{mHz}, 1.0 \mathrm{ppm} 20-40^{\circ} \mathrm{C}$ Optional Micro-power oven- $0.1 \mathrm{ppm} 20-40^{\circ} \mathrm{C}$ Power. $\quad 8-15$ VAC@ 250 ma

## 7 DIGITS 525 MHz \$99 $\frac{95}{\mathrm{w}}$ WIRED

SPECIEICATIONS
Range. $\quad 20 \mathrm{~Hz}$ to 525 MHz
Sensitivity. Less than 50 MV to 150 MHz
Less than 150 MV to 500 MHz
Resolution $\quad 1.0 \mathrm{~Hz}$ ( 5 MHz range) 10.0 Hz ( 50 MHz range) 100.0 Hz ( 500 MHz range)

Display: $\quad 7$ digits $0.4^{\prime \prime}$ LED
Time base. $\quad 1.0 \mathrm{ppm}$ TCXO $20-40^{\circ} \mathrm{C}$
Power. $12 \mathrm{VAC} @ 250 \mathrm{ma}$

The CT-70 breaks the price barrier on lab quality frequency counters. Deluxe features such as, three frequency ranges - each with pro- amplification, dual selectable gate times, and gate activity indication make measurements a snap. The wide frequency range enables you to accurately measure signals from audio thru UHF with 1.0 ppm accuracy- that's $.0001 \%$ ! The CT-70 is the answer to all your measurement needs, in the field, lab or ham shack

PRICES:
CT-70 wired, 1 year warranty $\$ 99.95$ CT-70 Kit, 90 day parts warranty
84.95 BP. 1 Nicad pack + AC adapter/charger

## $\begin{array}{r}\text { ron } \\ -20 \\ \hline\end{array}$

PRICES:
MINI-100 wired, 1 year
warranty
MINI-100 Kit, 90 day part warranty AC-Z Ac adapter for MINI100
BP-Z Nicad pack and AC adapter/charger

# 7 DIGITS 500 MHz 

Here's a handy, general purpose counter that provides most counter functions at an unbelievable price. The MINF-100 doesn't have the full frequency range or input impedance qualities found in higher price units, but * for basic RF signal measurements, it can't be beat' Accurate measurements can be made from 1 MHz all the way up to 500 MHz with excellent sensitivity throughout the range, and the two gate times let you select the resolution desired. Add the nicad pack option and the MINI-100 makes an ideal addition to your tool box for "in-the-field" frequency checks and repairs.


## 8 DIGITS $600 \mathrm{MHz} \$ 159$

SPECIFICATIONS: Range: $\quad 20 \mathrm{~Hz}$ to 600 MHz Sensitivity:
Resolution
Display.
Display.
Time base
Power.

20 Hz to 600 MHz
Less than 25 mv to 150 MHz Less than 150 mv to 600 MHz 1.0 Hz ( 60 MHz range) 10.0 Hz ( 600 MHz range) 8 digits $0.4^{\circ}$ LED 8 digits $0.4^{\circ}$ LED
$2.0 \mathrm{ppm} 20-40^{\circ} \mathrm{C}$ 110 VPAC or 12 VDC

The CT-50 is a versatile lab bench counter that will measure up to 600 MHz with 8 digit precision. And, one of its best features is the Receive Frequency Adapter, which turns the CT-50 into a digital readout for any receiver. The adapter is easily programmed for any receiver and a simple connection to the receiver's VFO is all that is required for use. Adding the receiver adapter in no way limits the operation of the CT-50, the adapter can be conveniently switched on or off. The CT-50, a counter that can work double-duty? ${ }^{-}$

PRICES:
CT-50 wired, 1 year warranty $\$ 159.95$ CT-50 Kit, 90 day parts warranty RA-1, receiver adapter kit RA-1 wired and pre programmed (send copy of teceiver schematic)

| SPECIFICATIONS: |  |
| :--- | :--- |
| Range: | 1 MHz to 500 MHz |
| Sensitivity: | Less than 25 MV |
| Resolution: | 100 Hz (slow gate) |
|  | 1.0 KHz (fast gate) |
| Display: | 7 digits, $0.4^{\prime \prime} \mathrm{LED}$ |
| Time base: | $2.0 \mathrm{ppm} 20-40^{\circ} \mathrm{C}$ |
| Power: | 5 VDC © 200 ma |

# DIGITAL MULTIMETER \$99 $\frac{95}{\mathrm{w}}$ 

The DM-700 offers professional quality performance at a hobbyist price. Features include; 26 different ranges and 5 functions, all arranged in a convenient, easy to use format. Measurements are displayed on a large $31 / 2$ digit, $1 / 2$ inch LED readout with automatic decimal placement, automatic polarity, overrange indication and overload protection up to 1250 volts on all ranges, making it virtually goof-proof? The DM-700 looks great, a handsome. jet black, ruged ABS case with convenient retractable tilt bail makes it an ideal addition to any shop.

SPECIFICATIONS:
DC/AC volts: 100 uV to $1 \mathrm{KV}, 5$ ranges DC/AC Nemvorial current $\quad 0.1$ A to $2.0 \mathrm{Amps}, 5$ ranges Resistance 0.1 ohms to 20 Megohms 6 ranges Input
impedance 10 Megohms, DC/AC volts Accuracy. $\quad 10.1 \%$ basic DC volts Power $\quad 4^{\prime} \mathrm{C}$ cells

## AUDIO SCALER

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ACCESSORIES

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## COUNTER PREAMP

- Great for sniffing RF with pick-up loop $\$ 34.95 \mathrm{Kit} \$ 44.95$ Wired

LASER VIDEODISC<br>continued from page 83

R3013. Output at pin 4 is applied through a 540 -nanosecond delay line, DL5103, to pin 4 of the Video Demodulator II module. Here the composite video signal is applied to point A of the electronic dropout switch.
The 64 -microsecond delayed videoFM signal is applied via pin 17 to the FM demodulator stage of the Video Demodulator II module. Its composite video output is available at Pin 1 and is applied to a video amplifier. The frequency response of that amplifier rolls off at 2.6 MHz because the 64 -microsecond delay line (DL5101 in Fig. 6) cannot pass signals above that frequency. As a result, dropout corrections occur in black-and-white only. The effect, however, is not noticeable on the TV screen because of the small areas of the picture involved. Gain control R3013 controls the gain of the video amplifier. The video input is applied to point B of the dropout switch and can be monitored at the test point on Pin 10 of the Video Demodulator II module.

The electronic-dropout switch therefore receives a 540 -nanosecond delayedvideo signal at point A and a line-delayed composite video signal at point B (64microsecond delay). The voltage at pin

15 determines which of those signals will be applied to the electronic burst switch. When a dropout occurs, the voltage at pin 15 goes high and creates a dropout pulse for the duration of the dropout. The dropout switch, normally in position A, receives the undelayed video signal ( 540 nanoseconds). However, when a dropout pluse is present, the dropout switch moves to position B and receives the delayed ( 64 -microsecond) video signal. The selected signal leaves the module at pins 7 and 11.

The undelayed video signal is actually delayed by 540 nanoseconds- the time required for the dropout circuitry to respond to the actual dropout. The actual time difference between the two types of video signals is 64 microseconds less 540 nanoseconds, or approximately 63.5 microseconds, the true scan time of one horizontal raster line.

The electronic burst switch in the Video Demodulator I module is used to maintain the 180 -degree phase difference in the $3.58-\mathrm{MHz}$ chroma (color) signal from frame to frame during special mdoes of operation, such as still-picture, reverse, fast forward, etc. The chroma signal is normally 180 degrees out of phase from track to track on the videodisc. That relationship is also true from frame to frame on normal TV broadcasts and its purpose is to cancel 3.58 MHz interference. However, when the


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same track is being played over and over again, the chroma signal would not be out of phase from frame to frame. To introduce that phase difference, the signal is therefore delayed by 140 nanoseconds during alternate revolutions when still-picture viewing is called for. That delay is equivalent to one half of a $3.58-\mathrm{MHz}$ period. The electronic burst switch switches the delay line DL5102 in or out depending upon the viewing mode being used (still picture, reverse play, fast forward, etc.)
The resultant composite video signal is then fed to a Video Processor module and to a Reference Control module, as shown in Fig. 8. A video-muting input at pin 16 blanks the video signal during return of the laser beam to the inside of the disc and during initial turn-on. The video signal is also processed by the DC clamp circuit that clamps the video signal to the correct DC level.

Composite video is clipped in the Reference Control module by the amplifier/clipper stage. The resulting clipped video contains the digital code that represents the picture or frame number. That code is applied to the decoder block on the Mode Control module. A video generator actually creates the separate video signal for displaying the picture number. A video signal that provides a gray background behind the numbers on the screen is also generated by those circuits. Those video signals are then applied to the 2nd video amplifier at pins 7 and 11 of the Video Processor module and the resultant video is then applied to the Rf Modulator as well as to the rearpanel monitor jack, via an emitter follower.

As elaborate as that video and audio processing circuitry may seem, it is really only one half the story of what takes place inside the new optical-laser video disc players. Fully as much electronic circuitry (not to mention the precision mechanical system) is associated with the servo-control circuitry that performs such functions as turntable motor control, tangential-tracking mirror control, radial-tracking mirror control, slide-drive control, objective-lens control (focus) and turn-on sequence logic. All of which makes the new videodisc players an almost miraculous achievement when one considers that they don't cost much more than many audiophiles spend for a super-performing audio turntable.

R-E

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The MicroAce is programmed in BASIC, and you can use it to do quite literally anything, from playing chess to managing a business.
The MicroAce is pleasantly straightforward to assemble, using a fine-tipped soldering iron. It immediately proves what a good job you've done: connect it to your TV ... link it to the mains adaptor and you're ready to go.
Fewer chips, compact design, volume production-more power per Dollar!
The MicroAce owes its remarkable low price to its remarkable design: the whole system is packed on to fewer, newer, more powerful and advanced LS chips. A single SUPER ROM, for instance, contains the BASIC interpreter, the character set, operating system, and monitor. And the MicroAce 1 K byte

RAM (expandable to 2 K on board) is roughly equivalent to $4 K$ bytes in a conventional computer - typically storing 100 lines of BASIC. (Key words occupy only a single byte.)
The display shows 32 characters by 24 lines.
And Benchmark tests show that the MicroAce is faster than all other personal computers.
No other personal computer offers this unique combination of high capability and low price.

## The MicroAce teach-yourself BASIC manual.

If the features of the BASIC interpreter mean little to you-don't worry. They're all explained in the specially-written book free with every kit! The book makes learning easy, exciting and enjoyable, and represents a complete course in BASIC programming-from first principles to complex programs. (Available separately-purchase price refunded if you buy a MicroAce later.)
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## NOISE FILTER

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minimum. A further advantage provided by multiple notch-filter techniques, is that the worst possible adjustment is limited to between 8 dB and 10 dB of attenuation. If a sliding cut-off filter were used, that might not be the limit and the error induced could be intolerable.

With the ASRU, the maximum permissible error can be set by the user and the possibility of obtaining an unnatural response-where a lowfrequency band might be attenuated more than a higher-frequency one-is avoided. The noise-reduction control and the spectral-tracking concept with feedback are two features that make the operation of the ASRU so effective, yet free of side effects.

Next month, we will provide the circuit details for both the noise reduction and dynamic range expander portions of the ASRU. The construction details will also be given.

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## MORE QUALITY:

It's not imported. It's not made in Taiwan, Korea or Japan. It's American made in an American town. It's made with better materials that cost more and by professional people we pay more. And we designed it right here in the U.S.A.
*Including optional mounts at extra cost.

## ...This Antenna is so DYNAMITE

 you receive aTimeS: - it's not often that a product bursts onto the market scene, dominates and improves CB'ing for everyone. American Antenna and the K40 are doing it-repeated tests showed the K40 could out-perform the major competitive brands."
RADIO-ELECTRONICS: "The results of our tests showed that, in three different positions of the monitoring receiver, the model K40 equaled or out-performed the competitive antenna. Apparently, American Antenna's advertising is not merely Madison Avenue showmanship.
PERSONAL COMMUNICATIONS:
an impressive
$95 \%$ of the trials, the K40 out-performed the existing mobile antennas. We had to try one for ourselves.
in every case, the K40 either equaled or out-performed its competitor
"No ifs, ands, or buts! The K40 Antenna from American Antenna would have to be just about the best antenna around.
CB MAGAZINE: "Introduced in October, 1977, the K40 quickly became the top seller and in mid 1978, became the number one selling antenna in the nation."

## ...Here's what CB'ers all across the country said.

ANTENNA SPECIALISTS:
truck driver and CB'er for 10 years . . . 50\% further than my M410 'Big Momma' '
$-J . H$. Collett, 207 McFee, Bastrop, LA
AVANTI: "I'm an electronic technician with a Second Class FCC license . . . I was able to transmit $70 \%$ further and tune the SWR 75\% lower than my Avanti.'
-H.R. Castro, VRB, Monserrante D-67, Salinas, Puerto Rico
PAL: ". . . 20\% better in transmission and reception than my $5 / 8$ wave Pal Firestik.'
-John A. Bum, Box 446, Zelienolple, PA
SHAKESPEARE: " . . . I've been a CB'er for three years and the K40 is the best l've ever had. Better in reception and transmission than my Shakespeare."
$-H$. Bachert, Jt., 15 King Rd., Park Ridge, NJ
HUSTLER: "Compared to my Hustler XBLT4, the K40 can consistently transmit 40\% further and the reception was better. The K40 is the perfect way to complete a CB system. -Jerome R. Brown, 7800 S. Linder, Burbank, il
(SPECIAL NOTE) IF YOU'RE A BEGINNER:
Our K40 Dealers will be happy to sell you any of the older style and less expensive antennas that are great bar[r40


[^0]:    To: Sinclair Research Ltd., 475 Main St., P.O. Box 3027, Wallingford, CT 06492.
    Please send me $\qquad$ ZX80 personal computer(s) at $\$ 199.95^{*}$ each (US dollars), plus $\$ 5$ shipping. (Your ZX80 may be tax deductible.)
    I enclose a check/money order payable to Sinclair Research Ltd. for \$
    Name
    Address
    

    ## Intended use of ZX 80 :

    Have you ever used a computer? $\square$ Yes $\square$ No.
    Do you own another personal computer? $\square$ Yes $\square$ No. $\quad$ *For Conn. deliveries, add sales tax.

[^1]:    *Due to the difficulty in obtaining positronic brains, Isaac Asimov's Three Laws of Robotics do not apply here, and we have to use our own judgment, rather than rely on the robot's.

[^2]:    Street/Apartment Number

